Advanced Monolithic Systems

AMS2954

250mA LOW DROPOUT VOLTAGE REGULATOR

FEATURES

- **2.5V, 3.0V, 3.3V and 5.0V Versions**
- **High Accuracy Output Voltage**
- **Extremely Low Quiescent Current**
- **Low Dropout Voltage**
- **Extremely Tight Load and Line Regulation**
- **Very Low Temperature Coefficient**
- **Current and Thermal Limiting**
- **Needs Minimum Capacitance (1µF) for Stability**
- **Unregulated DC Positive Transients 60V**

ADDITIONAL FEATURES (ADJ ONLY)

- **1.24V to 29V Programmable Output**
- **Error Flag Warning of Voltage Output Dropout**
- **Logic Controlled Electronic Shutdown**

APPLICATIONS

- **Battery Powered Systems**
- **Portable Consumer Equipment**
- **Cordless Telephones**
- **Portable (Notebook) Computers**
- **Portable Instrumentation**
- **Radio Control Systems**
- **Automotive Electronics**
- **Avionics**
- **Low-Power Voltage Reference**

GENERAL DESCRIPTION

The AMS2954 series are micropower voltage regulators ideally suited for use in battery-powered systems. These devices feature very low quiescent current (typ.75µA), and very low dropout voltage (typ.50mV at light loads and 380mV at 250mA) thus prolonging battery life. The quiescent current increases only slightly in dropout. The AMS2954 has positive transient protection up to 60V and can survive unregulated input transient up to 20V below ground. The AMS2954 was designed to include a tight initial tolerance (typ. 0.5%), excellent load and line regulation (typ. 0.05%), and a very low output voltage temperature coefficient, making these devices useful as a low-power voltage reference.

The AMS2954 is available in the 3L TO-220 package, 3L TO-263, SOT-223, TO-252 and in 8-pin plastic SOIC and DIP packages. In the 8L SOIC and PDIP packages the following additional features are offered: an error flag output warns of a low output voltage, often due to failing batteries on input; the logic-compatible shutdown input enables the regulator to be switched on and off; the device may be pin-strapped for a, 2.5, 3.0V, 3.3V or 5V output, or programmed from 1.24V to 29V with an external pair of resistors.

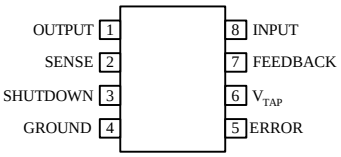
ORDERING INFORMATION

PACKAGE TYPE						OPERATING TEMP. RANGE
3 LEAD TO-220	3 LEAD TO-263	TO-252	SOT-223	8 LEAD SOIC	8 LEAD PDIP	
AMS2954ACT-X	AMS2954ACM-X	AMS2954ACD-X	AMS2954AC-X	AMS2954ACS-X	AMS2954CP-X	IND.
AMS2954CT-X	AMS2954CM-X	AMS2954CD-X	AMS2954C-X	AMS2954CS-X	AMS2954CP-X	IND

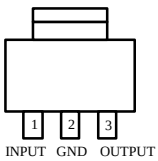
X = 2.5V, 3.0V, 3.3V, 5.0V

PIN CONNECTIONS

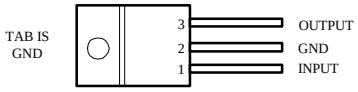
8L SOIC/ 8L PDIP



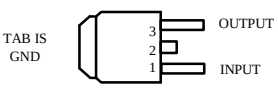
SOT-223 TOP VIEW



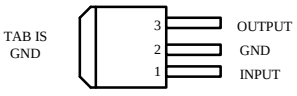
3L TO-220 FRONT VIEW



TO-252 FRONT VIEW



3L TO-263 FRONT VIEW



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ABSOLUTE MAXIMUM RATINGS (Note 1)

Input Supply Voltage -0.3 to +30V
SHUTDOWN Input Voltage, Error Comparator Output Voltage,(Note 9)
FEEDBACK Input Voltage (Note 9) (Note 10) -1.5 to +30V
Power Dissipation Internally Limited
Junction Temperature +150°C
Storage Temperature -65°C to +150°C
ESD 2kV

Soldering Dwell Time, Temperature
Wave 4 seconds, 260°C
Infrared 4 seconds, 240°C
Vapor Phase 4 seconds, 219°C

OPERATING RATINGS (Note 1)

Max. Input Supply Voltage 40V
Junction Temperature Range (T_J) (Note 8)
AMS2954AC-X -40°C to +125°C
AMS2954C-X

ELECTRICAL CHARACTERISTICS at V_S=Vout+1V, Ta=25°C, unless otherwise noted.

Parameter	Conditions (Note 2)	AMS2954AC			AMS2954C			Units
		Min.	Typ.	Max.	Min.	Typ.	Max.	
2.5 V Versions (Note 16)								
Output Voltage	T_J = 25°C (Note 3)	2.488	2.5	2.512	2.475	2.5	2.525	V
	-25°C ≤T_J ≤85°C	2.475	2.5	2.525	2.450	2.5	2.550	V
	Full Operating Temperature Range	**2.470**	2.5	**2.530**	**2.440**	2.5	**2.560**	V
Output Voltage	100 µA ≤I_L ≤250 mA T_J ≤T_{JMAX}	**2.463**	2.5	**2.537**	**2.448**	2.5	**2.562**	V
3.0 V Versions (Note 16)								
Output Voltage	T_J = 25°C (Note 3)	2.985	3.0	3.015	2.970	3.0	3.030	V
	-25°C ≤T_J ≤85°C	2.970	3.0	3.030	2.955	3.0	3.045	V
	Full Operating Temperature Range	**2.964**	3.0	**3.036**	**2.940**	3.0	**3.060**	V
Output Voltage	100 µA ≤I_L ≤250 mA T_J ≤T_{JMAX}	**2.958**	3.0	**3.042**	**2.928**	3.0	**3.072**	V
3.3 V Versions (Note 16)								
Output Voltage	T_J = 25°C (Note 3)	3.284	3.3	3.317	3.267	3.3	3.333	V
	-25°C ≤T_J ≤85°C	3.267	3.3	3.333	3.251	3.3	3.350	V
	Full Operating Temperature Range	**3.260**	3.3	**3.340**	**3.234**	3.3	**3.366**	V
Output Voltage	100 µA ≤I_L ≤250 mA T_J ≤T_{JMAX}	**3.254**	3.3	**3.346**	**3.221**	3.3	**3.379**	V
5 V Versions (Note 16)								
Output Voltage	T_J = 25°C (Note 3)	4.975	5.0	5.025	4.95	5.0	5.05	V
	-25°C ≤T_J ≤85°C	4.95	5.0	5.050	4.925	5.0	5.075	V
	Full Operating Temperature Range	**4.94**	5.0	**5.06**	**4.90**	5.0	**5.10**	V
Output Voltage	100 µA ≤I_L ≤250 mA T_J ≤T_{JMAX}	**4.925**	5.0	**5.075**	**4.88**	5.0	**5.12**	V
All Voltage Options								
Output Voltage Temperature Coefficient	(Note 12) (Note 4)		**20**	**100**		**50**	**150**	ppm/°C
Line Regulation (Note 14)	6V ≤V_{in} ≤30V (Note 15)		0.03	0.1		0.04	0.2	%
Load Regulation (Note 14)	100 µA ≤I_L ≤ 250 mA		0.04	0.16		0.1	0.2	%

ELECTRICAL CHARACTERISTICS (Note 2) (Continued)

PARAMETER	CONDITIONS (Note 2)	AMS2954AC			AMS2954C			Units
		Min.	Typ.	Max.	Min.	Typ.	Max.	
Dropout Voltage (Note 5)	I_L = 100µ A I_L = 250 mA		50 380	80 600		50 380	80 600	mV mV
Ground Current	I_L = 100 µA I_L = 250 mA		75 15	120 20		75 15	120 20	µA mA
Current Limit	V_{out} = 0		200	500		200	500	mA
Thermal Regulation	(Note 13)		0.05	0.2		0.05	0.2	%/W
Output Noise, 10Hz to 100KHz	C_L = 1µF C_L = 200 µF C_L = 13.3 µF (Bypass = 0.01 µF pins 7 to 1)		430 160 100			430 160 100		µV rms µV rms µV rms
8-Pin Versions only		**AMS2954AC**			**AMS2954C**			
Reference Voltage		1.22	1.235	1.25	1.21	1.235	1.26	V
Reference Voltage	Over Temperature (Note 7)	**1.19**		**1.27**	**1.185**		**1.285**	V
Feedback Pin Bias Current			40	60		40	60	nA
Reference Voltage Temperature Coefficient	(Note 12)		20				50	ppm/°C
Feedback Pin Bias Current Temperature Coefficient			0.1				0.1	nA/°C
Error Comparator								
Output Leakage Current	V_{OH} = 30V		0.01	1		0.01	1	µA
Output Low Voltage	V_{in} = 4.5V I_{OL} = 400µA		150	250		150	250	mV
Upper Threshold Voltage	(Note 6)	40	60		40	60		mV
Lower Threshold Voltage	(Note 6)		75	95		75	95	mV
Hysteresis	(Note 6)		15			15		mV
Shutdown Input								
Input logic Voltage	Low (Regulator ON) High (Regulator OFF)	 **2**	1.3	**0.7**	 **2**	1.3	**0.7**	V V
Shutdown Pin Input Current (Note 3)	Vs = 2.4V V_s = 30V		30 450	50 600		30 450	50 600	µA µA
Regulator Output Current in Shutdown (Note 3)	(Note 11)		3	10		3	10	µA

Note 1: Absolute Maximum Ratings are limits beyond which damage to the device may occur. Operating Ratings are conditions under which operation of the device is guaranteed. Operating Ratings do not imply guaranteed performance limits. For guaranteed performance limits and associated test conditions, see the Electrical Characteristics tables.

Note 2: Unless otherwise specified all limits guaranteed for V_{IN} = (V_{ONOM} +1)V, I_L = 100 µA and C_L = 1 µF for 5V versions and 2.2µF for 3V and 3.3V versions. Limits appearing in **boldface** type apply over the entire junction temperature range for operation. Limits appearing in normal type apply for $T_A = T_J$ = 25°C Additional conditions for the 8-pin versions are FEEDBACK tied to V_{TAP}, OUTPUT tied to SENSE and $V_{SHUTDOWN} \leq 0.8V$.

Note 3: Guaranteed and 100% production tested.

Note 4: Guaranteed but not 100% production tested. These limits are not used to calculate outgoing AQL levels.

Note 5: Dropout voltage is defined as the input to output differential at which the output voltage drops 100 mV below its nominal value measured at 1V differential. At very low values of programmed output voltage, the minimum input supply voltage of 2V (2.3V over temperature) must be taken into account.

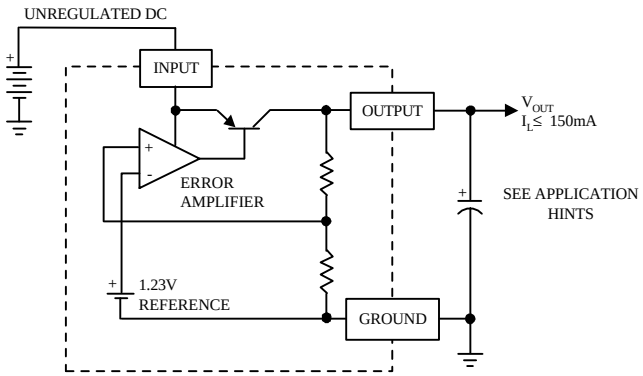
Note 6: Comparator thresholds are expressed in terms of a voltage differential at the feedback terminal below the nominal reference voltage measured at V_{IN} = (V_{ONOM} +1)V. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain = Vout/Vref = (R1 + R2)/R2. For example, at a programmed output voltage of 5V, the error output is guaranteed to go low when the output drops by 95 mV x 5V/1.235 = 384 mV. Thresholds remain constant as a percent of V_{out} as V_{out} is varied, with the dropout warning occurring at typically 5% below nominal, 7.5% guaranteed.

Note 7: $V_{ref} \leq V_{out} \leq (V_{in} - 1V)$, $2.3 \leq V_{in} \leq 30V$, $100\mu A \leq I_L \leq 250$ mA, $T_J \leq T_{JMAX}$.

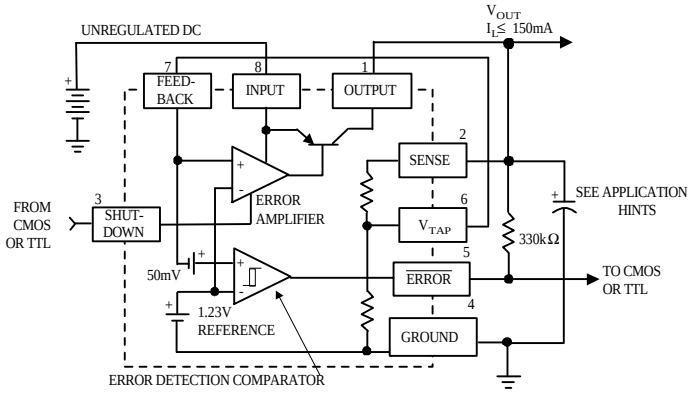
Note 8: The junction-to-ambient thermal resistance are as follows:60°C/W for the TO-220 (T), 73°C/W for the TO-263 (M), 80°C/W for the TO-252 (D), 90°C/W for the SOT-223 (with package soldering to copper area over backside ground plane or internal power plane φ_{JA} can vary from 46°C/W to >90°C/W depending on mounting technique and the size of the copper area), 105°C/W for the molded plastic DIP (P) and 160°C/W for the molded plastic SO-8 (S).
Note 9: May exceed input supply voltage.
Note 10: When used in dual-supply systems where the output terminal sees loads returned to a negative supply, the output voltage should be diode-clamped to ground.
Note 11: $V_{shutdown} \geq 2V$, $V_{in} \leq 30V$, $V_{out} = 0$, Feedback pin tied to $5V_{TAP}$.
Note 12: Output or reference voltage temperature coefficients defined as the worst case voltage change divided by the total temperature range.
Note 13: Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 50mA load pulse at $V_{IN} = 30V$ (1.25W pulse) for T =10 ms.
Note 14: Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specification for thermal regulation.
Note 15: Line regulation for the AMS2954 is tested at 150°C for I_L = 1 mA. For I_L = 100 µA and T_J = 125°C, line regulation is guaranteed by design to 0.2%. See typical performance characteristics for line regulation versus temperature and load current.

BLOCK DIAGRAM AND TYPICAL APPLICATIONS

AMS2954-XX
3 Lead Packages

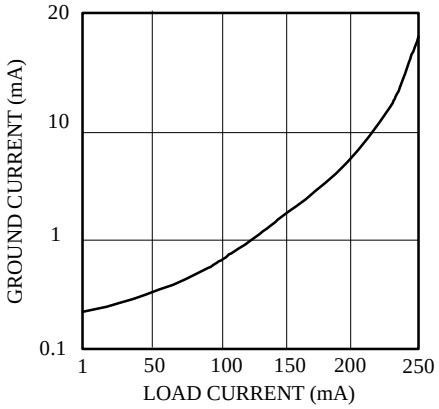


AMS2954-XX
8 Lead Packages

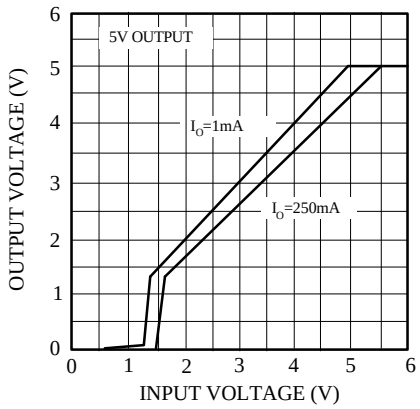


TYPICAL PERFORMANCE CHARACTERISTICS

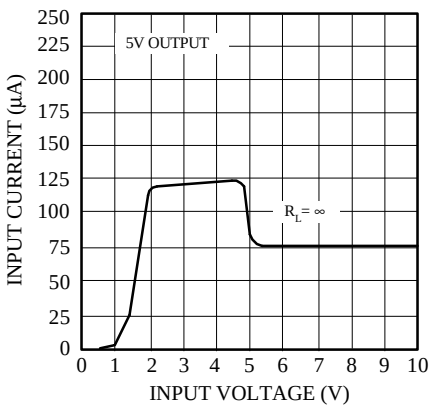
Quiescent Current



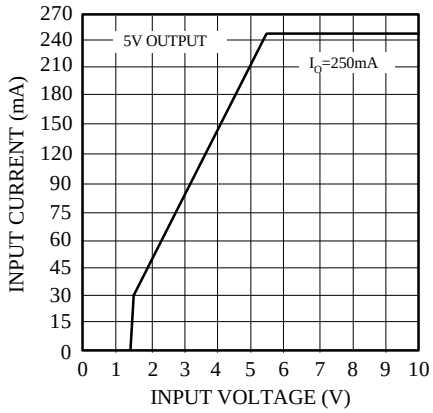
Dropout Characteristics



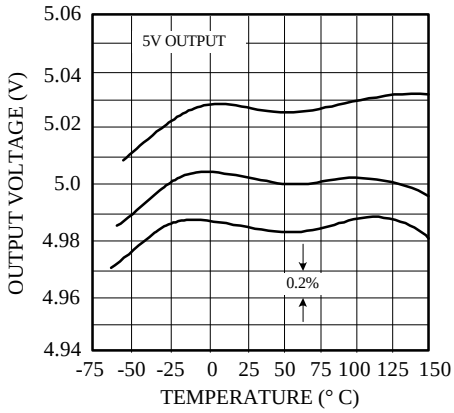
Input Current



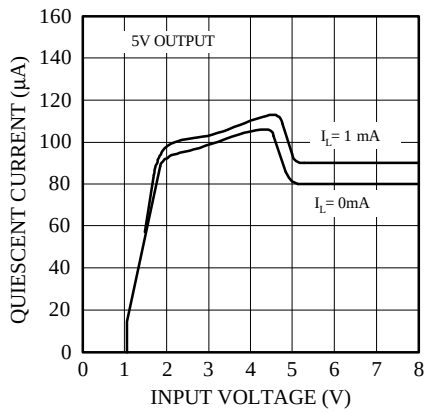
Input Current



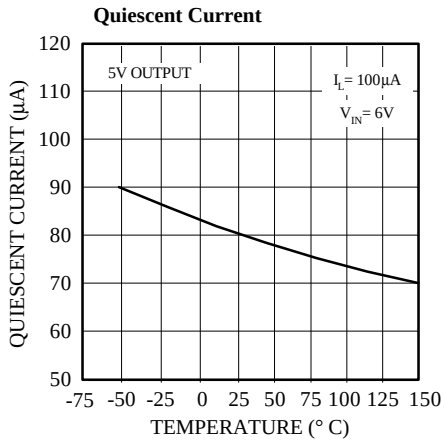
Temperature Drift of 3 Representative Units



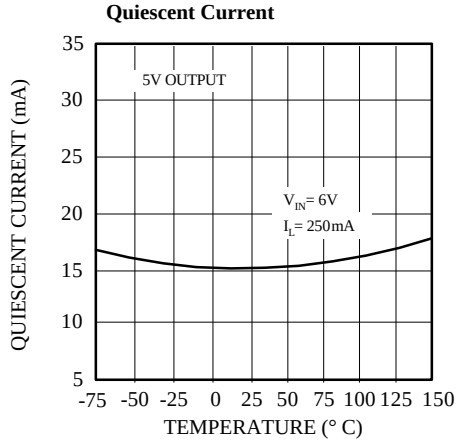
Quiescent Current



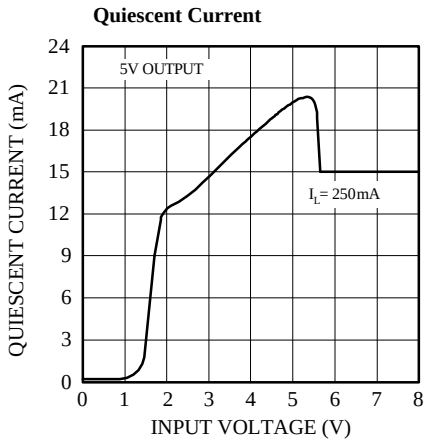
Quiescent Current



Quiescent Current

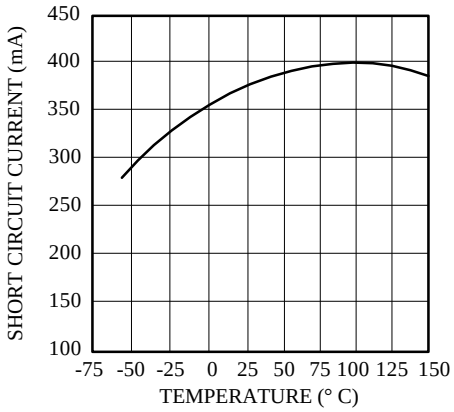


Quiescent Current

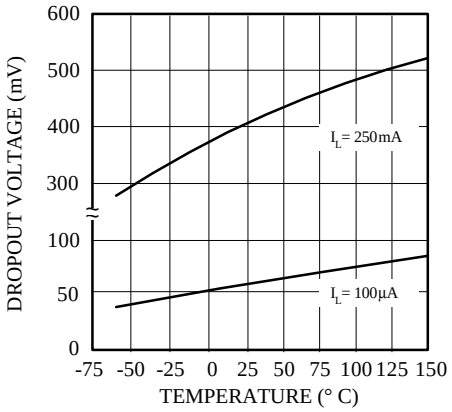


TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

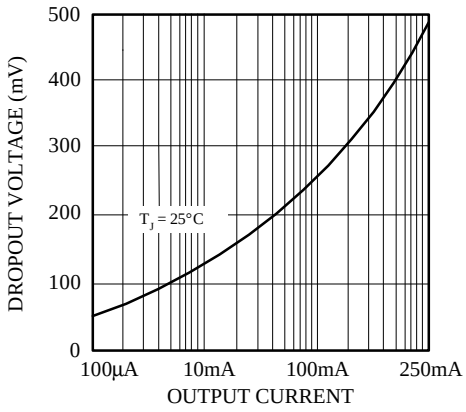
Short Circuit Current



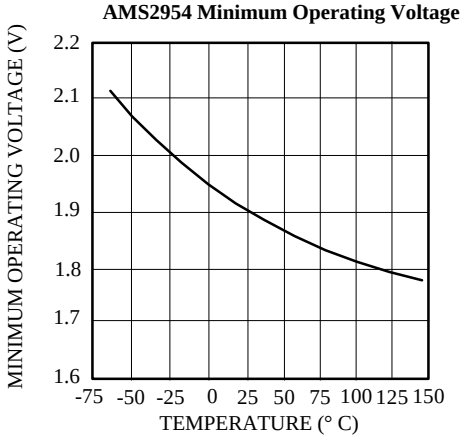
Dropout Voltage



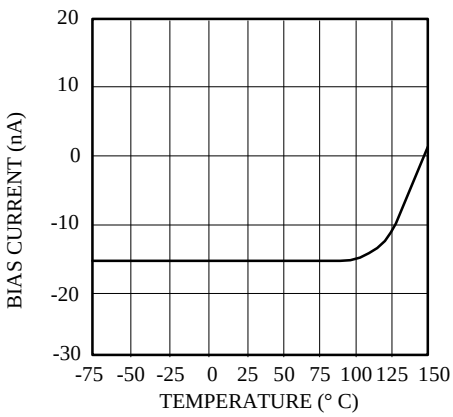
Dropout Voltage



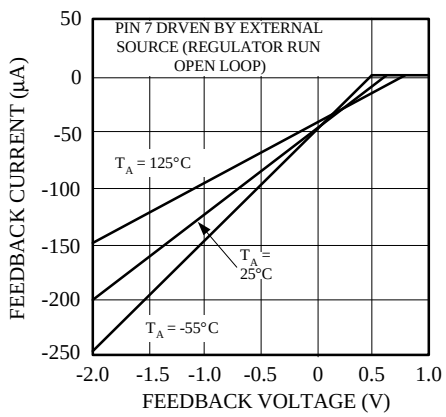
AMS2954 Minimum Operating Voltage



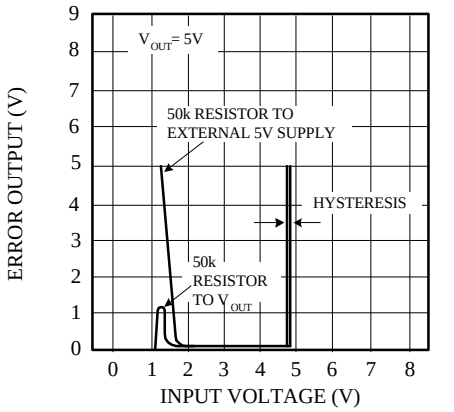
AMS2954 Feedback Bias Current



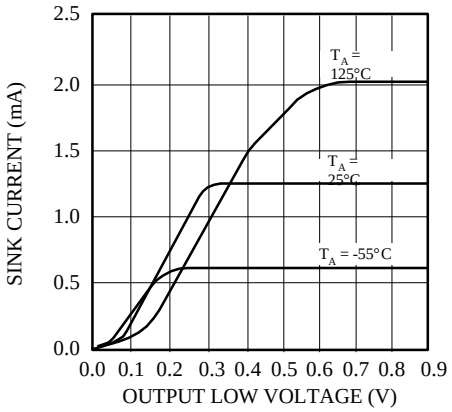
AMS2954 Feedback Pin Current



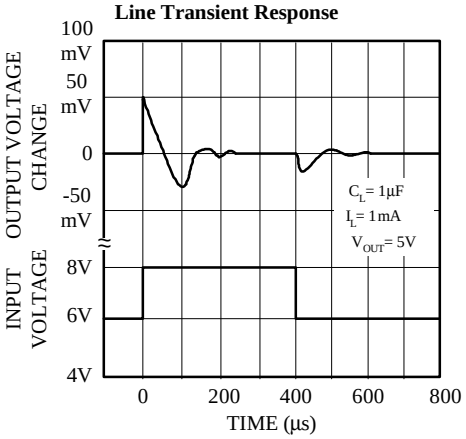
AMS2954 Error Comparator Output



AMS2954 Comparator Sink Current

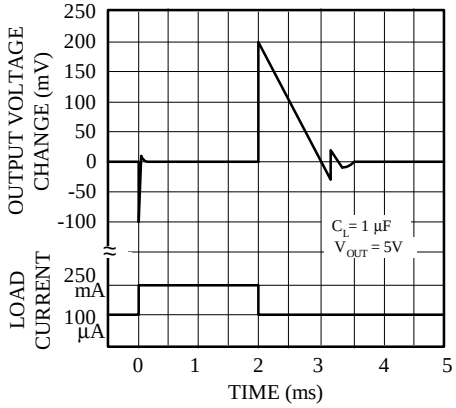


Line Transient Response

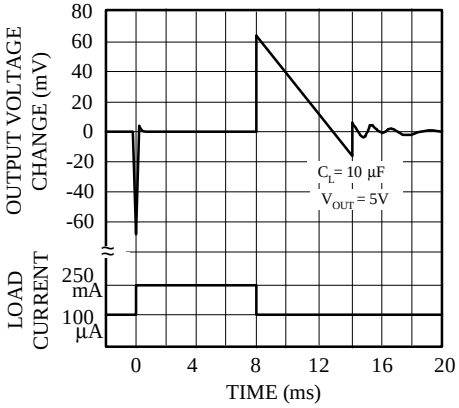


TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

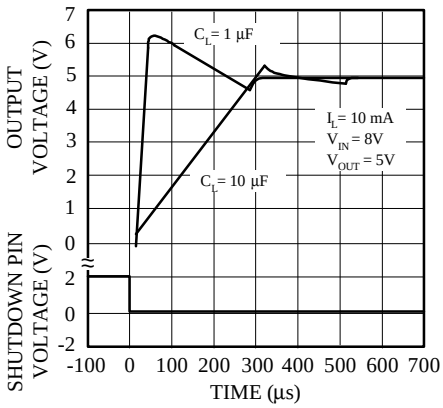
Load Transient Response



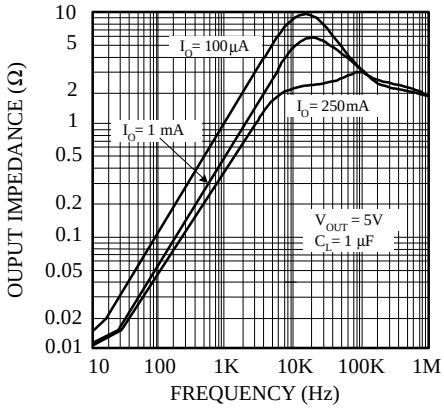
Load Transient Response



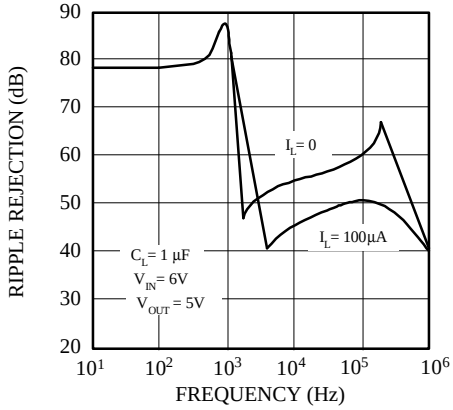
AMS2954 Enable Transient



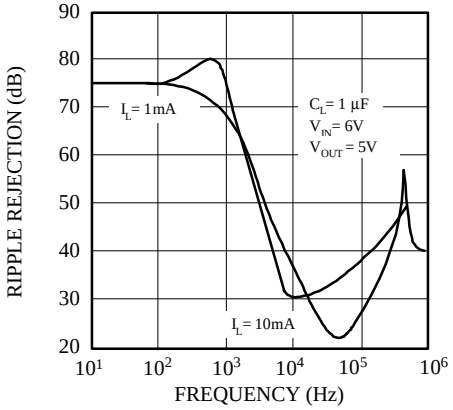
Output Impedance



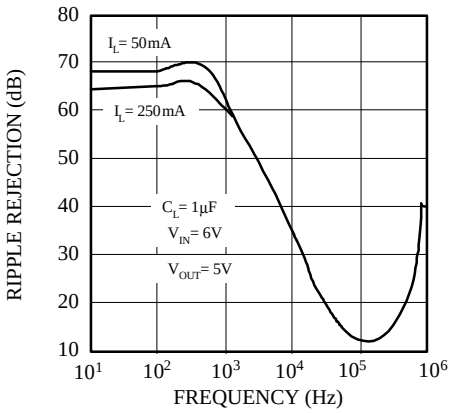
Ripple Rejection



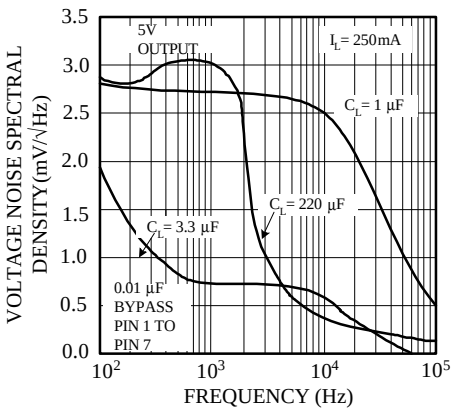
Ripple Rejection



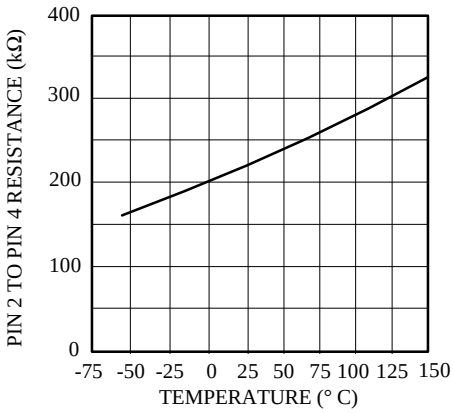
Ripple Rejection



AMS2954 Output Noise

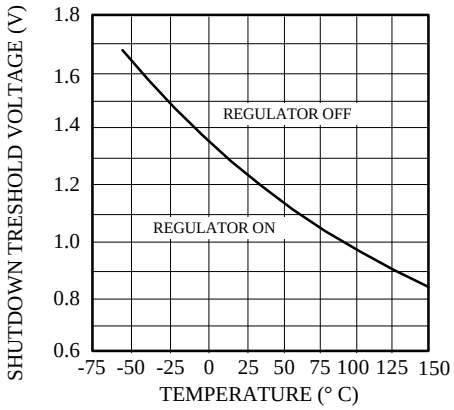


AMS2954 Divider Resistance

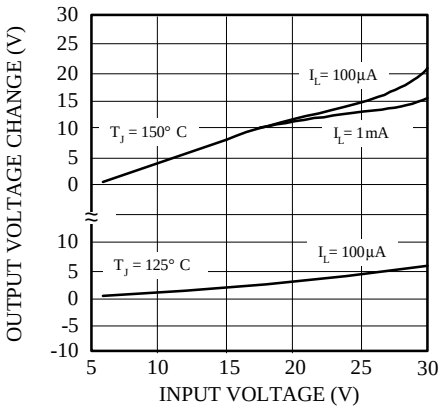


TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

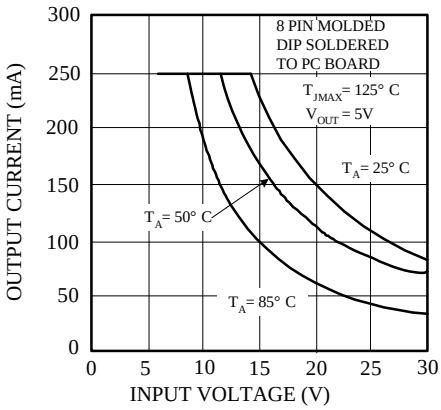
Shutdown Treshold Voltage



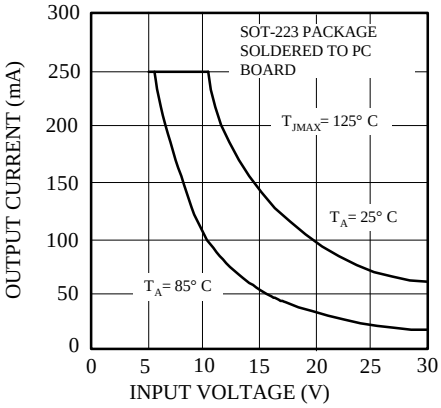
Line Regulation



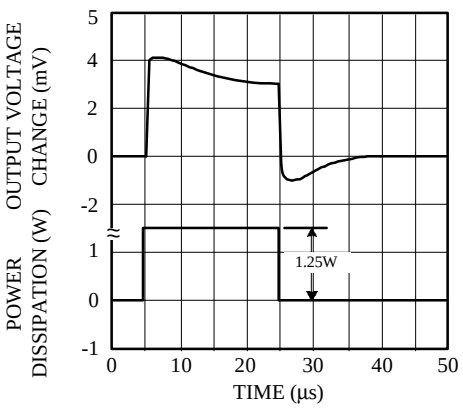
AMS2954 Maximum Rated Output Current



AMS2954 Maximum Rated Output Current



Thermal Response



APPLICATION HINTS

External Capacitors

A 1.0 µF or greater capacitor is required between output and ground for stability at output voltages of 5V or more. At lower output voltages, more capacitance is required (2.2µ or more is recommended for 2.5V, 3.0V and 3.3V versions). Without this capacitor the part will oscillate. Most types of tantalum or aluminum electrolytic works fine here; even film types work but are not recommended for reasons of cost. Many aluminum types have electrolytes that freeze at about -30°C, so solid tantalums are recommended for operation below -25°C. The important parameters of the capacitor are an ESR of about 5 Ω or less and resonant frequency above 500 kHz parameters in the value of the capacitor. The value of this capacitor may be increased without limit.

At lower values of output current, less output capacitance is required for stability. The capacitor can be reduced to 0.33 µF for currents below 10 mA or 0.1 µF for currents below 1 mA. Using the adjustable versions at voltages below 5V runs the error amplifier at lower gains so that more output capacitance is needed. For the worst-case situation of a 300mA load at 1.23V output (Output shorted to Feedback) a 3.3µF (or greater) capacitor should be used.

Unlike many other regulators, the AMS2954, will remain stable and in regulation with no load in addition to the internal voltage divider. This is especially important in CMOS RAM keep-alive applications. When setting the output voltage of the AMS2954 version with external resistors, a minimum load of 1µA is recommended.

A 1µF tantalum or aluminum electrolytic capacitor should be placed from the AMS2954/AMS2954 input to the ground if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

Stray capacitance to the AMS2954 Feedback terminal can cause instability. This may especially be a problem when using a higher value of external resistors to set the output voltage. Adding a 100 pF capacitor between Output and Feedback and increasing the output capacitor to at least 3.3 µF will fix this problem.

Error Detection Comparator Output

The comparator produces a logic low output whenever the AMS2954 output falls out of regulation by more than approximately 5%. This figure is the comparator's built-in offset of about 60 mV divided by the 1.235 reference voltage (Refer to the block diagram). This trip level remains "5% below normal" regardless of the programmed output voltage of the 2951. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V output. The out of regulation condition may be due either to low input voltage, current limiting, or thermal limiting.

Figure 1 gives a timing diagram depicting the ERROR signal and the regulator output voltage as the AMS2954 input is ramped up and down. For 5V versions the ERROR signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at which Vout = 4.75).

Since the AMS2954's dropout voltage is load dependent (see curve in typical performance characteristics), the input voltage trip point (about 5V) will vary with the load current. The output voltage trip point (approx. 4.75V) does not vary with load.

The error comparator has an open-collector output which requires an external pull-up resistor. This resistor may be returned to the output or some other supply voltage depending on system requirements. In determining a value for this resistor, note that the output is rated to sink 400µA, this sink current adds to battery drain in a low battery condition. Suggested values range from 100K to 1MΩ. The resistor is not required if this output is unused.

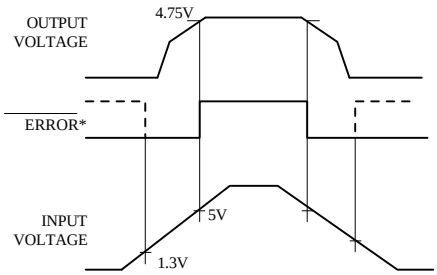


FIGURE 1. ERROR Output Timing

*When V_{IN} 1.3V the error flag pin becomes a high impedance, and the error flag voltage rises to its pull-up voltage. Using V_{out} as the pull-up voltage (see Figure 2), rather than an external 5V source, will keep the error flag voltage under 1.2V (typ.) in this condition. The user may wish to drive down the error flag voltage using equal value resistors (10 k suggested), to ensure a low-level logic signal during any fault condition, while still allowing a valid high logic level during normal operation.

Programming the Output Voltage

The AMS2954 may be pin-strapped for the nominal fixed output voltage using its internal voltage divider by tying the output and sense pins together, and also tying the feedback and V_{TAP} pins together. Alternatively, it may be programmed for any output voltage between its 1.235V reference and its 30V maximum rating. As seen in Figure 2, an external pair of resistors is required.

The complete equation for the output voltage is:

$$V_{out} = V_{REF} \times (1 + R_1/R_2) + I_{FB}R_1$$

where V_{REF} is the nominal 1.235 reference voltage and I_{FB} is the feedback pin bias current, nominally -20 nA. The minimum recommended load current of 1 µA forces an upper limit of 1.2 MΩ on value of R_2, if the regulator must work with no load (a condition often found in CMOS in standby) I_{FB} will produce a 2% typical error in V_{OUT} which may be eliminated at room temperature by trimming R_1. For better accuracy, choosing R2 = 100k reduces this error to 0.17% while increasing the resistor program current by 12 µA. Since the AMS2954 typically draws 60 µA at no load with Pin 2 open-circuited, this is a small price to pay.

APPLICATION HINTS (Continued)

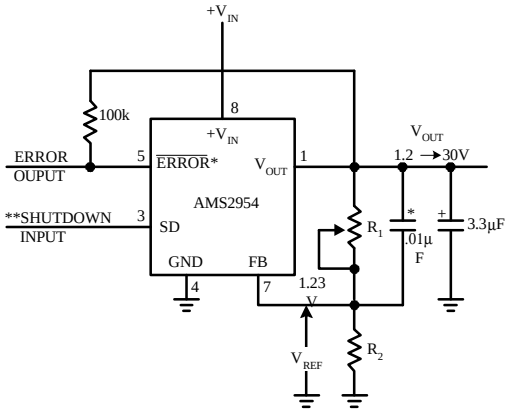


FIGURE 2. Adjustable Regulator

*See Application Hints.

$V_{out} = V_{REF} \times (1 + R_1/ R_2)$

**Drive with TTL- high to shut down. Ground or leave if shutdown feature is not used.

Note: Pins 2 and 6 are left open.

Reducing Output Noise

In reference applications it may be an advantageous to reduce the AC noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This is the only way that noise can be reduced on the 3 lead AMS2954 but is relatively inefficient, as increasing the capacitor from 1 µF to 220 µF only decreases the noise from 430 µV to 160 µV rms for a 100 kHz bandwidth at 5V output.

Noise could also be reduced fourfold by a bypass capacitor across R_1, since it reduces the high frequency gain from 4 to unity. Pick

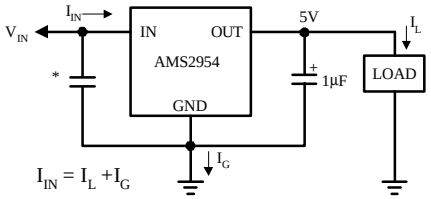
$$C_{BYPASS} \cong 1 / 2\pi R_1 \times 200 \text{ Hz}$$

or about 0.01 µF. When doing this, the output capacitor must be increased to 3.3 µF to maintain stability. These changes reduce the output noise from 430 µV to 100 µV rms for a 100 kHz bandwidth at 5V output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.

Heatsink Requirements

A heatsink might be required when using AMS2954, depending on the maximum power dissipation and maximum ambient temperature of the application. The heatsink must be chosen considering that under all operating condition, the junction temperature must be within the range specified under Absolute Maximum Ratings.

To determine if a heatsink is required, the maximum power dissipated by the regulator must be calculated. It is important to consider, that if the regulator is powered from a transformer connected to the AC line, the maximum specified AC input voltage must be used.



* See external capacitors

$P_{Total} = (V_{IN} -5)I_L + (V_{IN})I_G$

FIGURE 3. Basic 5V Regulator

Figure 3 shows the voltages and currents which are present in a 5V regulator circuit. The formula for calculating the power dissipated in the regulator is also shown in Figure 3.

The next parameter which must be calculated is the maximum allowable temperature rise, $T_{R(max)}$. This is calculated using the formula:

$$T_{R(max)} = T_{J(max)} - T_{A(max)}$$

Where $T_{J(max)}$ is the maximum allowable junction temperature, and $T_{A(max)}$ is the maximum ambient temperature.

Using the calculated values for $T_{R(max)}$ and $P_{(max)}$, the required value for junction to ambient thermal resistance $\theta_{(J-A)}$, can be determined:

$$\theta_{(J-A)} = T_{R(max)} / P_{(max)}$$

If the value obtained is 60°C/W or higher, the regulator may be operated without an external heatsink. If the calculated value is below 60°C/W, an external heatsink is required. To calculate the thermal resistance of this heatsink use the formula:

$$\theta_{(H-A)} = \theta_{(J-A)} - \theta_{(J-C)} - \theta_{(C-H)}$$

where:

$\theta_{(J-C)}$ is the junction-to-case thermal resistance, which is specified as 3°C/W maximum for the AMS2954.

$\theta_{(C-H)}$ is the case-to-heatsink thermal resistance, which is dependent on the interfacing material (if used).

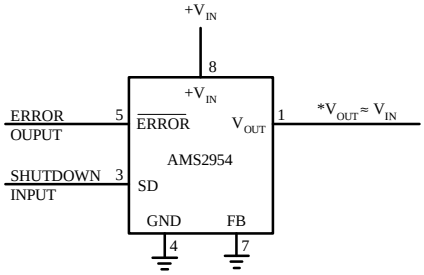
$\theta_{(H-A)}$ is the heatsink-to-ambient thermal resistance. It is this specification which defines the effectiveness of the heatsink. The heatsink selected must have a thermal resistance equal or lower than the value of $\theta_{(H-A)}$ calculated from the above listed formula.

Output Isolation

The regulator output can be left connected to an active voltage source with the regulator input power turned off, as long as the regulator ground pin is connected to ground. If the ground pin is left floating, damage to the regulator can occur if the output is pulled up by an external voltage source.

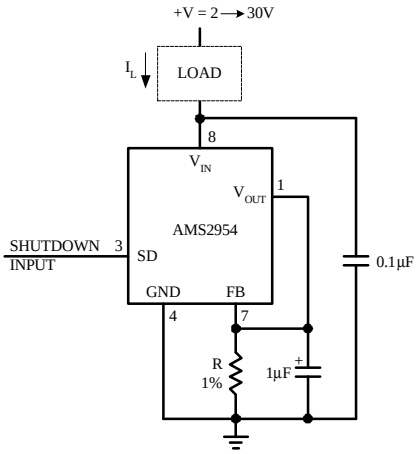
TYPICAL APPLICATIONS (Continued)

Wide Input Voltage Range Current Limiter

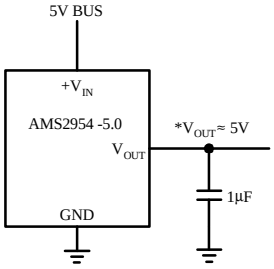


*Minimum Input-Output voltage ranges from 40mV to 400mV, depending on load current. Current limit is typically 260 mA

Low Drift Current Source

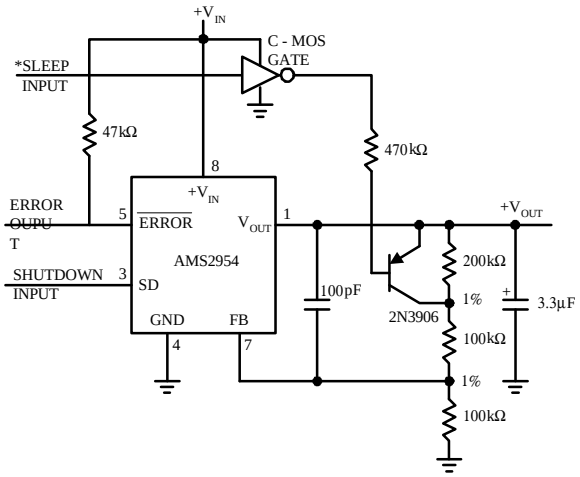


5Volt Current Limiter

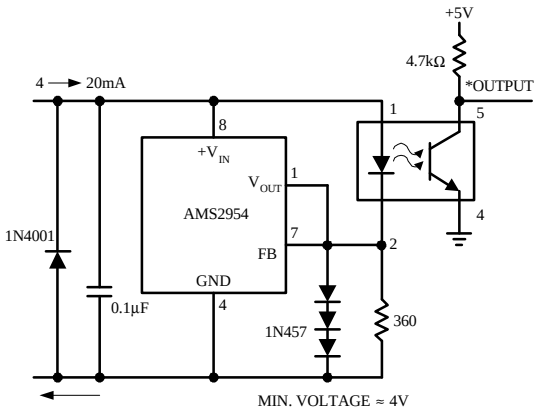


*Minimum Input-Output voltage ranges from 40mV to 400mV, depending on load current. Current limit is typically 260 mA

5V Regulator with 2.5V Sleep Function

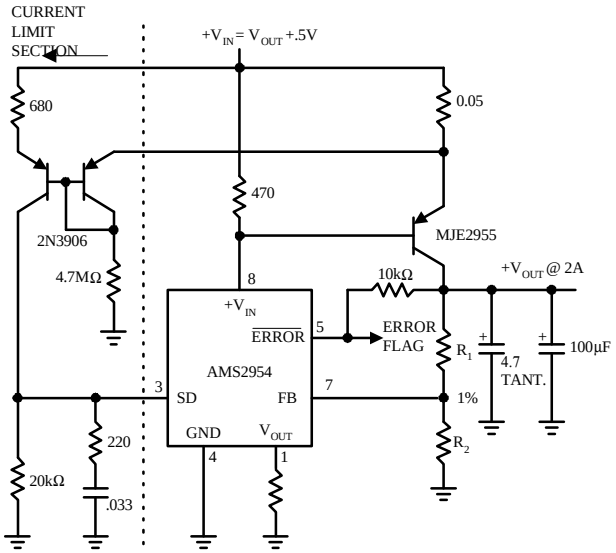


Open Circuit Detector for 4 to 20mA Current Loop



TYPICAL APPLICATIONS (Continued)

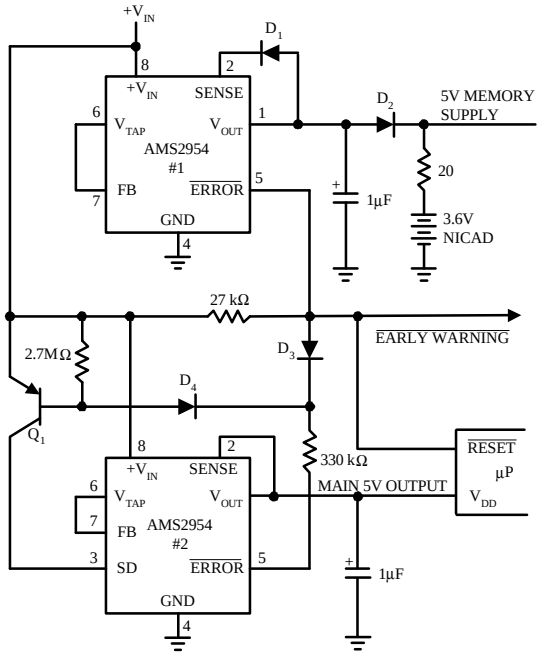
2 Ampere Low Dropout Regulator



VOUT = 1.23V(1+R1/R2)

For 5V V_{OUT}, use internal resistors. Wire pin 6 to 7 and pin 2 to $+V_{OUT}$Buss.

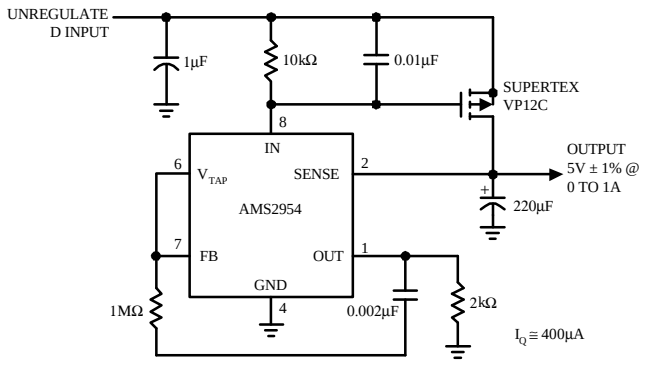
Regulator with Early Warning and Auxiliary Output



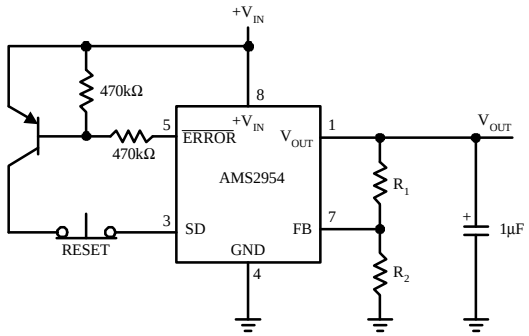
- Early warning flag on low input voltage
- Main output latches off at lower input voltages
- Battery backup on auxiliary output

Operation: Reg.#1's V_{OUT} is programmed one diode drop above 5V. It's error flag becomes active when $V_{IN} \leq 5.7V$. When V_{IN} drops below 5.3V, the error flag of Reg.#2 becomes active and via Q1 latches the main output off. When V_{IN} again exceeds 5.7V Reg.#1 is back in regulation and the early warning signal rises, unlatching Reg.#2 via D3.

1A Regulator with 1.2V Dropout

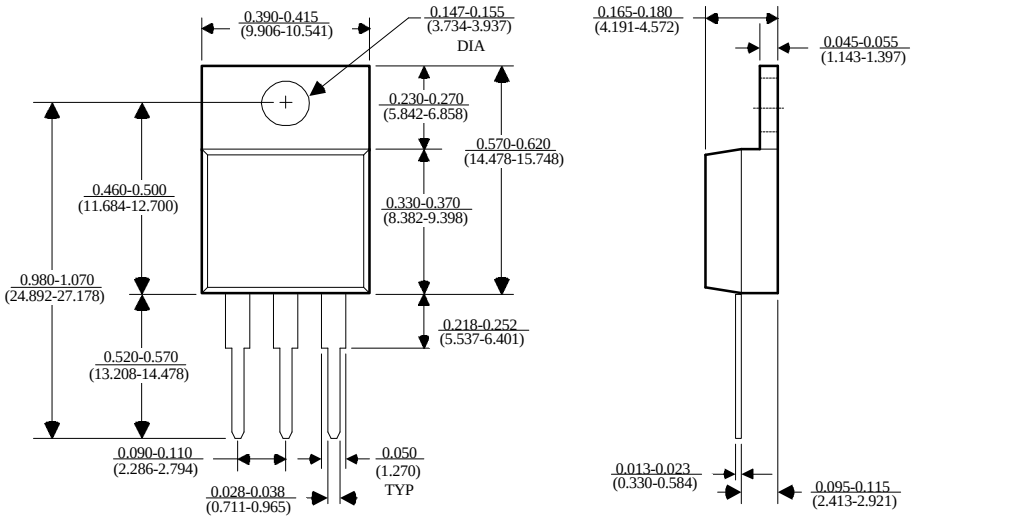


Latch Off When Error Flag Occurs

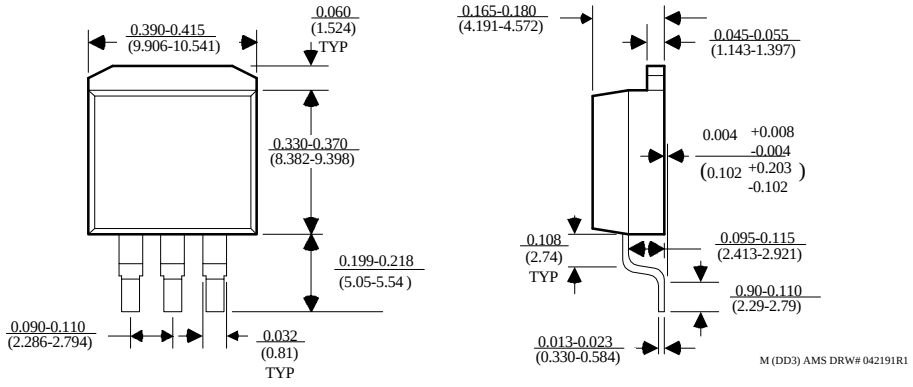


PACKAGE DIMENSIONS inches (millimeters) unless otherwise noted.

3 LEAD TO-220 PLASTIC PACKAGE (T)

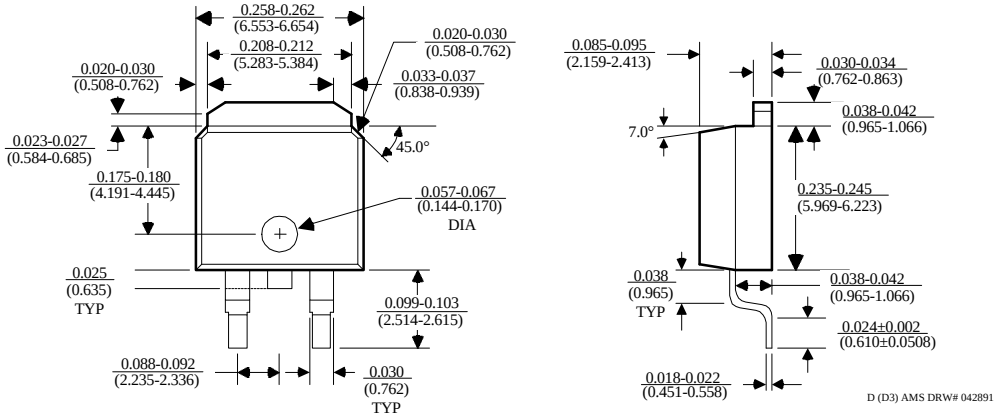


3 LEAD TO-263 PLASTIC DD (M)

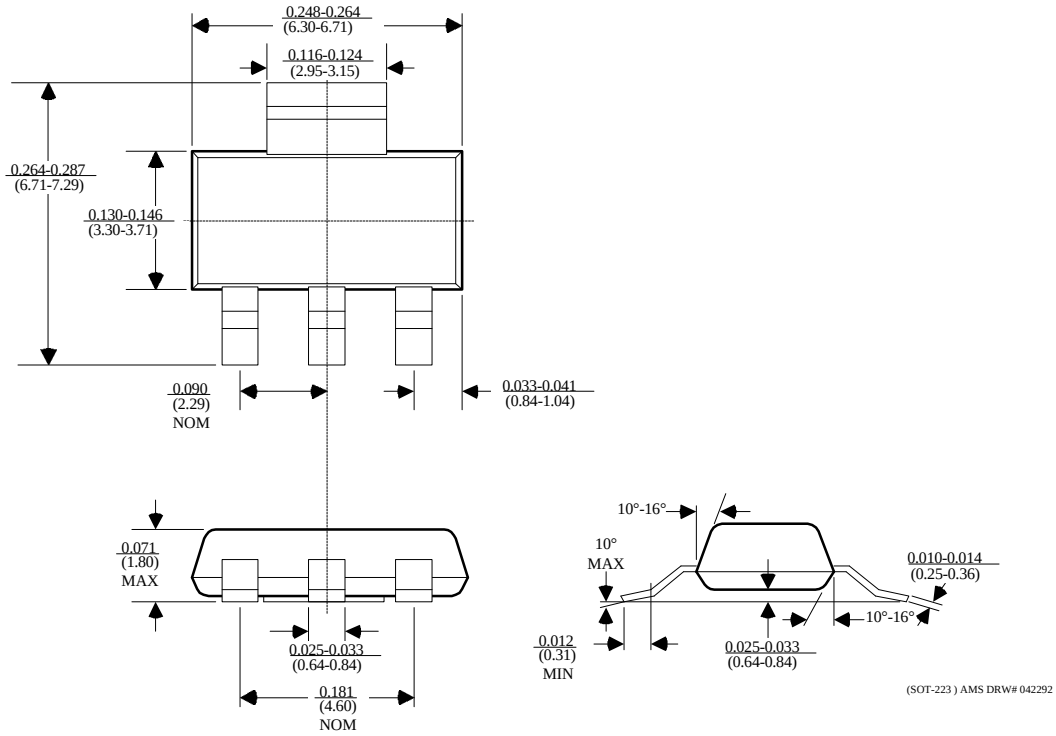


PACKAGE DIMENSIONS inches (millimeters) unless otherwise noted (Continued).

TO-252 PLASTIC PACKAGE (D)

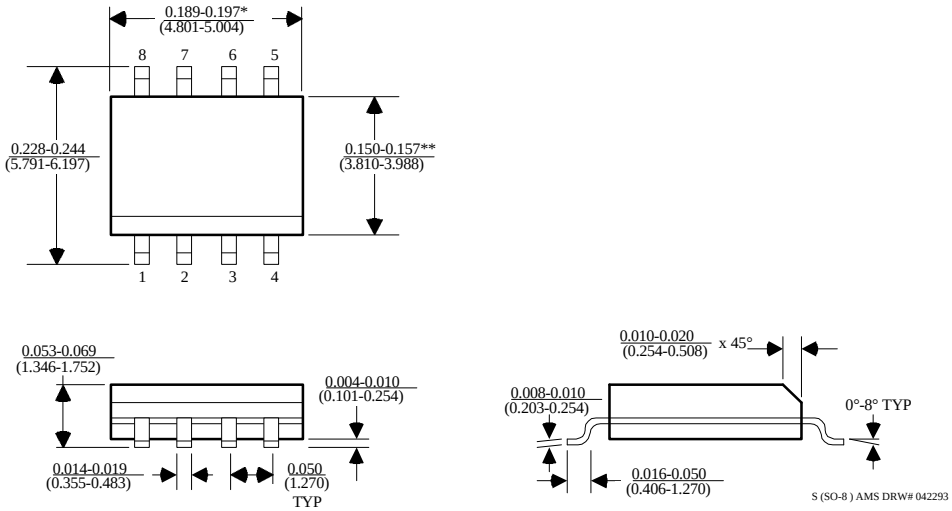


3 LEAD SOT-223 PLASTIC PACKAGE



PACKAGE DIMENSIONS inches (millimeters) unless otherwise noted (Continued).

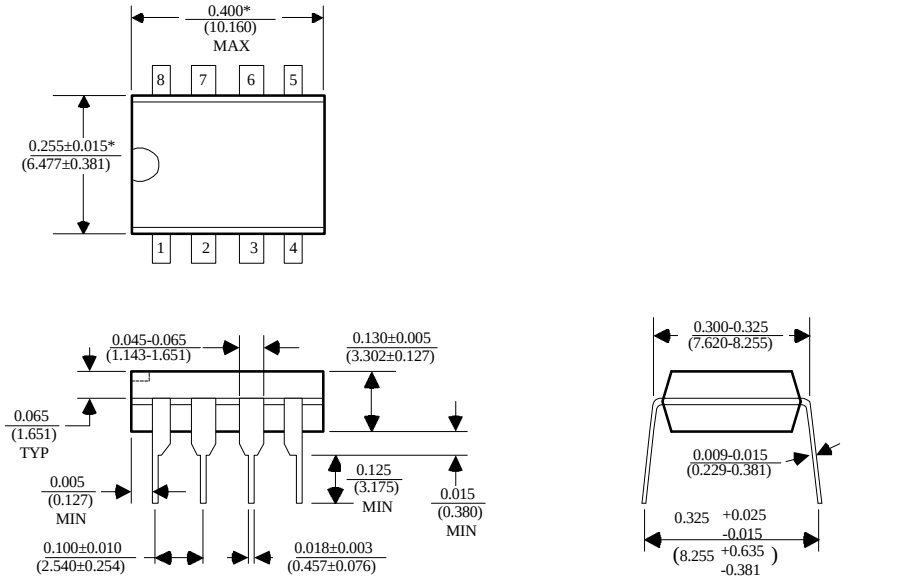
8 LEAD SOIC PLASTIC PACKAGE (S)



*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

**DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

8 LEAD PLASTIC DIP PACKAGE (P)



*DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTUSIONS. MOLD FLASH OR PROTUSIONS SHALL NOT EXCEED 0.010" (0.254mm)