

# High-Voltage, Current-Mode PWM Controller

#### **Features**

- ▶ 10 to 450V input voltage range
- <1.3mA supply current</p>
- >1.0MHz clock
- >20:1 dynamic range @ 500KHz
- 99% Maximum duty cycle version
- Low internal noise

### **Applications**

- Off-line high frequency power supplies
- Universal input power supplies
- High density power supplies
- Very high efficiency power supplies
- Extra wide load range power supplies

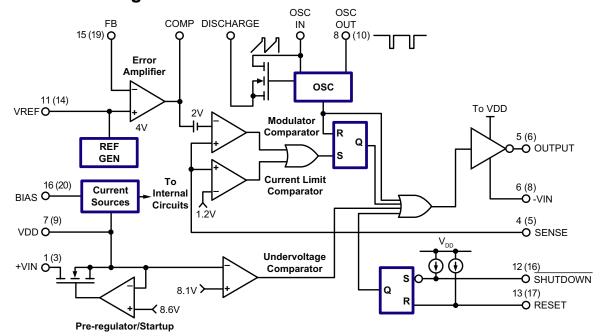
### **General Description**

The Supertex HV9123 is a Switch Mode Power Supply (SMPS) controller subsystem that can start and run directly from almost any DC input, from a 12V battery to a rectified and filtered 240VAC line. It contains all the elements required to build a single-switch converter except for the switch, magnetic assembly, output rectifier(s) and filter(s).

A unique input circuit allows the HV9123 to self-start directly from a high voltage input, and subsequently take the power to operate from one of the outputs of the converter it is controlling, allowing very efficient operation while maintaining input-to-output galvanic isolation limited in voltage only by the insulation system of the associated magnetic assembly. A ±2% internal bandgap reference, internal operational amplifier, very high speed comparator, and output buffer allow production of rugged, high performance, high efficiency power supplies of 50W or more, which can still be over 80% efficient at outputs of 1.0W or less. The wide dynamic range of the controller system allows designs with extremely wide line and load variations with much less difficulty and much higher efficiency than usual. The exceptionally wide input voltage range also allows better usage of energy stored in input dropout capacitors than with other PWM ICs. Remote on/off controls allow either latching or nonlatching remote shutdown. During shutdown, the power required is under 6.0mW.

For detailed circuit and application information, please refer to application notes AN-H13, AN-H21 to AN-H24.

### **Functional Block Diagram**



Note:

Pin numbers in parentheses are for PLCC package.

### **Ordering Information**

		Package Options	
Device	16-Lead SOIC 9.90x3.90mm body 1.75mm height (max) 1.27mm pitch	16-Lead PDIP .790x.250in body .210in height (max) .100in pitch	20-Lead PLCC .353x.353in body .180in height (max) .050in pitch
HV9123	HV9123NG-G	HV9123P-G	HV9123PJ-G

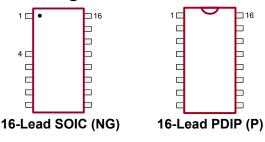


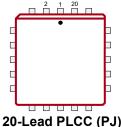
### **Absolute Maximum Ratings**

Parameter	Value
Input voltage, V <sub>IN</sub>	450V
Device supply voltage, V <sub>DD</sub>	15.5V
Logic input voltage	-0.3V to V <sub>DD</sub> +0.3V
Linear input voltage	-0.3V to V <sub>DD</sub> +0.3V
Preregulator input current, I <sub>IN</sub> (continuous)	2.5mA
Operating junction temperature (T <sub>J</sub> )	150°C
Storage temperature	-65°C to +150°C
Power dissipation:	
16-Lead SOIC	900mW
16-Lead PDIP	1000mW
20-Lead PLCC	1400mW

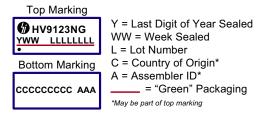
Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **Pin Configurations**



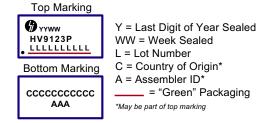


### **Product Marking**



Package may or may not include the following marks: Si or

#### 16-Lead SOIC (NG)



Package may or may not include the following marks: Si or

#### 16-Lead PDIP (P)



Package may or may not include the following marks: Si or

20-Lead PLCC (PJ)

### **Electrical Characteristics**

(Unless otherwise specified,  $V_{DD}$  = 10V, + $V_{IN}$  = 48V, Discharge = - $V_{IN}$  = 0V,  $R_{BIAS}$  = 390K $\Omega$ ,  $R_{OSC}$  = 330K $\Omega$ ,  $T_A$  = 25°C.)

			Тур	Max	Units	Conditions			
ce									
Output valtage		3.92	4.00	4.08		R <sub>L</sub> = 10MΩ			
Output voltage	-	3.84	4.00	4.16	] <b>V</b>	$R_{L} = 10M\Omega, T_{A} = -55 \text{ to } 125^{\circ}\text{C}$			
Output impedance	#	15	30	45	ΚΩ				
Short circuit current	-	-	125	250	μA	V <sub>REF</sub> = -V <sub>IN</sub>			
Change in V <sub>REF</sub> with temperature	#	-	0.25	-	mV/°C	T <sub>A</sub> = -55 to 125°C			
or									
Oscillator frequency	-	1.0	3.0	-	MHz	$R_{OSC} = 0\Omega$			
haitial account of	-	80	100	120	121.1-	R <sub>osc</sub> = 330KΩ			
initial accuracy	-	160	200	240	KHZ	$R_{\rm osc}$ = 150K $\Omega$			
Voltage stability	-	-	-	15	%	9.5V < V <sub>DD</sub> < 13.5V			
Sc Voltage stability  Temperature coefficient		-	170	-	ppm/°C	T <sub>A</sub> = -55 to 125°C			
Maximum duty cycle	#	95	97	99	%				
Deadtime	#	-	225	-	ns				
Minimum duty cycle	-	-	-	0	%				
Maximum pulse width before pulse drops out	#	-	80	125	ns				
Limit									
Maximum input signal	-	1.0	1.2	1.4	V	V <sub>FB</sub> = 0V			
Delay to output	#	-	80	120	ns	V <sub>SENSE</sub> = 1.5V, V <sub>COMP</sub> ≤ 2.0V			
nplifier									
Feedback voltage	-	3.92	4.00	4.08	V	V <sub>FB</sub> shorted to COMP			
Input bias current	-	-	25	500	nA	V <sub>FB</sub> = 4.0V			
Input offset voltage	-	nulled	during t	rim	-				
Open loop voltage gain	#	60	80	-	dB				
Unity gain bandwidth	#	1.0	1.3	-	MHz				
Out impedance	#	see Fig. 1		Ω					
Output source current	-	-1.4	-2.0	-	mA	V <sub>FB</sub> = 3.4V			
Output sink current	-	0.12	0.15	-	mA	V <sub>FB</sub> = 4.5V			
Power supply rejection	#	se	e Fig. 2	1	dB				
	Short circuit current Change in V <sub>REF</sub> with temperature or Oscillator frequency Initial accuracy¹ Voltage stability Temperature coefficient  Maximum duty cycle Deadtime Minimum duty cycle Maximum pulse width before pulse drops out Limit Maximum input signal Delay to output aplifier Feedback voltage Input bias current Input offset voltage Open loop voltage gain Unity gain bandwidth Out impedance Output source current Output sink current	Output impedance # Short circuit current - Change in V <sub>REF</sub> with temperature #  or Oscillator frequency - Initial accuracy¹ - Voltage stability - Temperature coefficient #  Maximum duty cycle # Deadtime # Minimum duty cycle - Maximum pulse width before pulse drops out #  Limit Maximum input signal - Delay to output #  applifier Feedback voltage - Input bias current - Input offset voltage gain # Unity gain bandwidth # Out impedance # Output sink current - Output sink current - Output sink current - Output sink current -	Output voltage  Output impedance # 15  Short circuit current	Output voltage	Output voltage  Output impedance # 15 30 45  Short circuit current - 125 250  Change in V <sub>REF</sub> with temperature # - 0.25 -  Initial accuracy'  Voltage stability - 160 200 240  Voltage stability 170 -  Maximum duty cycle # 95 97 99  Deadtime # - 225 -  Minimum duty cycle # 95 97 99  Deadtime # - 225 -  Minimum duty cycle # - 80 125  Limit  Maximum pulse width before pulse drops out # - 80 125  Limit  Maximum input signal - 1.0 1.2 1.4  Delay to output # - 80 120  Input offset voltage - nulled during trim  Open loop voltage gain # 60 80 -  Unity gain bandwidth # 1.0 1.3 -  Output source current1.4 -2.0 -	Output voltage         -         3.84         4.00         4.16         V           Short circuit current         -         -         15         30         45         KΩ           Short circuit current         -         -         125         250         μA           Change in V <sub>REF</sub> with temperature         #         -         0.25         -         mV/°C           Volcage in Value of the part			

- # Guaranteed by design.
  1. Stray capacitance on OSC In pin must be ≤5pF.

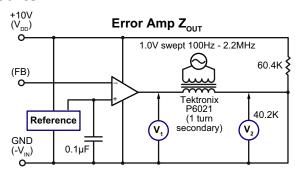
Electrical Characteristics (cont.) (Unless otherwise specified,  $V_{DD}$  = 10V,  $+V_{IN}$  = 48V, Discharge =  $-V_{IN}$  = 0V,  $R_{BIAS}$  = 390K $\Omega$ ,  $R_{OSC}$  = 330K $\Omega$ ,  $T_A$  = 25°C.)

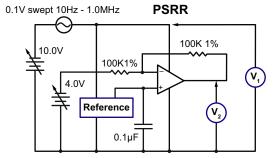
Pre-regutation/Startup	Sym	Parameter		#	Min	Тур	Max	Units	Conditions
+I <sub>N</sub> Input leakage current         -         -         10         μA         V <sub>Oo</sub> > 9.4V           V <sub>TM</sub> V <sub>Do</sub> pre-regulator turn-off threshold voltage         -         8.0         8.7         9.4         V         I <sub>PRERES</sub> = 10μA           Supply           Undervoltage lockout         -         7.0         8.1         8.9         V            Supply           I <sub>DO</sub> Supply current         -         -         0.75         1.3         mA         C <sub>C</sub> < 75pF	Pre-regu	ulator/Startup		•			•		
V   V   V   V   V   V   V   V   V   V	+V <sub>IN</sub>	Input voltage		-	10	-	450	V	$I_{IN} < 10 \mu A; V_{CC} > 9.4 V$
V   V   V   V   V   V   V   V   V   V	+1	Input leakage current			-	-	10	μA	V <sub>DD</sub> > 9.4V
Supply current         -         -         0.75         1.3         mA         C₁ < 75pF           I₀ Oulsescent supply current         -         -         0.55         -         mA         SHUTDOWN = -V <sub>N</sub> Include the property of the property		V <sub>DD</sub> pre-regulator turn-off threshold voltage			8.0	8.7	9.4	V	
Supply current         -         -         0.75         1.3         mA         C₁ < 75pF           I₀ Oulsescent supply current         -         -         0.55         -         mA         SHUTDOWN = -V <sub>N</sub> Include the property of the property	V <sub>LOCK</sub>	Undervoltage lockou	t	-	7.0	8.1	8.9	V	
I									
Nominal bias current   -   -   20   -   μA	I <sub>DD</sub>	Supply current		_	-	0.75	1.3	mA	C <sub>L</sub> < 75pF
Shutdown Logic   Test	I <sub>Q</sub>	Quiescent supply cur	rent	_	-	0.55	-	mA	SHUTDOWN = -V <sub>IN</sub>
Shutdown Logic   Total Shutdown   To	I <sub>BIAS</sub>	Nominal bias current		_	-	20	-	μΑ	
The state of th	V <sub>DD</sub>	Operating range		-	9.0	-	13.5	V	
t <sub>sw</sub> SHUTDOWN pulse width         #         50         -         -         ns            t <sub>RW</sub> RESET pulse width         #         50         -         -         ns            t <sub>LW</sub> Latching pulse width         #         25         -         -         ns         SHUTDOWN and RESET low           V <sub>IL</sub> Input low voltage         -         -         -         2.0         V            V <sub>IH</sub> Input low voltage         -         -         -         2.0         V            I <sub>IH</sub> Input current, input high voltage         -         -         1.0         5.0         μA         V <sub>IN</sub> = V <sub>DD</sub> I <sub>IL</sub> Input current, input low voltage         -         -         -25         -35         μA         V <sub>IN</sub> = 0V           Output           V <sub>OH</sub> Output high voltage         -         -         -25         -3         V         I <sub>OUT</sub> = 10mA         I <sub>OUT</sub> = 10mA         I <sub>I</sub> = -55 to 125°C           V <sub>OL</sub> Output high voltage         -         -         -         -         V         I <sub>OUT</sub> = 10mA         I <sub>I</sub> = -55 to 125°C           V <sub>OL</sub>	Shutdov	wn Logic							
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	t <sub>sd</sub>	SHUTDOWN delay		#	-	50	100	ns	$C_L = 500pF, V_{SENSE} = -V_{IN}$
The second color of the	t <sub>sw</sub>	SHUTDOWN pulse width		#	50	-	-	ns	
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	t <sub>RW</sub>	RESET pulse width		#	50	-	-	ns	
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	t <sub>LW</sub>	Latching pulse width		#	25	-	-	ns	SHUTDOWN and RESET low
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	V <sub>IL</sub>	Input low voltage		-	-	-	2.0	V	
Input current, input low voltage   -   -   -25   -35   μA   V <sub>IN</sub> = 0V	V <sub>IH</sub>	Input high voltage		-	7.0	-	-	V	
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	I <sub>IH</sub>	Input current, input h	igh voltage	-	-	1.0	5.0	μA	$V_{IN} = V_{DD}$
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	I <sub>IL</sub>	Input current, input lo	w voltage	-	-	-25	-35	μA	V <sub>IN</sub> = 0V
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	Output								
$V_{\text{OL}} = -55 \text{ to } 125^{\circ}\text{C}$ $V_{\text{OL}} = -10 \text{ mA}$ $V_{\text{DU}} = -10 \text{ mA}$ $V_{\text{DU}} = -10 \text{ mA}$ $V_{\text{A}} = -55 \text{ to } 125^{\circ}\text{C}$ $V_{\text{DU}} = -10 \text{ mA}$ $V_{\text{A}} = -55 \text{ to } 125^{\circ}\text{C}$ $V_{\text{DU}} = -10 \text{ mA}$ $V_{\text{A}} = -55 \text{ to } 125^{\circ}\text{C}$ $V_{\text{DU}} = -10 \text{ mA}$ $V_{\text{DU}} = \pm 10 \text{ mA}$ $V_{D$				-	V <sub>DD</sub> -0.25	-	-	V	I <sub>OUT</sub> = 10mA
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	V <sub>OH</sub>	Output high voltage		-	V <sub>DD</sub> -0.3	-	-	V	I <sub>OUT</sub> = 10mA, T <sub>A</sub> = -55 to 125°C
$R_{\text{OUT}} = \frac{-1}{25} = -$				_	-	-	0.2	V	I <sub>OUT</sub> = -10mA
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	V <sub>OL</sub>	Output low voltage		-	-	-	0.3	V	I <sub>OUT</sub> = -10mA, T <sub>A</sub> = -55 to 125°C
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$			Pull up	-	-	15	25	Ω	- L - +10mΛ
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	P	Output resistance	Pull down	-	-	8.0	20	Ω	OUT - TIOIIIA
Pull down 10 30 Ω $T_A$ = -55 to 125°C $t_R$ Rise time # - 30 75 ns $C_L$ = 500pF	OUT	Output resistance	Pull up	-	-	20	30	Ω	I <sub>out</sub> = ±10mA,
			Pull down	-	-	10	30	Ω	T <sub>A</sub> = -55 to 125°C
$t_{\rm F}$ Fall time # - 20 75 ns $C_{\rm L}$ = 500pF	t <sub>R</sub>	Rise time		#	-	30	75	ns	C <sub>L</sub> = 500pF
	t <sub>F</sub>	Fall time		#	-	20	75	ns	C <sub>L</sub> = 500pF

Note:

# Guaranteed by design.

#### **Test Circuits**





Note:

Set feedback voltage so that  $V_{COMP} = V_{DIVIDE} \pm 1.0 \text{mV}$  before connecting transformer.

### **Detailed Description**

#### **Preregulator**

The preregulator/startup circuit for the HV9123 consists of a high-voltage n-channel depletion-mode DMOS transistor driven by an error amplifier to form a variable current path between the VIN terminal and the VDD terminal. Maximum current (about 20 mA) occurs when  $V_{\rm DD}$  = 0, with current reducing as  $V_{\rm DD}$  rises. This path shuts off altogether when  $V_{\rm DD}$  rises to somewhere between 7.8 and 9.4V, so that if  $V_{\rm DD}$  is held at 10 or 12V by an external source (generally the supply the chip is controlling), no current other than leakage is drawn through the high voltage transistor. This minimizes dissipation.

An external capacitor between VDD and VSS is generally required to store energy used by the chip in the time between shutoff of the high voltage path and the VDD supply's output rising enough to take over powering the chip. This capacitor should have a value of 100X or more the effective gate capacitance of the MOSFET being driven, i.e.,

$$C_{STORAGE} \ge 100 x$$
 (gate charge of FET at 10V)

as well as very good high frequency characteristics. Stacked polyester or ceramic caps work well. Electrolytic capacitors are generally not suitable. A common resistor divider string is used to monitor  $V_{\text{DD}}$  for both the undervoltage lockout circuit and the shutoff circuit of the high voltage FET. Setting the undervoltage sense point about 0.6V lower on the string than the FET shutoff point guarantees that the undervoltage lockout always releases before the FET shuts off.

#### **Bias Circuit**

An external bias resistor, connected between the bias pin and VSS is required by the HV9123 to set currents in a series of current mirrors used by the analog sections of the chip. Nominal external bias current requirement is 15 to 20μA, which can be set by a 390 to 510KΩ resistor if a 10V  $V_{DD}$  is used, or a 510 to 680KΩ resistor if  $V_{DD}$  will be 12V. A precision resistor is not required;  $\pm$  5% is fine.

#### **Clock Oscillator**

The clock oscillator of the HV9123 consists of a ring of CMOS inverters, timing capacitors, and a capacitor discharge FET. A single external resistor between the OSC IN and OSC OUT pins is required to set oscillator frequency (see graph). The discharge can either be connected to VSS directly or connected to VSS through a resistor used to set a dead time.

One difference exists between the Supertex HV9123 and competitive 9123s: The oscillator is shut off when a shutoff command is received. This saves about 150 $\mu$ A of quiescent current, which aids in the construction of power supplies to meet CCITT specification I-430, and in other situations where an absolute minimum of quiescent power dissipation is required.

#### Reference

The Reference of the HV9123 consists of a stable bandgap reference followed by a buffer amplifier which scales the voltage up to approximately 4.0V. The scaling resistors of the reference buffer amplifier are trimmed during manufacture so that the output of the error amplifier, when connected in a gain of -1 configuration, is as close to 4.0V as possible. This nulls out any input offset of the error amplifier. As a consequence, even though the observed reference voltage of a specific part may not be exactly 4.0V, the feedback voltage required for proper regulation will be.

A  $\approx$  50K $\Omega$  resistor is placed internally between the output of the reference buffer amplifier and the circuitry it feeds (reference output pin and non-inverting input to the error

amplifier). This allows overriding the internal reference with a low-impedance voltage source ≤6.0V. Using an external reference reinstates the input offset voltage of the error amplifier, and its effect of the exact value of feedback voltage required. In general, because the reference voltage of the Supertex HV9123 is not noisy, as some previous examples have been, overriding the reference should seldom be necessary.

Because the reference of the HV9123 is a high impedance node, and usually there will be significant electrical noise near it, a bypass capacitor between the reference pin and VSS is strongly recommended. The reference buffer amplifier is intentionally compensated to be stable with a capacitive load of 0.01 to  $0.1\mu F$ .

#### **Error Amplifier**

The error amplifier in the HV9123 is a true low-power differential input operational amplifier intended for around-the-amplifier compensation. It is of mixed CMOS-bipolar construction: A PMOS input stage is used so the common-mode range includes ground and the input impedance is very high. This is followed by bipolar gain stages which provide high gain without the electrical noise of all-MOS amplifiers. The

amplifier is unity-gain stable.

#### **Current Sense Comparators**

The HV9123 uses a true dual-comparator system with independent comparators for modulation and current limiting. This allows the designer greater latitude in compensation design, as there are no clamps (except ESD protection) on the compensation pin. Like the error amplifier, the comparators are of low-noise BiCMOS construction.

#### Remote Shutdown

The SHUTDOWN and RESET pins of the HV9123 can be used to perform either latching or non-latching shutdown of a converter as required. These pins have internal current source pull-ups so they can be driven from open-drain logic. When not used, they should be left open, or connected to VDD.

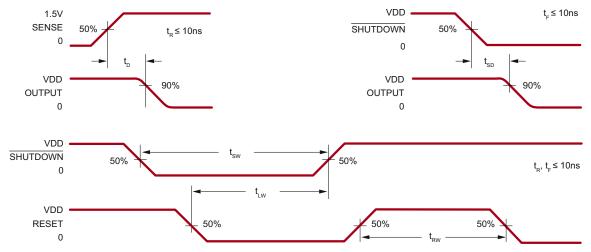
#### **Output Buffer**

The output buffer of the HV9123 is of standard CMOS construction (P-channel pull-up, N-channel pull-down). Thus the body-drain diodes of the output stage can be used for spike clipping if necessary, and external Schottky diode clamping of the output is not required.

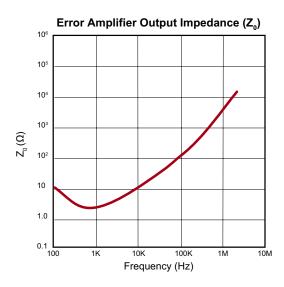
#### **Truth Table**

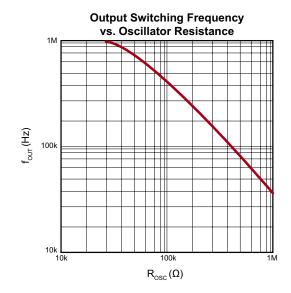
SHUTDOWN	RESET	Output
Н	Н	Normal operation
Н	$H \rightarrow L$	Normal operation, no change
L	Н	Off, not latched
L	L	Off, latched
$L \rightarrow H$	L	Off, latched, no change

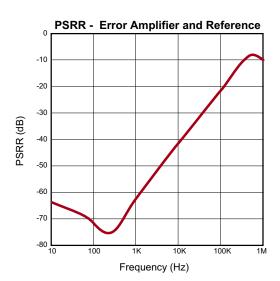
### **Shutdown Timing Waveforms**

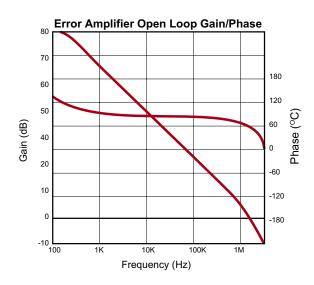


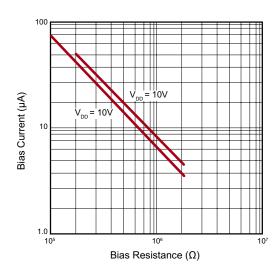
### **Typical Performance Curves**

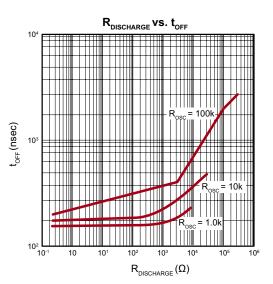












### **Pin Descriptions**

### 16-Lead SOIC (NG)

Pin #	Description				
1	+VIN				
-	-				
-	-				
4	SENSE				
5	OUTPUT				
6	-VIN				
7	VDD				
8	OSC OUT				

Pin#	Description					
9	OSC IN					
10	DISCHARGE					
11	VREF					
12	SHUTDOWN					
13	RESET					
14	COMP					
15	FB					
16	BIAS					

### 16-Lead PDIP (P)

10-Ecad i Dii (i )							
Pin #	Description						
1	+VIN						
2	NC						
3	NC						
4	SENSE						
5	OUTPUT						
6	-VIN						
7	VDD						
8	OSC OUT						

Pin #	Description
9	OSC IN
10	DISCHARGE
11	VREF
12	SHUTDOWN
13	RESET
14	COMP
15	FB
16	BIAS

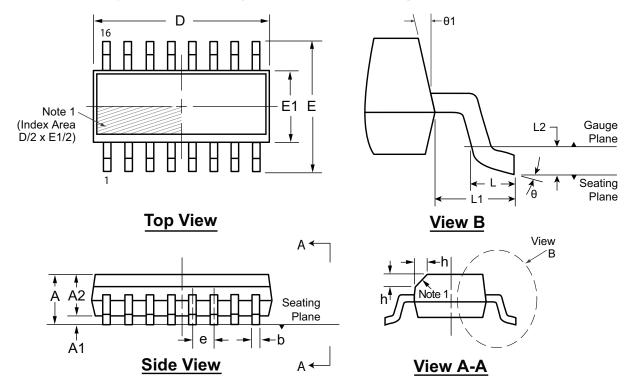
### 20-Lead PLCC (PJ)

Pin #	Description					
1	NC					
2	NC					
3	+VIN					
4	NC					
5	SENSE					
6	OUTPUT					
7	NC					
8	-VIN					
9	VDD					
10	OSC OUT					

Pin#	Description					
11	OSC IN					
12	DISCHARGE					
13	NC					
14	VREF					
15	NC					
16	SHUTDOWN					
17	RESET					
18	COMP					
19	FB					
20	BIAS					

# 16-Lead SOIC (Narrow Body) Package Outline (NG)

9.90x3.90mm body, 1.75mm height (max), 1.27mm pitch



#### Note:

1. This chamfer feature is optional. If it is not present, then a Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

Symbo	ol	Α	<b>A</b> 1	A2	b	D	E	E1	е	h	L	L1	L2	θ	θ1
	MIN	1.35*	0.10	1.25	0.31	9.80*	5.80*	3.80*		0.25	0.40			<b>0</b> °	5°
Dimension (mm)	NOM	-	-	-	-	9.90	6.00	3.90	1.27 - BSC -	-	-	1.04   0.25 REF   BSC	-	-	
(111111)	MAX	1.75	0.25	1.65*	0.51	10.00*	6.20*	4.00*		0.50	1.27			<b>8</b> º	15°

JEDEC Registration MS-012, Variation AC, Issue E, Sept. 2005.

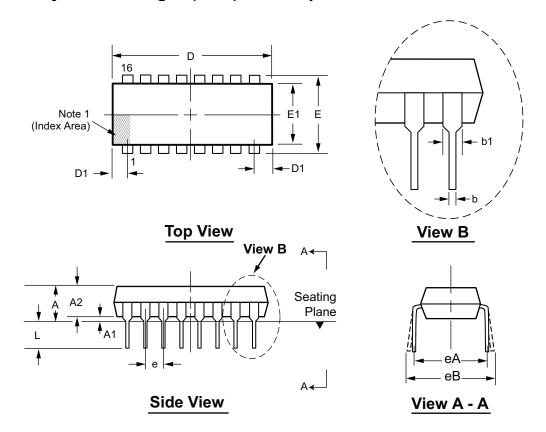
\* This dimension is not specified in the JEDEC drawing.

Drawings are not to scale.

Supertex Doc. #: DSPD-16SONG, Version G041309.

# 16-Lead PDIP (.300in Row Spacing) Package Outline (P)

.790x.250in body, .210in height (max), .100in pitch



#### Note:

A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

Symbol		Α	A1	A2	b	b1	D	D1	E	E1	е	eА	еВ	L
Dimension (inches)	MIN	.130*	.015	.115	.014	.045	.780	.005	.290 <sup>†</sup>	.240	.100 BSC	.300 BSC	.300*	.115
	NOM	-	-	.130	.018	.060	.790	-	.310	.250			-	.130
	MAX	.210	.035*	.195	.023 <sup>†</sup>	.070	.810 <sup>†</sup>	.050*	.325	.280			.430	.150

JEDEC Registration MS-001, Variation AB, Issue D, June, 1993.

\* This dimension is not specified in the JEDEC drawing.

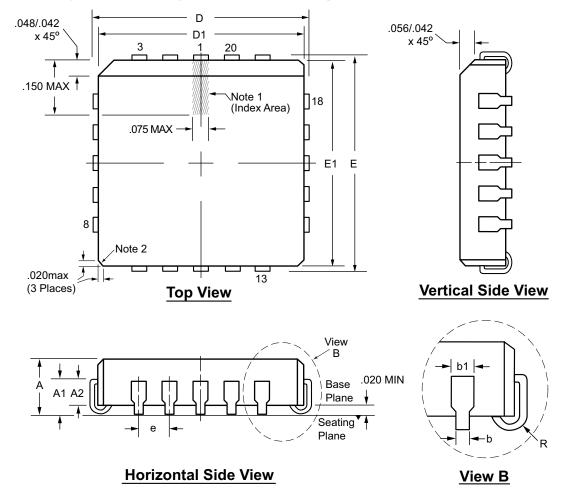
Drawings not to scale.

Supertex Doc. #: DSPD-16DIPP, Version B041009.

<sup>†</sup> This dimension differs from the JEDEC drawing.

### 20-Lead PLCC Package Outline (PJ)

.353x.353in body, .180in height (max), .050in pitch



#### Notes:

- A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.
- Actual shape of this feature may vary.

Symbol		Α	A1	A2	b	b1	D	D1	E	E1	е	R
Dimension (inches)	MIN	.165	.090	.062	.013	.026	.385	.350	.385	.350		.025
	NOM	.172	.105	-	-	-	.390	.353	.390	.353	.050 BSC	.035
	MAX	.180	.120	.083	.021	.032	.395	.356	.395	.356		.045

JEDEC Registration MS-018, Variation AA, Issue A, June, 1993.

Drawings not to scale.

Supertex Doc. #: DSPD-20PLCCPJ, Version C031111

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <a href="http://www.supertex.com/packaging.html">http://www.supertex.com/packaging.html</a>.)

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