

Rail-to-rail high output current quad operational amplifier

Features

- Rail-to-rail input and output
- Low noise: $9 \text{ nV}/\sqrt{\text{Hz}}$
- Low distortion
- High output current: 80 mA (able to drive 32Ω loads)
- High-speed: 4 MHz, $1.3 \text{ V}/\mu\text{s}$
- Operating range from 2.7 V to 12 V
- Low input offset voltage: 900 μV max (TS924A)
- ESD internal protection: 3 kV
- Latch-up immunity
- Macromodel included in this specification

Applications

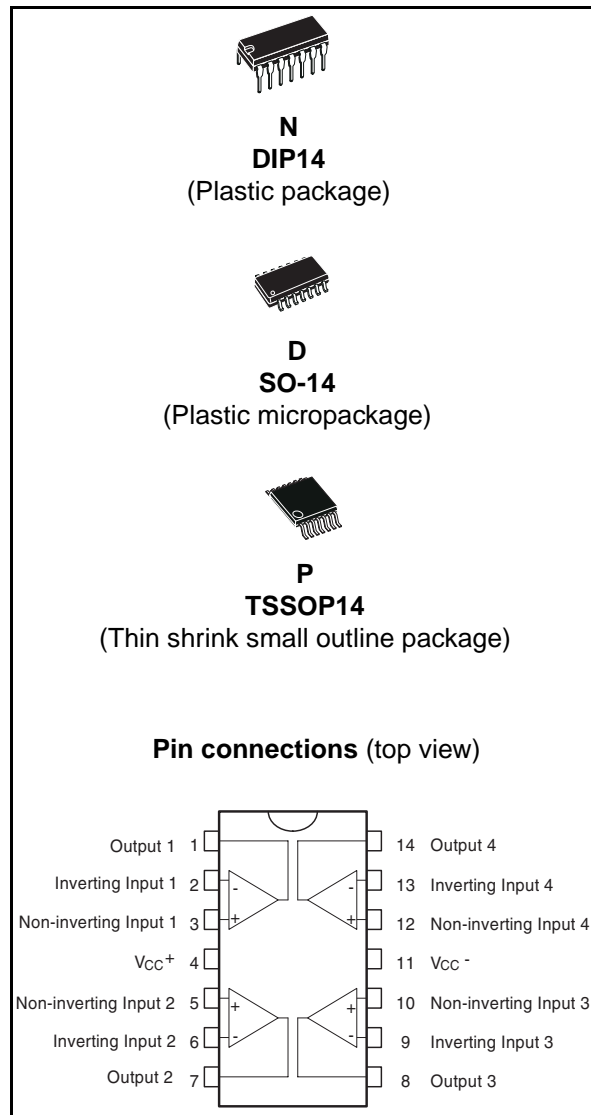
- Headphone amplifier
- Piezoelectric speaker driver
- Sound cards
- MPEG boards, multimedia systems
- Line driver, buffer
- Cordless telephones and portable communication equipment
- Instrumentation with low noise as key factor

Description

The TS924 is a rail-to-rail quad BiCMOS operational amplifier optimized and fully specified for 3 V and 5 V operation.

High output current allows low load impedances to be driven.

The TS924 exhibits a very low noise, low distortion, low offset and high output current capability making this device an excellent choice for high quality, low voltage or battery operated audio systems.



The device is stable for capacitive loads up to 500 pF.

1 Absolute maximum ratings and operating conditions

Table 1. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{CC}	Supply voltage ⁽¹⁾	14	V
V_{id}	Differential input voltage ⁽²⁾	± 1	V
V_{in}	Input voltage ⁽³⁾	$V_{DD} - 0.3$ to $V_{CC} + 0.3$	V
T_{stg}	Storage temperature	-65 to +150	°C
T_j	Maximum junction temperature	150	°C
R_{thja}	Thermal resistance junction to ambient ⁽⁴⁾	103	°C/W
	DIP14	66	
	SO-14 TSSOP14	100	
ESD	HBM: human body model ⁽⁵⁾	3	kV
	MM: machine model ⁽⁶⁾	100	V
	CDM: charged device model ⁽⁷⁾	1	kV
	Output short-circuit duration	see note ⁽⁸⁾	
	Latch-up immunity	200	mA
	Soldering temperature (10 sec), leaded version	250	°C
	Soldering temperature (10 sec), unleaded version	260	°C

1. All voltage values, except differential voltage are with respect to network ground terminal.
2. Differential voltages are the non-inverting input terminal with respect to the inverting input terminal. If $V_{id} > \pm 1$ V, the maximum input current must not exceed ± 1 mA. In this case ($V_{id} > \pm 1$ V), an input series resistor must be added to limit input current.
3. Do not exceed 14 V.
4. Short-circuits can cause excessive heating and destructive dissipation. R_{th} are typical values.
5. Human body model: A 100 pF capacitor is charged to the specified voltage, then discharged through a 1.5 kΩ resistor between two pins of the device. This is done for all couples of connected pin combinations while the other pins are floating.
6. Machine model: A 200 pF capacitor is charged to the specified voltage, then discharged directly between two pins of the device with no external series resistor (internal resistor < 5 Ω). This is done for all couples of connected pin combinations while the other pins are floating.
7. Charged device model: all pins and the package are charged together to the specified voltage and then discharged directly to the ground through only one pin. This is done for all pins.
8. There is no short-circuit protection inside the device: short-circuits from the output to V_{CC} can cause excessive heating. The maximum output current is approximately 80 mA, independent of the magnitude of V_{CC} . Destructive dissipation can result from simultaneous short-circuits on all amplifiers.

Table 2. Operating conditions

Symbol	Parameter	Value	Unit
V_{CC}	Supply voltage	2.7 to 12	V
V_{icm}	Common mode input voltage range	$V_{DD} - 0.2$ to $V_{CC} + 0.2$	V
T_{oper}	Operating free air temperature range	-40 to +125	°C

2 Electrical characteristics

Table 3. Electrical characteristics at $V_{CC} = +3V$ with $V_{DD} = 0V$, $V_{icm} = V_{CC}/2$, $T_{amb} = 25^{\circ}C$, and R_L connected to $V_{CC}/2$ (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
V_{io}	Input offset voltage TS924 TS924A $T_{min} \leq T_{amb} \leq T_{max}$			3 0.9	mV
	TS924 TS924A			5 1.8	
DV_{io}	Input offset voltage drift		2		$\mu V/^{\circ}C$
I_{io}	Input offset current - $V_{out} = V_{CC}/2$		1	30	nA
I_{ib}	Input bias current - $V_{out} = V_{CC}/2$		15	100	nA
V_{OH}	High level output voltage				V
	$R_L = 10k\Omega$	2.90			
	$R_L = 600\Omega$ $R_L = 32\Omega$	2.87	2.63		
V_{OL}	Low level output voltage				mV
	$R_L = 10k\Omega$			50	
	$R_L = 600\Omega$ $R_L = 32\Omega$		180	100	
A_{vd}	Large signal voltage gain ($V_{out} = 2V_{pk-pk}$)				V/mV
	$R_L = 10k\Omega$		200		
	$R_L = 600\Omega$ $R_L = 32\Omega$		35 16		
I_{CC}	Total supply current - no load, $V_{out} = V_{CC}/2$		4.5	7	mA
GBP	Gain bandwidth product - $R_L = 600\Omega$		4		MHz
CMR	Common mode rejection ratio	60	80		dB
SVR	Supply voltage rejection ratio - $V_{CC} = 2.7$ to $3.3V$	60	85		dB
I_o	Output short-circuit current	50	80		mA
SR	Slew rate	0.7	1.3		V/ μs
ϕ_m	Phase margin at unit gain - $R_L = 600\Omega$, $C_L = 100pF$		68		Degrees
G_m	Gain margin - $R_L = 600\Omega$, $C_L = 100pF$		12		dB
e_n	Equivalent input noise voltage - $f = 1kHz$		9		$\frac{nV}{\sqrt{Hz}}$
THD	Total harmonic distortion $V_{out} = 2V_{pk-pk}$, $F = 1kHz$, $A_v = 1$, $R_L = 600\Omega$		0.005		%
C_s	Channel separation		120		dB

Table 4. $V_{CC} = +5V$, $V_{DD} = 0V$, $V_{icm} = V_{CC}/2$, $T_{amb} = 25^{\circ}C$, R_L connected to $V_{CC}/2$
(unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
V_{io}	Input offset voltage TS924 TS924A $T_{min} \leq T_{amb} \leq T_{max}$			3 0.9	mV
	TS924 TS924A			5 1.8	
DV_{io}	Input offset voltage drift		2		$\mu V/^{\circ}C$
I_{io}	Input offset current - $V_{out} = V_{CC}/2$		1	30	nA
I_{ib}	Input bias current - $V_{out} = V_{CC}/2$		15	100	nA
V_{OH}	High level output voltage				V
	$R_L = 10k\Omega$	4.90			
	$R_L = 600\Omega$ $R_L = 32\Omega$	4.85	4.4		
V_{OL}	Low level output voltage				mV
	$R_L = 10k\Omega$			50	
	$R_L = 600\Omega$ $R_L = 32\Omega$		300	120	
A_{vd}	Large signal voltage gain ($V_{out} = 2V_{pk-pk}$)				V/mV
	$R_L = 10k\Omega$		200		
	$R_L = 600\Omega$ $R_L = 32\Omega$		40 17		
I_{CC}	Total supply current - no load, $V_{out} = V_{CC}/2$		4.5	7	mA
GBP	Gain bandwidth product - $R_L = 600\Omega$		4		MHz
CMR	Common mode rejection ratio	60	80		dB
SVR	Supply voltage rejection ratio - $V_{CC} = 3V$ to $5V$	60	85		dB
I_o	Output short-circuit current	50	80		mA
SR	Slew rate	0.7	1.3		V/ μs
ϕ_m	Phase margin at unit gain - $R_L = 600\Omega$, $C_L = 100pF$		68		Degrees
G_m	Gain margin - $R_L = 600\Omega$, $C_L = 100pF$		12		dB
e_n	Equivalent input noise voltage - $f = 1kHz$		9		$\frac{nV}{\sqrt{Hz}}$
THD	Total harmonic distortion $V_{out} = 2V_{pk-pk}$, $F = 1kHz$, $A_v = 1$, $R_L = 600\Omega$		0.005		%
C_s	Channel separation		120		dB

Figure 1. Output short circuit current vs. output voltage

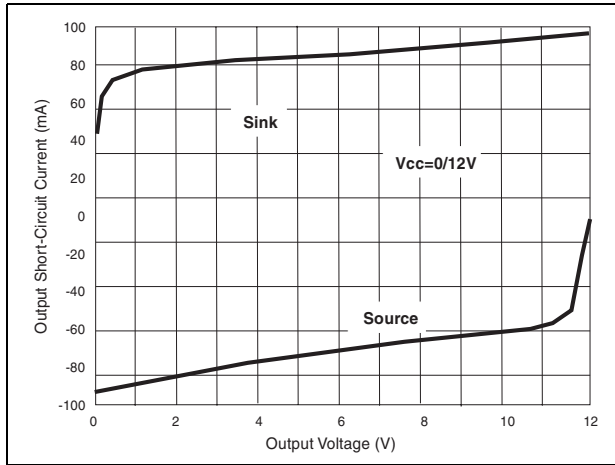


Figure 2. Output short circuit current vs. output voltage

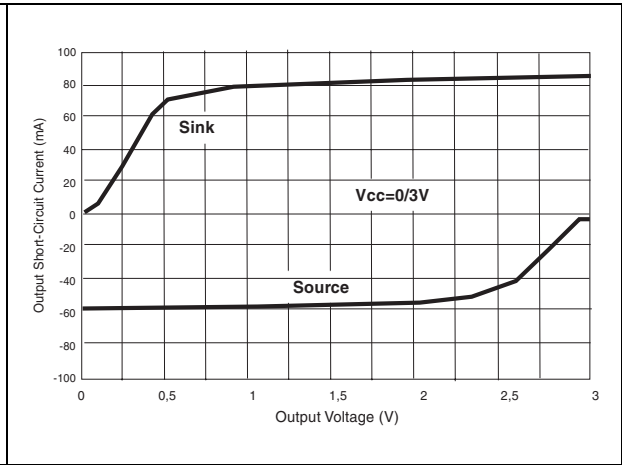


Figure 3. Voltage gain and phase vs. frequency

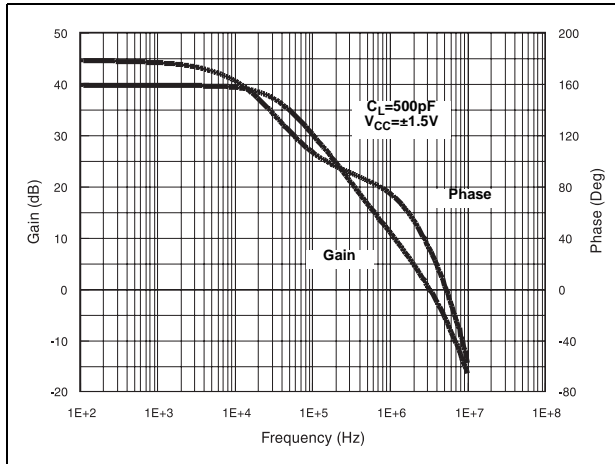


Figure 4. Output short circuit current vs. output voltage

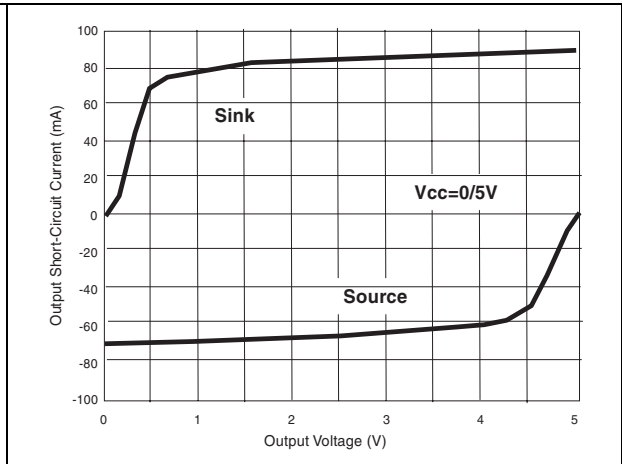


Figure 5. Voltage gain and phase vs. frequency

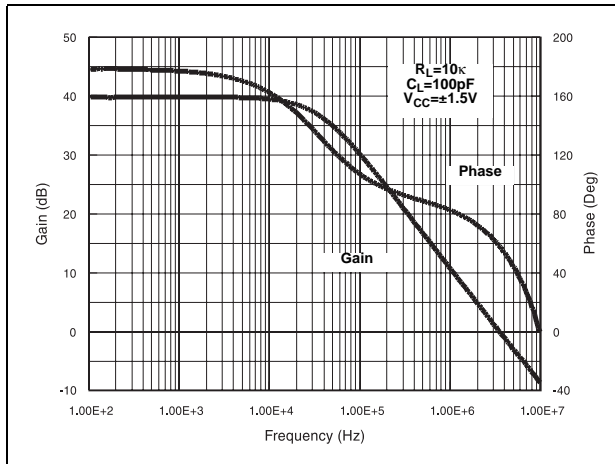


Figure 6. THD + noise vs. frequency

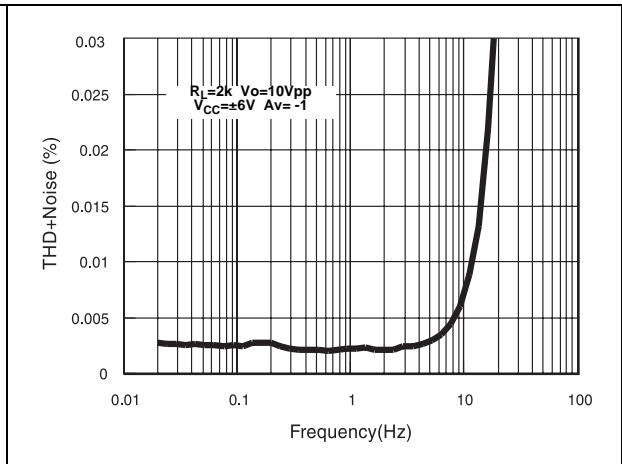


Figure 7. THD + noise vs. frequency

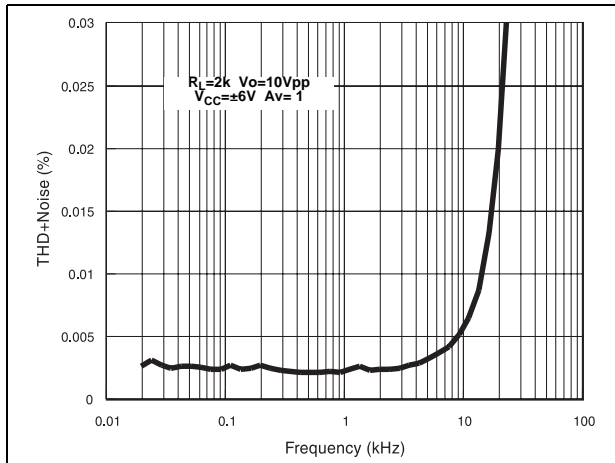


Figure 8. THD + noise vs. frequency

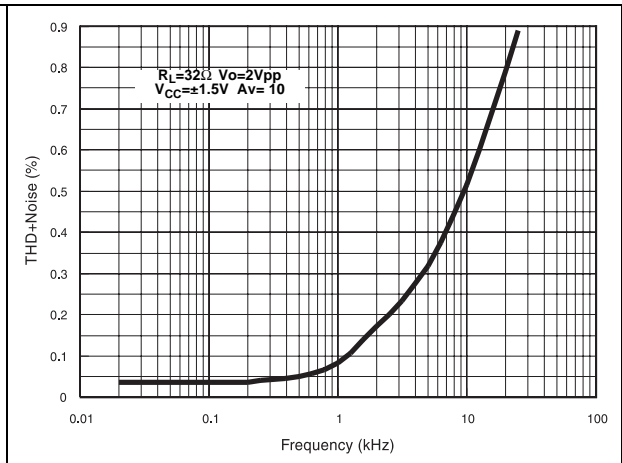


Figure 9. THD + noise vs. V_{out}

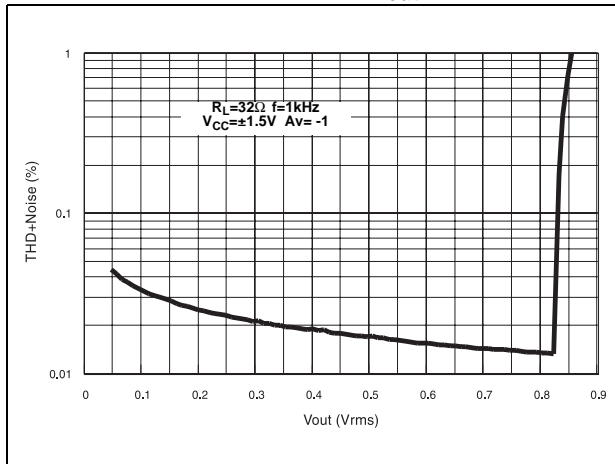


Figure 10. THD + noise vs. frequency

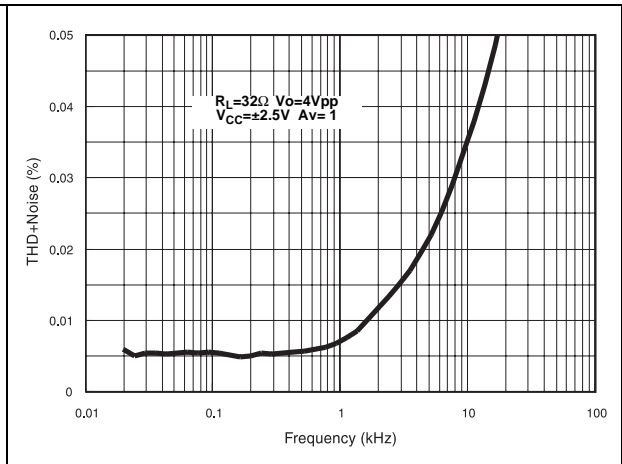


Figure 11. THD + noise vs. V_{out}

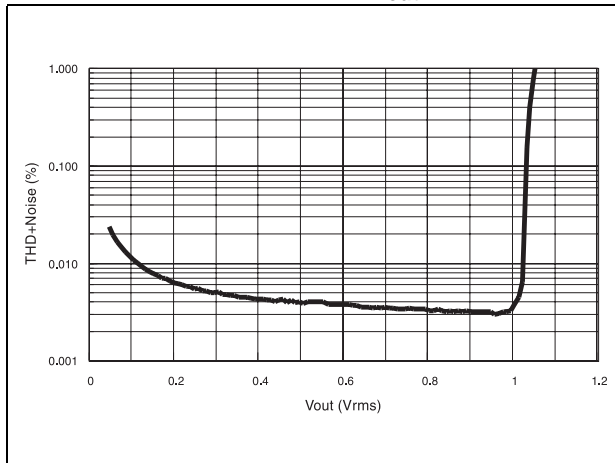
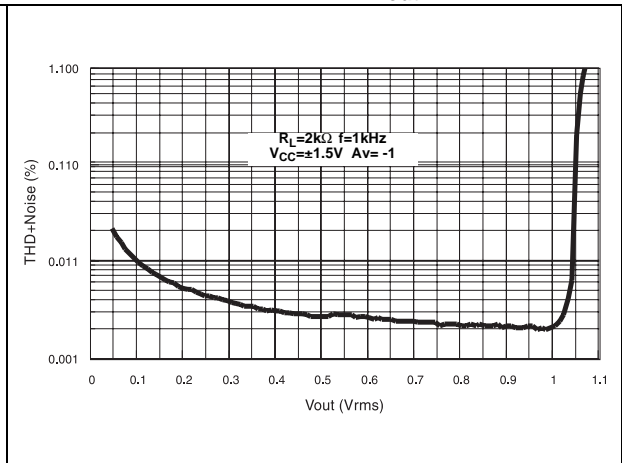


Figure 12. THD + noise vs. V_{out}



3 Macromodel

3.1 Important note concerning this macromodel

Please consider the following remarks before using this macromodel.

- All models are a trade-off between accuracy and complexity (i.e. simulation time).
- Macromodels are not a substitute to breadboarding; rather, they confirm the validity of a design approach and help to select surrounding component values.
- A macromodel emulates the **nominal** performance of a **typical** device within **specified operating conditions** (temperature, supply voltage, for example). Thus the macromodel is often not as exhaustive as the datasheet, its purpose is to illustrate the main parameters of the product.

Data derived from macromodels used outside of the specified conditions (V_{CC} , temperature, for example) or even worse, outside of the device operating conditions (V_{CC} , V_{icm} , for example), is not reliable in any way.

[Section 3.2](#) presents the electrical characteristics resulting from the use of these macromodels.

3.2 Electrical characteristics from macromodelization

Table 5. Macromodel simulation at $V_{CC} = 3V$, $V_{DD} = 0V$, R_L , C_L connected to $V_{CC}/2$, and $T_{amb} = 25^\circ C$ (unless otherwise specified)

Symbol	Conditions	Value	Unit
V_{io}		0	mV
A_{vd}	$R_L = 10k\Omega$	200	V/mV
I_{CC}	No load, per operator	1.2	mA
V_{icm}		-0.2 to 3.2	V
V_{OH}	$R_L = 10k\Omega$	2.95	V
V_{OL}	$R_L = 10k\Omega$	25	mV
I_{sink}	$V_O = 3V$	80	mA
I_{source}	$V_O = 0V$	80	mA
GBP	$R_L = 600k\Omega$	4	MHz
SR	$R_L = 10k\Omega$, $C_L = 100pF$	1	V/ μs
ϕ_m	$R_L = 600k\Omega$	68	Degrees

3.3 Macromodel code

```

** Standard Linear Ics Macromodels, 1996.
** CONNECTIONS:
* 1 INVERTING INPUT
* 2 NON-INVERTING INPUT
* 3 OUTPUT
* 4 POSITIVE POWER SUPPLY
* 5 NEGATIVE POWER SUPPLY

.SUBCKT TS92X 1 2 3 4 5
*
.MODEL MDTH D IS=1E-8 KF=2.664234E-16 CJO=10F
*
* INPUT STAGE
CIP 2 5 1.000000E-12
CIN 1 5 1.000000E-12
EIP 10 5 2 5 1
EIN 16 5 1 5 1
RIP 10 11 8.125000E+00
RIN 15 16 8.125000E+00
RIS 11 15 2.238465E+02
DIP 11 12 MDTH 400E-12
DIN 15 14 MDTH 400E-12
VOFP 12 13 DC 153.5u
VOFN 13 14 DC 0
IPOL 13 5 3.200000E-05
CPS 11 15 1e-9
DINN 17 13 MDTH 400E-12
VIN 17 5 -0.100000e+00
DINR 15 18 MDTH 400E-12
VIP 4 18 0.400000E+00
FCP 4 5 VOFN 1.865000E+02
FCN 5 4 VOFN 1.865000E+02
FIBP 2 5 VOFN 6.250000E-03
FIBN 5 1 VOFN 6.250000E-03
* GM1 STAGE *****
FGM1P 119 5 VOFN 1.1
FGM1N 119 5 VOFN 1.1
RAP 119 4 2.6E+06
RAN 119 5 2.6E+06
* GM2 STAGE *****
G2P 19 5 119 5 1.92E-02
G2N 19 5 119 4 1.92E-02
R2P 19 4 1E+07
R2N 19 5 1E+07
*****
VINT1 500 0 5
GCONVP 500 501 119 4 19.38
VP 501 0 0
GCONVN 500 502 119 5 19.38
VN 502 0 0

```



```
***** orientation isink isource *****
VINT2 503 0 5
FCOPY 503 504 VOUT 1
DCOPYP 504 505 MDTH 400E-9
VCOPYP 505 0 0
DCOPYN 506 504 MDTH 400E-9
VCOPYN 0 506 0
*****
F2PP 19 5 poly(2) VCOPYP VP 0 0 0 0 0.5
F2PN 19 5 poly(2) VCOPYP VN 0 0 0 0 0.5
F2NP 19 5 poly(2) VCOPYN VP 0 0 0 0 1.75
F2NN 19 5 poly(2) VCOPYN VN 0 0 0 0 1.75
* COMPENSATION *****
CC 19 119 25p
* OUTPUT *****
DOPM 19 22 MDTH 400E-12
DONM 21 19 MDTH 400E-12
HOPM 22 28 VOUT 6.250000E+02
VIPM 28 4 5.000000E+01
HONM 21 27 VOUT 6.250000E+02
VINM 5 27 5.000000E+01
VOUT 3 23 0
ROUT 23 19 6
COUT 3 5 1.300000E-10
DOP 19 25 MDTH 400E-12
VOP 4 25 1.052
DON 24 19 MDTH 400E-12
VON 24 5 1.052
.ENDS ;TS92X
```

4 Package information

In order to meet environmental requirements, STMicroelectronics offers these devices in ECOPACK® packages. These packages have a lead-free second level interconnect. The category of second level interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label. ECOPACK is an STMicroelectronics trademark. ECOPACK specifications are available at: www.st.com.

4.1 DIP14 package information

Figure 13. DIP14 package mechanical drawing

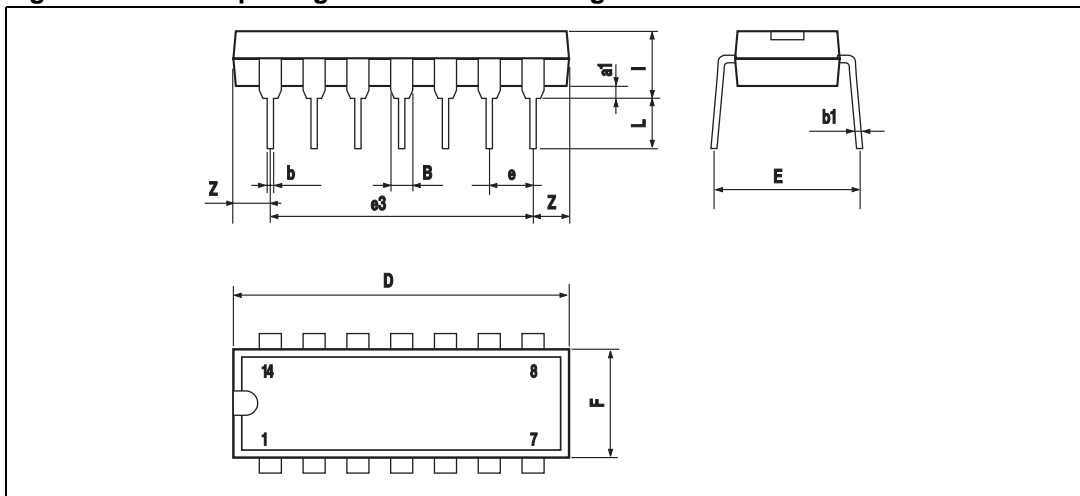


Table 6. DIP14 package mechanical data

Ref.	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
a1	0.51			0.020		
B	1.39		1.65	0.055		0.065
b		0.5			0.020	
b1		0.25			0.010	
D			20			0.787
E		8.5			0.335	
e		2.54			0.100	
e3		15.24			0.600	
F			7.1			0.280
I			5.1			0.201
L		3.3			0.130	
Z	1.27		2.54	0.050		0.100

4.2 SO-14 package information

Figure 14. SO-14 package mechanical drawing

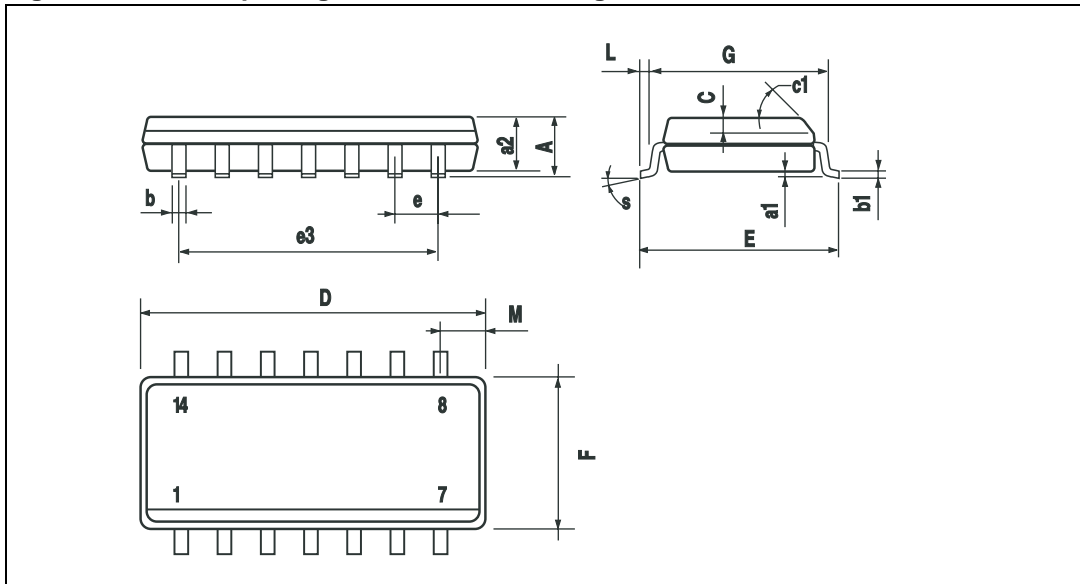


Table 7. SO-14 package mechanical data

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.75			0.068
a1	0.1		0.2	0.003		0.007
a2			1.65			0.064
b	0.35		0.46	0.013		0.018
b1	0.19		0.25	0.007		0.010
C		0.5			0.019	
c1	45° (typ.)					
D	8.55		8.75	0.336		0.344
E	5.8		6.2	0.228		0.244
e		1.27			0.050	
e3		7.62			0.300	
F	3.8		4.0	0.149		0.157
G	4.6		5.3	0.181		0.208
L	0.5		1.27	0.019		0.050
M			0.68			0.026
S	8° (max.)					

4.3 TSSOP14 package information

Figure 15. TSSOP14 package mechanical drawing

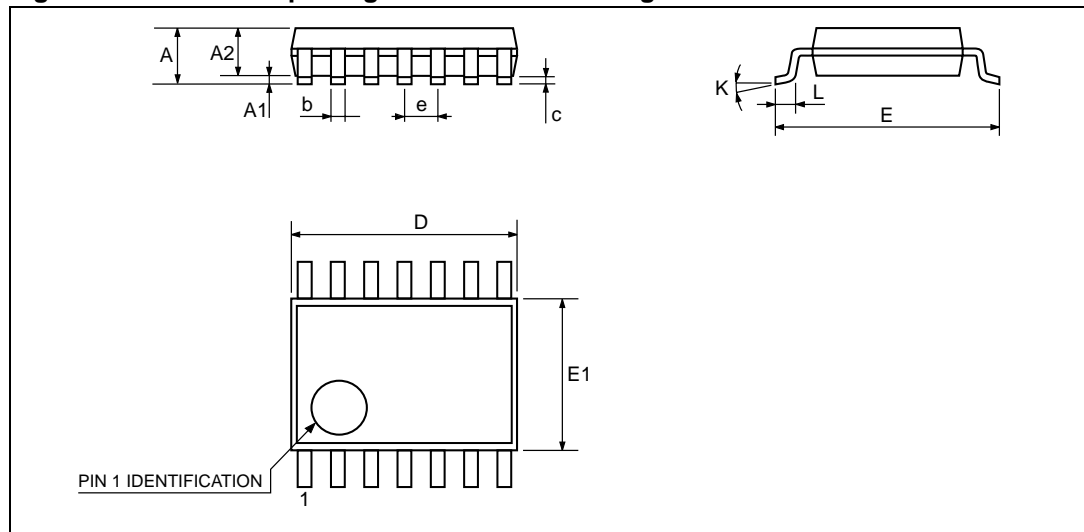


Figure 16. TSSOP14 package mechanical data

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.2			0.047
A1	0.05		0.15	0.002	0.004	0.006
A2	0.8	1	1.05	0.031	0.039	0.041
b	0.19		0.30	0.007		0.012
c	0.09		0.20	0.004		0.0089
D	4.9	5	5.1	0.193	0.197	0.201
E	6.2	6.4	6.6	0.244	0.252	0.260
E1	4.3	4.4	4.48	0.169	0.173	0.176
e		0.65 BSC			0.0256 BSC	
K	0°		8°	0°		8°
L1	0.45	0.60	0.75	0.018	0.024	0.030

5 Ordering information

Table 8. Order codes

Order code	Temperature range	Package	Packaging	Marking
TS924IN	-40°C, +125°C	DIP14	Tube	TS924IN
TS924AIN				TS924AIN
TS924ID TS924IDT		SO-14	Tube or Tape and reel	924I
TS924AID TS924AIDT				924AI
TS924IYD ⁽¹⁾ TS924IYDT ⁽¹⁾		SO-14 (Automotive grade)	Tube or Tape and reel	924IY
TS924AIYD ⁽¹⁾ TS924AIYDT ⁽¹⁾				924AIY
TS924IPT ⁽²⁾		TSSOP14	Tape and reel	924I
TS924AIPT ⁽²⁾				924AI
TS924IYPT ⁽²⁾		TSSOP14 (Automotive grade)	Tape and reel	924IY
TS924AIYPT ⁽²⁾				924AIY

1. Qualified and characterized according to AEC Q100 and Q003 or equivalent, advanced screening according to AEC Q001 & Q 002 or equivalent.
2. Qualification and characterization according to AEC Q100 and Q003 or equivalent, advanced screening according to AEC Q001 & Q 002 or equivalent are on-going.

6 Revision history

Table 9. Document revision history

Date	Revision	Changes
28-May-2001	1	First release.
12-May-2005	2	Modifications on AMR Table on page 2 (explanation of V_{id} and V_{in} limits, ESD MM and CDM values added, R_{thja} added).
31-Jul-2005	3	PPAP references inserted in the datasheet see Table 1 on page 3 .
30-Nov-2005	4	Package mechanical data modified. TS924IYPT/TS924AIYPT PPAP reference inserted in order code table. Macromodel modified.
11-Mar-2008	5	Added footnotes for automotive grade order codes in order code table. Updated document format.

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