

LND TRM33

LOW POWER 300 TO 500 MHz ASK TRANSMITTER

General Description

The LND TRM33 is a low power Amplitude Shift Keying (ASK) transmitter designed to operate at transmit frequencies between 300 and 500MHz. It is comprised of a frequency synthesizer and modulator, which accepts a simple CMOS digital data input and generates an ASK modulated output. Output power can be adjusted to a maximum of 2dBm. The LND TRM33 has been optimized for use in batterypowered applications with low current consumption in both transmit and standby modes.

How Does it Work

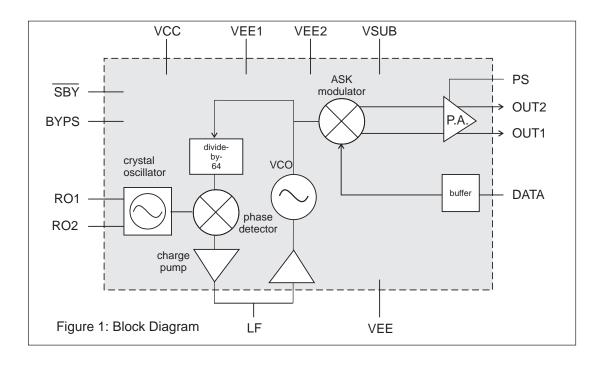
Simply connect the supply (as low as 2.6V), an external oscillator to RO1 and RO2, and all ground connections. The chip converts logic level inputs to balanced, complimentary, amplitude modulated outputs, Out1 and Out 2. Output power can be adjusted using resistor values in Table 5, and can be easily switched with logic input to Standby Mode (extremely low power) when not transmitting. Component selections, network, and antenna options are included in this document.

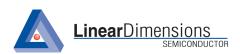
Applications

- Wireless office and home security systems
- Automotive remote locking/starting
- Consumer remote control units
- Radio data transmission
- Low power telemetry

Features

- Low current consumption in transmit mode
- Low standby current (typically a few nA)
- Conforms to requirements of ETSI-300-220, PT1340 and FCC Part 15
- Easy to use with low external component count and cost
- No discrete external inductors required
- Adjustable output power level





Theory of operation

The LND TRM33 comprises three main elements:

- 1 Input buffer and modulator
- 2 Frequency synthesizer for generation of the RF output carrier from a low frequency external crystal
- 3 Power amplifier (PA) for driving an external antenna

The IC accepts a digital CMOS input, and modulates a fixed RF carrier frequency on or off according to the digital input level. The RF carrier is generated by the on-chip phase locked loop (PLL) frequency synthesizer. This contains a fixed divider which ensures that the RF output frequency is 64 times the crystal reference

frequency. The power amplifier provides a variable output level of up to 2 dBm of output power depending upon the value of an external resistor R1.

The PA has balanced outputs OUT1 and OUT2 which are in open collector configuration. A balanced output is used since it provides a high degree of suppression of even-order harmonics of the fundamental. Odd order harmonic suppression is achieved using on-chip cancellation techniques.

Electrical characteristics

Table 1: Absolute maximum ratings					
Parameter	Conditions	Symbol	Min	Max	Unit
Supply voltage		VCC	-0.3	+7.0	V
Input voltage - Logic Inputs	DATA & SBY pins		-0.3V	VCC + 0.3V	
Input Current	DATA & SBY pins		-1.0	+1.0	mA
Storage temperature			-40	150	°C
Junction temperature			-40	150	°C
Power Dissipation (SO16)				200	mW

Table 2: Operating conditions @ 434 MHz; @ 318 MHz					
Parameter	Symbol	Min	Max	Unit	
Supply voltage	vcc	2.6	4.8	V	
Ambient temperature	Та	-40	65	°C	

Note

All specification parameters guaranteed only with the following components (please refer to figure 2 for application circuit):

RF transformer Mini-Circuits TC4-14

Crystal Euroquartz B537 6.78-MHz crystal (Cload = 10 pF)

R1 = 3.9 k R2 = 3.3 k C7 = 1 nF C8 = 47 pF C1 = C2 = 22 pF

Parameter	Conditions	Symbol	Min	Тур	Max	Unit
Standby Current	SBY = low	Iccstb		0.1	1	μΑ
Supply Current	DATA = low	Icclow			16.5	mA
Supply Current	DATA = high, including open-collector output currents	Icchigh			25	mA
DATA logic high level	guaranteed by design	Vhigh	0.7*VCC		VCC+0.3V	
DATA logic low level	guaranteed by design	Vlow	-0.3V		0.3*VCC	
DATA input current	guaranteed by design -0.3 < Vin < Vcc+0.3	Idata	-10		+10	μA
Bias voltage on pin PS	guaranteed by design I(PS) = 100 μA	V(PS)		0.4		V



Electrical characteristics (continued)

Table 4: AC Characteristics at 3.0V VCC and 23°C ambient temperature, fc = 433.92 MHz, unless otherwise stated all parameters 100 % production tested under these conditions unless otherwise stated Conditions Min Max Unit Parameter Symbol Тур Maximum Carrier Frequency 440 MHz fcmax Minimum Carrier Frequency fcmin 290 MHz Maximum Data Rate received BER<1% with receiver Rmfsk 200 kbits/s pre-detection bandwidth matched to data rate R=1.5 kbit/s 50 dΒ **Extinction Ratio** Μ Reference Frequency guaranteed by design, 4.5 7.0 MHz fref external crystal with R1=3.9 kOhms Output Power Po 0.5 dBm 2nd Harmonic P2 -34 dBc 3rd Harmonic P3 -35 -28 dBc Reference Spurious at fc + (fref to 3fref) Pref -45 -40 dBc Phase Noise -75 dBc/Hz guaranteed by design, PΝ at 10 kHz offset VCO gain guaranteed by design **KVCO** 125 MHz/V Phase detector gain guaranteed by design KPD 16 µA/rad

Matching network and antenna options

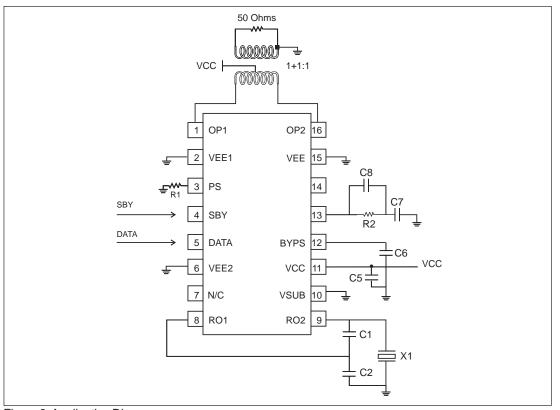


Figure 2: Application Diagram

Figure 2 shows a typical application diagram.

In order to minimise second harmonic distortion, a differential output configuration is recommended.



Matching network and antenna options (continued)

The LND TRM33 needs the following:

1) a dc bias reference to VCC (for instance through use of printed bias inductors with a connection to 1) the IC can provide 7.5mA of current into the load VCC, or direct connection of a loop antenna connected to VCC at its centre). Linear's TRM3x series of evaluation boards provide large and small loop antenna solutions for customers to copy. These boards require a minimum number of external components, and importantly no discrete inductors are required. For driving an external single-ended antenna, a high-efficiency printed balun is also available.

Table 5: Output Power Settings				
R (kOhms)	Output power (dBm)	Current cons. (mA)		
2.8	2	25		
3.8	0	22		
5.2	-2	20		
6.7	-4	19		
8.5	-6	18		
9.7	-8	17		
10.3	-10	16		
10.7	-12	16		

For optimum operation, the following points are highlighted:

- 2) for maximum output power, it is recommended that the antenna load be transformed to an impedance of 2000hms
- 3) a PLL synthesizer is designed to operate with an external second order loop filter as shown in figure 2, with component values as shown in Table 6. These values determine the loop bandwidth and dynamics and operation of the loop is not guaranteed for any other values
- 4) the bias current for the power amplifier directly controls the output current and hence the O/P power. This bias current is set by an external resistor connected between PS and ground. Output power versus R1 values is shown in Table 5, for a differential load impedance of 200 Ω .

Table 6: Recommended External Components					
Component	Function	Value	Tolerance	Units	
C1	Crystal Oscillator	22	± 5%	pF	
C2	Crystal Oscillator	22	± 5%	pF	
C5	Supply Decoupling	100	± 20%	nF	
C6	Regulator Decoupling	100	± 20%	nF	
C7	PLL Loop Filter	1	± 5%	nF	
C8	PLL Loop Filter	47	± 5%	pF	
R1	O/P power set	3.9	± 5%	kOhms	
R2	PLL Loop Filter	3.3	± 5%	kOhms	
X1	Crystal	6.78	< 20ppm	MHz	
Note: Recommended Crystal Oscillator type: B537					

Table 7:	Table 7: I/O interfacing				
No.	Signal	Description			
1	OP1	PA Output			
2	VEE1	Ground (0V)			
3	PS	Power Set			
4	SBY	Standby (complete power down)			
5	DATA	Data			
6	VEE2	Ground (0V)			
7	N/C	Not connected			
8	RO1	Reference Oscillator (emitter)			
9	RO2	Reference Oscillator (base)			
10	VSUB	Substrate Connection (0V)			
11	VCC	Supply (Battery)			
12	BYPS	Power supply decoupling to ground			
13	LF	Low Pass Filter			
15	VEE	Ground (0V)			
16	OP2	PA Output (complement)			

Input and output interface diagrams

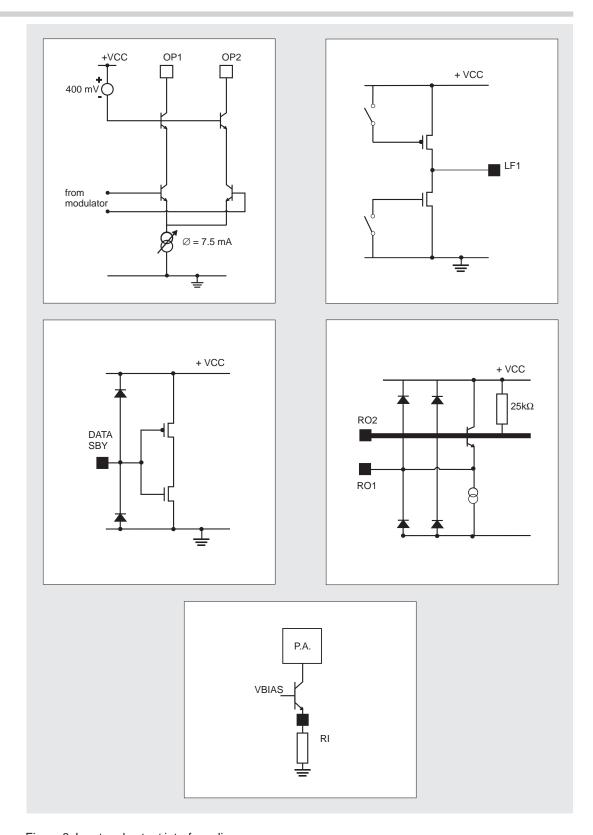
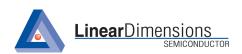
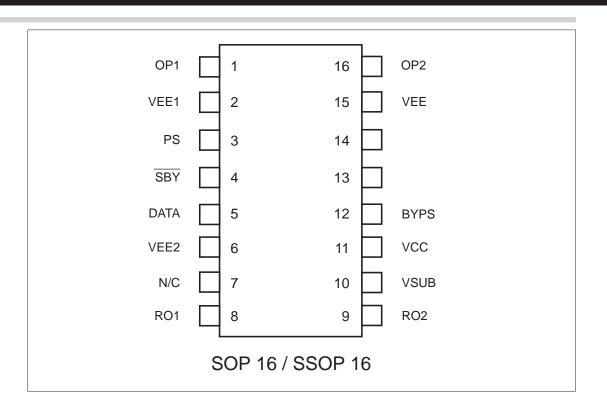


Figure 3: Input and output interface diagrams Pinout Information





Package Information

