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October 2008

## FPF2123-FPF2125 IntelliMAX™ Advanced Load Management Products

#### **Features**

- 1.8 to 5.5V Input Voltage Range
- Controlled Turn-On
- 0.15-1.5A Adjustable Current Limit
- Undervoltage Lockout
- Thermal Shutdown
- <2µA Shutdown Current
- Auto Restart
- Fast Current limit Response Time
  - 3µs to Moderate Over Currents
- Fault Blanking
- Reverse Current Blocking
- RoHS Compliant

### **Applications**

- PDAs
- Cell Phones
- GPS Devices
- MP3 Players
- Digital Cameras
- Peripheral Ports
- Hot Swap Supplies



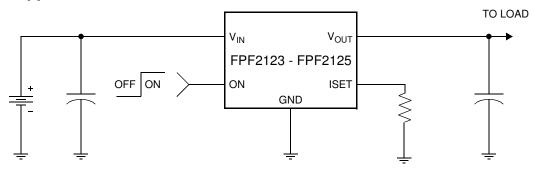
#### **General Description**

The FPF2123, FPF2124, and FPF2125 are a series of load switches which provide full protection to systems and loads which may encounter large current conditions. These devices contain a  $0.125\Omega$  current-limited P-channel MOSFET which can operate over an input voltage range of 1.8-5.5V. The current limit is settable using an external resistor. Internally, current is prevented from flowing when the MOSFET is off and the output voltage is higher than the input voltage. Switch control is by a logic input (ON) capable of interfacing directly with low voltage control signals. Each part contains thermal shutdown protection which shuts off the switch to prevent damage to the part when a continuous over-current condition causes excessive heating.

When the switch current reaches the current limit, the parts operate in a constant-current mode to prohibit excessive currents from causing damage. For the FPF2123 and FPF2124 if the constant current condition still persists after 10ms, these parts will shut off the switch. The FPF2123 has an auto-restart feature which will turn the switch on again after 160ms if the ON pin is still active. The FPF2124 does not have this auto-restart feature so the switch will remain off after a current limit fault until the ON pin is cycled. The FPF2125 will not turn off after a current limit fault, but will rather remain in the constant current mode indefinitely. The minimum current limit is 150mA.

These parts are available in a space-saving 5 pin SOT23 package

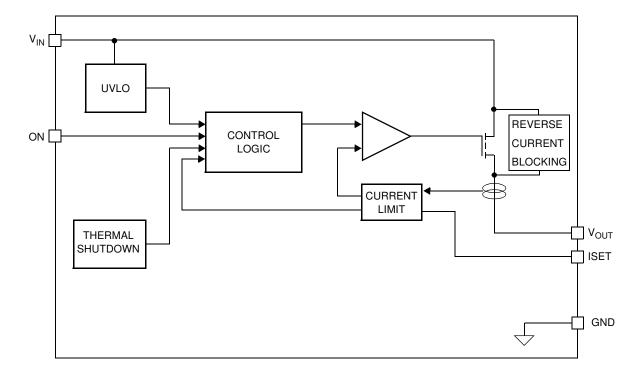
## **Typical Application Circuit**



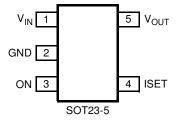
## **Ordering Information**

Part	Current Limit [A]	Current Limit Blanking Time [ms]	Auto-Restart Time [ms]	ON Pin Activity	Top Mark
FPF2123	0.15 - 1.5	5/10/20	80/160/320	Active HI	2123
FPF2124	0.15 - 1.5	5/10/20	NA	Active HI	2124
FPF2125	0.15 - 1.5	Infinite	NA	Active HI	2125

## **Functional Block Diagram**



## **Pin Configuration**



## **Pin Description**

Pin	Name	Function
1	V <sub>IN</sub>	Supply Input: Input to the power switch and the supply voltage for the IC
2	GND	Ground
3	ON	ON Control Input
4	ISET	Current Limit Set Input: A resistor from ISET to ground sets the current limit for the switch.
5	V <sub>OUT</sub>	Switch Output: Output of the power switch

## **Absolute Maximum Ratings**

Parameter	Min.	Max.	Unit	
V <sub>IN</sub> , V <sub>OUT</sub> , ON, ISET to GND		-0.3	6	V
Power Dissipation @ T <sub>A</sub> = 25 °C (note 1)			667	mW
Operating Temperature Range	-40	125	℃	
Storage Temperature	-65	150	℃	
Thermal Resistance, Junction to Ambient		150	°C/W	
Floatroatatia Disabarra Protection	НВМ	4000		V
Electrostatic Discharge Protection	MM	400		V

## **Recommended Operating Range**

Parameter	Min.	Max.	Unit
V <sub>IN</sub>	1.8	5.5	V
Ambient Operating Temperature, T <sub>A</sub>	-40	85	∞

#### **Electrical Characteristics**

 $V_{IN}$  = 1.8 to 5.5V,  $T_A$  = -40 to +85 °C unless otherwise noted. Typical values are at  $V_{IN}$  = 3.3V and  $T_A$  = 25 °C.

Parameter	Symbol	Conditions		Min.	Тур.	Max	Units
Basic Operation	l .	l		ı		ı	
Operating Voltage	V <sub>IN</sub>			1.8		5.5	V
Outresent Ouwent		I <sub>OUT</sub> = 0mA	$V_{IN} = 1.8 \text{ to } 3.3 \text{V}$		75		
Quiescent Current	IQ		$V_{IN} = 3.3 \text{ to } 5.5 \text{V}$		80	120	μA
Shutdown Current	I <sub>SHDN</sub>					2	μΑ
Reverse Block Leakage Current	I <sub>BLOCK</sub>					1	μΑ
Latch-Off Current	I <sub>LATCHOFF</sub>	FPF2124			50		μΑ
		V <sub>IN</sub> = 3.3V, I <sub>OUT</sub> = 50mA, T <sub>A</sub> = 25 °C			125	160	mΩ
On-Resistance	R <sub>ON</sub>	V <sub>IN</sub> = 3.3V, I <sub>OUT</sub> = 50mA, T <sub>A</sub> = 85 °C			150	200	
		$V_{IN} = 3.3V$ , $I_{OUT} = 50$ mA, $T_A = -40$ °C to $+85$ °C		65		200	
ON Input Logic High Voltage (ON)	V <sub>IH</sub>	V <sub>IN</sub> = 1.8V		0.75			V
		V <sub>IN</sub> = 5.5V		1.30			
	V <sub>IL</sub>	V <sub>IN</sub> = 1.8V				0.5	V
ON Input Logic Low Voltage		V <sub>IN</sub> = 5.5V				1.0	
ON Input Leakage		$V_{ON} = V_{IN}$ or GND				1	μA
Off Switch Leakage	I <sub>SWOFF</sub>	$V_{ON} = 0V, V_{OUT} = 0V$				1	μA
Protections	l .	l					
Current Limit	I <sub>LIM</sub>	$V_{IN} = 3.3V, V_{OUT} = 3.0V,$ RSET=576 $\Omega$		600	800	1000	mA
Min. Current Limit	I <sub>LIM(min.)</sub>	V <sub>IN</sub> = 3.3V, V <sub>OUT</sub> = 3.0V			150		mA
	, ,	Shutdown Threshold			140		
Thermal Shutdown		Return from Shutdown			130		℃
		Hysteresis			10		1
Under Voltage Shutdown	UVLO	V <sub>IN</sub> Increasing		1.5	1.6	1.7	٧
Under Voltage Shutdown Hysteresis					50		mV

## **Electrical Characteristics Cont.**

 $V_{IN}$  = 1.8 to 5.5V,  $T_A$  = -40 to +85 °C unless otherwise noted. Typical values are at  $V_{IN}$  = 3.3V and  $T_A$  = 25 °C.

Parameter	Symbol	Conditions		Тур.	Max	Units
Dynamic						
Turn on time	t <sub>ON</sub>	$R_L = 500\Omega, C_L = 0.1 \mu F$		25		μs
Turn off time	t <sub>OFF</sub>	$R_L = 500\Omega, C_L = 0.1 \mu F$		70		μs
V <sub>OUT</sub> Rise Time	t <sub>R</sub>	$R_L = 500\Omega, C_L = 0.1 \mu F$		12		μs
V <sub>OUT</sub> Fall Time	t <sub>F</sub>	$R_L = 500\Omega, C_L = 0.1 \mu F$		200		μs
Over Current Blanking Time	t <sub>BLANK</sub>	FPF2123, FPF2124	5	10	20	ms
Auto-Restart Time		FPF2123	80	160	320	ma
Auto-nestart fillie	<sup>T</sup> RESTART	FPF2124, FPF2125		NA		ms
Short Circuit Response Time		V <sub>IN</sub> = V <sub>ON</sub> = 3.3V. Moderate Over-Current Condition.		3		μs
		V <sub>IN</sub> = V <sub>ON</sub> = 3.3V. Hard Short.		20		μs

Note 1: Package power dissipation on 1square inch pad, 2 oz. copper board.

## **Typical Characteristics**

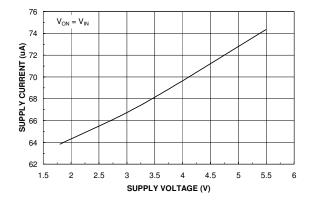


Figure 1. Quiescent Current vs. Input Voltage

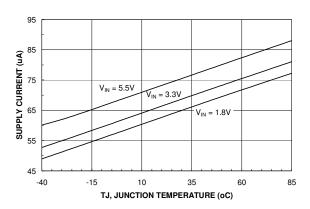


Figure 2. Quiescent Current vs. Temperature

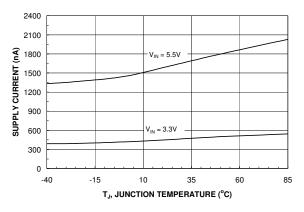


Figure 3. I<sub>SHUTDOWN</sub> Current vs. Temperature

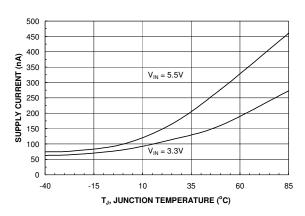


Figure 4. I<sub>SWITCH-OFF</sub> Current vs. Temperature

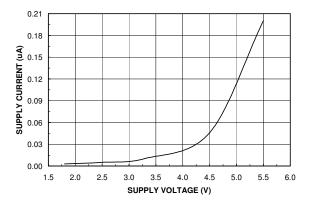


Figure 5. Reverse Current vs.  $V_{OUT}$ 

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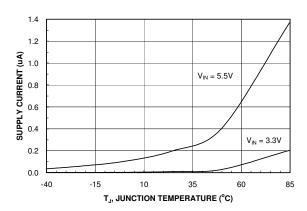


Figure 6. Reverse Current vs. Temperature

## **Typical Characteristics**

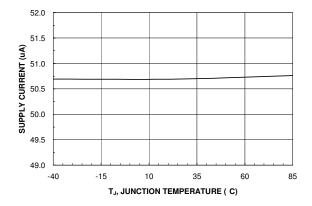


Figure 7. I<sub>LATCH-OFF</sub> Current vs. Temperature

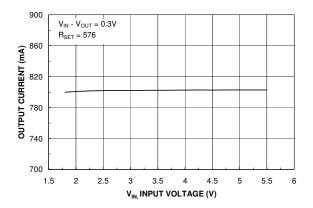


Figure 8. Current Limit vs. Input Voltage

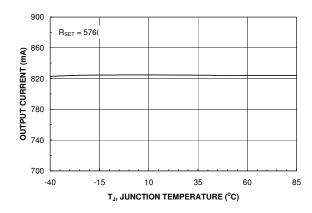


Figure 9. Current Limit vs. Temperature

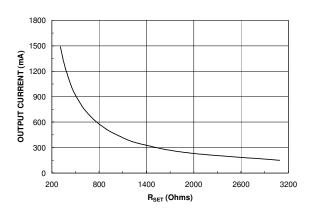


Figure 10. Current Limit vs. Rest

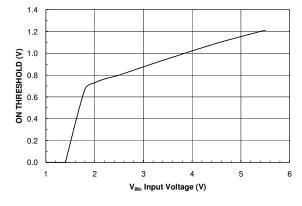


Figure 11.  $\,\mathrm{V_{IH}}\,\mathrm{vs.}\,\mathrm{V_{IN}}$ 

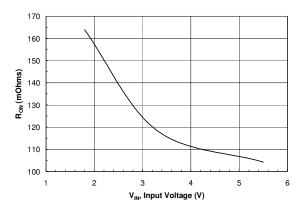


Figure 12. R<sub>ON</sub> vs. V<sub>IN</sub>

## **Typical Characteristics**

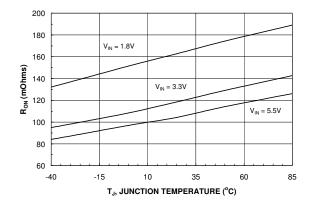


Figure 13.  $R_{(ON)}$  vs. Temperature

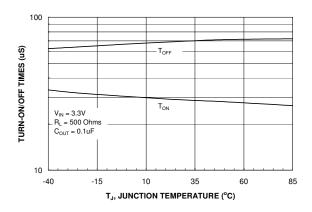


Figure 14.  $T_{ON}/T_{Off}$  vs. Temperature

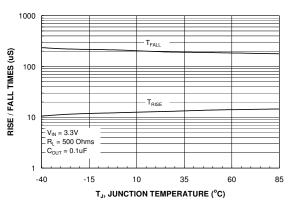


Figure 15.  $T_{RISE}/T_{FALL}$  vs. Temperature

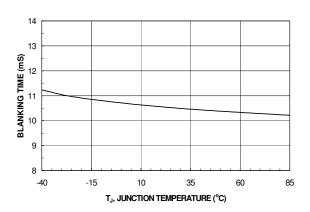


Figure 16.  $T_{BLANK}$  vs. Temperature

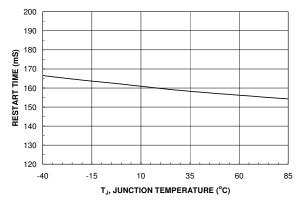


Figure 17. T<sub>RESTART</sub> vs. Temperature

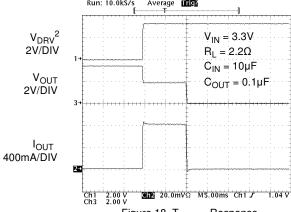


Figure 18. T<sub>BLANK</sub> Response

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#### **Typical Characteristics** Average IIII Run: 2.50kS/s Run: 500kS/s Average IIII $V_{DRV}^2$ V<sub>ON</sub> 2V/DIV $V_{IN} = 3.3V$ $V_{IN} = 3.3V$ 2V/DIV $R_L = 500\Omega$ $R_L = 2.2\Omega$ $C_{IN} = 10 \mu F$ $C_{IN} = 10 \mu F$ $C_{OUT} = 0.1 \mu F$ $V_{OUT}$ $C_{OUT} = 0.1 \mu F$ 2V/DIV I<sub>OUT</sub> 10mA/DIV $I_{OUT}$ 400mA/DIV Ch2 20.0mVΩ M20.0ms Ch1 J Ch2 10.0mVΩ M 100μs Ch1 🗲 Figure 19. $T_{RESTART}$ Response Figure 20. T<sub>ON</sub> Response Run: 250MS/s Average IIII Run: 2.50MS/s Average IIIG V<sub>ON</sub> 2V/DIV V<sub>IN</sub> 2V/DIV $V_{IN} = 3.3V$ $R_L = 500\Omega$ $C_{IN} = 10 \mu F$ $C_{IN} = 10 \mu F$ $C_{OUT} = 0.1 \mu F$ $C_{OUT} = 0.1 \mu F$ $I_{OUT}$ I<sub>OUT</sub> 10mA/DIV 4A/DIV $V_{OUT}$ 2V/DIV **©12** 20.0mVΩ M 20.0μs Ch3 \ **Ch1** 2.00 V Ch2 10.0mVΩ M 200ns Ch1 \ Figure 22. Short Circuit Response Figure 21. T<sub>OFF</sub> Response (Output Shorted to GND) Run: 1.00MS/s Average IIIgr Run: 1.00MS/s Average III $V_{\text{IN}}$ 2V/DIV $C_{IN} = 10 \mu F$ $R_L = 2.2\Omega$ $V_{IN} = V_{ON}$ $C_{OUT} = 0.1 \mu F$ $C_{IN} = 10 \mu F$ 2V/DIV $V_{ON}$ $C_{OUT} = 0.1 \mu F$ 2V/DIV I<sub>OUT</sub> 400mA/DIV I<sub>OUT</sub> 400mA/DIV Ch2 20.0mVΩ M 50.0μs Ch3 J Ch1 2.00 V (12) 20.0 mVΩ M 50.0 μs Ch1 F 960 mV Figure 23. Current Limit Response Figure 24. Current Limit Response (Switch power up to hard short) (Output Shorted to GND by 2.2Ω, moderate short)

Note 2:  $V_{DRV}$  signal forces the device to go into overcurrent condition by loading a 2.2 $\Omega$  resistor.

#### **Description of Operation**

The FPF2123, FPF2124, and FPF2125 are current limited switches that protect systems and loads which can be damaged or disrupted by the application of high currents. The core of each device is a 0.125 $\Omega$  P-channel MOSFET and a controller capable of functioning over a wide input operating range of 1.8-5.5V. The controller protects against system malfunctions through current limiting under-voltage lockout and thermal shutdown. The current limit is adjustable from 150mA to 1.5A through the selection of an external resistor.

#### On/Off Control

The ON pin controls the state of the switch. When ON is high, the switch is in the on state. Activating ON continuously holds the switch in the on state so long as there is no fault. For all versions, an under-voltage on  $\rm V_{IN}$  or a junction temperature in excess of 140 °C overrides the ON control to turn off the switch. In addition, excessive currents will cause the switch to turn off in the FPF2123 and FPF2124. The FPF2123 has an Auto-Restart feature which will automatically turn the switch on again after 160ms. For the FPF2124, the ON pin must be toggled to turn-on the switch again. The FPF2125 does not turn off in response to an over current condition but instead remains operating in a constant current mode so long as ON is active and the thermal shutdown or under-voltage lockout have not activated.

The ON pin control voltage and  $V_{\rm IN}$  pin have independent recommended operating ranges. The ON pin voltage can be driven by a voltage level higher than the input voltage.

#### **Current Limiting**

FPF2123-FPF2125 Rev. F

The current limit ensures that the current through the switch doesn't exceed a maximum value while not limiting at less than a minimum value. The current at which the parts will limit is adjustable through the selection of an external resistor connected to ISET. Information for selecting the resistor is found in the Application Info section. The FPF2123 and FPF2124 have a blanking time of 10ms, nominally, during which the switch will act as a constant current source. At the end of the blanking time, the switch will be turned-off. The FPF2125 has no current limit blanking period so it will remain in a constant current state until the ON pin is deactivated or the thermal shutdown turns-off the switch.

#### **Under-Voltage Lockout**

The under-voltage lockout turns-off the switch if the input voltage drops below the under-voltage lockout threshold. With the ON pin active, the input voltage rising above the under-voltage lockout threshold will cause a controlled turn-on of the switch which limits current over-shoots.

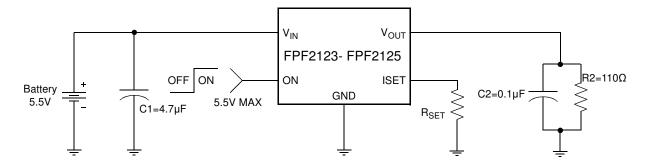
#### **Thermal Shutdown**

The thermal shutdown protects the die from internally or externally generated excessive temperatures. During an over-temperature condition the switch is turned-off. The switch automatically turns-on again if the temperature of the die drops below the threshold temperature.

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### **Application Information**

#### **Typical Application**



#### **Setting Current Limit**

The FPF2123, FPF2124, and FPF2125 have a current limit which is set with an external resistor connected between ISET and GND. This resistor is selected by using the following equation,

$$R_{SET} = \frac{460}{I_{LIM}}$$

 $R_{\mbox{\footnotesize SET}}$  is in Ohms and that of  $I_{\mbox{\footnotesize LIM}}$  is Amps

The table below can also be used to select  $R_{SET}$ . A typical application would be the 500mA current that is required by a single USB port. Using the table below an appropriate selection for the  $R_{SET}$  resistor would be  $604\Omega$ . This will ensure that the port load could draw 570mA, but not more than 950mA. Likewise for a dual port system, an  $R_{SET}$  of  $340\Omega$  would always deliver at least 1120mA and never more than 1860mA.

#### **Input Capacitor**

To limit the voltage drop on the input supply caused by transient in-rush currents when the switch turns-on into a discharged load capacitance or a short-circuit, a capacitor needs to be placed between  $V_{IN}$  and GND. A 4.7 $\mu F$  ceramic capacitor,  $C_{IN}$ , must be placed close to the  $V_{IN}$  pin. A higher value of  $C_{IN}$  can be used to further reduce the voltage drop experienced as the switch is turned on into a large capacitive load.

#### **Output Capacitor**

A  $0.1\mu F$  capacitor,  $C_{OUT}$ , should be placed between  $V_{OUT}$  and GND. This capacitor will prevent parasitic board inductances from forcing  $V_{OUT}$  below GND when the switch turns-off. For the FPF2123 and FPF2124, the total output capacitance needs to be kept below a maximum value,  $C_{OUT(max)}$ , to prevent the part from registering an over-current condition and turning-off the switch. The maximum output capacitance can be determined from the following formula,

$$C_{OUT}(max) = \frac{I_{LIM}(min) \times t_{BLANK}(min)}{V_{IN}}$$

Current Limit Various R<sub>SET</sub> Values

		JEI	
R <sub>SET</sub> [Ω]	Min. Current Limit [mA]	Typ. Current Limit [mA]	Max. Current Limit [mA]
309	1120	1490	1860
340	1010	1350	1690
374	920	1230	1540
412	840	1120	1400
453	760	1010	1270
499	690	920	1150
549	630	840	1050
576	600	800	1000
604	570	760	950
732	470	630	790
887	390	520	650
1070	320	430	540
1300	260	350	440
1910	180	240	300
3090	110	150	190

#### **Power Dissipation**

During normal operation as a switch, the power dissipated in the part will depend upon the level at which the current limit is set. The maximum allowed setting for the current limit is 1.5A and this will result in a typical power dissipation of,

$$P = (I_{LIM})^2 \times R_{ON} = (1.5)^2 \times 0.125 = 281 \text{mW}$$

If the part goes into current limit the maximum power dissipation will occur when the output is shorted to ground. For the FPF2123 the power dissipation will scale by the Auto-Restart Time,  $t_{\mbox{\scriptsize RESTART}}$ , and the Over Current Blanking Time,  $t_{\mbox{\scriptsize BLANK}}$ , so that the maximum power dissipated is,

$$P(max) = \frac{t_{BLANK}(max)}{t_{RESTART}(min) + t_{BLANK}(max)} \times V_{IN}(max) \times I_{LIM}(max) \quad (4)$$
$$= \frac{20}{80 + 20} \times 5.5 \times 1.5 = 1.65W$$

This is more power than the package can dissipate, but the thermal shutdown of the part will activate to protect the part from damage due to excessive heating. When using the FPF2124, attention must be given to the manual resetting of the part. Continuously resetting the part when a short on the output is present will cause the temperature of the part to increase. The junction temperature will only be able to increase to the thermal shutdown threshold. Once this temperature has been reached, toggling ON will not turn-on the switch until the junction temperature drops. For the FPF2125, a short on the output will cause the part to operate in a constant current state dissipating a worst case power of,

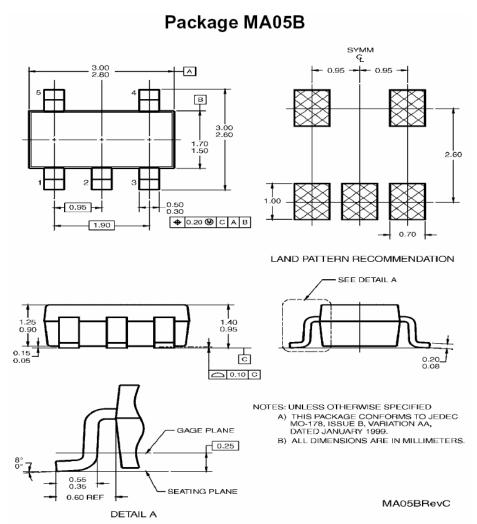
$$P(max) = V_{1N}(max) \times I_{LIM}(max)$$
= 5.5 \times 1.5 = 8.25 W (5)

This large amount of power will activate the thermal shutdown and the part will cycle in and out of thermal shutdown so long as the ON pin is active and the short is present.

#### **Board Layout**

For best performance, all traces should be as short as possible. To be most effective, the input and output capacitors should be placed close to the device to minimize the effects that parasitic trace inductances may have on normal and short-circuit operation. Using wide traces for  $V_{\text{IN}}$ ,  $V_{\text{OUT}}$  and GND will help minimize parasitic electrical effects along with minimizing the case to ambient thermal impedance.

## **Dimensional Outline and Pad Layout**







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