



### **General Description**

The MAX11508/MAX11509 integrated 3-channel video filters for high-definition (HD), progressive-scan (PS), standard-definition (SD), and bypass (BP) video include an output buffer with +6dB gain and are ideal for digital video disc (DVD) players, set-top box (STB) receivers, high-definition television (HDTV), digital video recorders (DVRs), and similar devices.

The MAX11508/MAX11509 video inputs feature a transparent clamp compatible with AC- and DC-coupled input signals and allow DAC outputs to be directly coupled, eliminating the need for bulky coupling capacitors. The filter bandwidths are selectable to HD, PS, and SD. A BP mode is provided for 1080p and highbandwidth RGB signals. Selectable input bias circuitry on 2 filter channels offers simple connection to bipolar video signals such as C, Pb, and Pr.

The MAX11508 filters provide -3dB bandwidths of 9MHz (SD), 16MHz (PS), 33MHz (HD), and 60MHz (BP). The MAX11509 filters provide -3dB bandwidths of 10MHz (SD), 17MHz (PS), 34MHz (HD), and 60MHz (BP).

Each channel includes an output buffer with +6dB gain that provides a full  $2V_{P-P}$  video signal into a  $150\Omega$  video load. The buffers drive either AC- or DC-coupled loads and assure a blanking level of below 1V after the backmatch resistor. The shutdown mode provided reduces device current to 1µA (typ).

The MAX11508 offers a flat passband and the MAX11509 features +0.8dB peaking to compensate for DAC rolloff. The MAX11508/MAX11509 operate from a 5V power supply and operate over the upper commercial (0°C to +85°C) temperature range. The parts are offered in the 14-pin TSSOP package.

### **Applications**

Cable and Satellite STB Receivers

**HDTV** 

**DVD Players** 

Personal Video Recorders

**DVRs** 

Video-On-Demand

**HD** Recorders

Typical Operating Circuit appears at end of data sheet.

#### **Features**

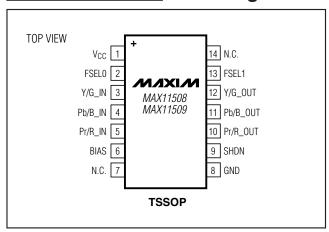
- ♦ Three Selectable 6th-Order 9MHz/16MHz/33MHz (SD/PS/HD) Filters
- **♦** Bypass Mode for High Bandwidth Signals
- **♦ Transparent Input Clamp**
- ♦ Output Buffers Drive a Standard 150Ω Video Load
- ♦ ±12kV HBM ESD Protection on Outputs
- **♦** AC- or DC-Coupled Inputs
- **♦ AC- or DC-Coupled Outputs**
- ♦ +0.8dB Peaking Passband Response (MAX11509) in SD, PS, and HD Modes
- ♦ Single +5V Power Supply
- ♦ Lead-Free, 14-Pin TSSOP Package

### **Ordering Information**

PART	TEMP RANGE	PIN-PACKAGE		
MAX11508UUD+	0°C to +85°C	14 TSSOP		
MAX11509UUD+	0°C to +85°C	14 TSSOP		

<sup>+</sup>Denotes a lead-free/RoHS-compliant package.

#### Pin Configuration



#### **ABSOLUTE MAXIMUM RATINGS**

V <sub>CC</sub> to GND0.3V to +6V All Other Pins to GND0.3V to the lower of (V <sub>CC</sub> + 0.3V) and +6V	Operating Temperature Range0°C to +85°C Storage Temperature Range65°C to +150°C Lead Temperature (soldering, 10s)+300°C
Continuous Power Dissipation (T <sub>A</sub> = +70°C)	Junction Temperature+150°C
14-Pin TSSOP (derate 10mW/°C above +70°C)796.8mW	
Maximum Current into Any Pin Except VCC and GND±50mA	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **ELECTRICAL CHARACTERISTICS**

 $(V_{CC} = +5V, R_{LOAD} = 150\Omega)$  to GND,  $C_{IN} = 0.1 \mu F$ ,  $T_A = 0^{\circ}C$  to  $+85^{\circ}C$ , all frequency responses are relative to 100kHz, unless otherwise noted.) (Note 1)

PARAMETER SYMBOL C		CONDITIONS	MIN	TYP	MAX	UNITS	
DC CHARACTERISTICS	•		•				
Supply Current	Icc	No load		24	35	mA	
Supply Voltage Range	V <sub>DD</sub>		4.75	5.00	5.25	V	
Input Voltage Range	VIN	Referenced to GND if DC-coupled		1.4		V	
Power-Down Current	lee	BIAS = high		1		μА	
Power-Down Current	I <sub>PD</sub>	BIAS = low		15			
Digital Input High Voltage	VIH		2.0			V	
Digital Input Low Voltage	VIL				0.8	V	
STANDARD-DEFINITION VID	EO						
1dD Dandwidth	f	MAX11508, T <sub>A</sub> = +25°C	4.5	7.9		NAL I—	
-1dB Bandwidth	f <sub>1dB</sub>	MAX11509, $T_A = +25^{\circ}C$	5	8.6		MHz	
0.10.0		MAX11508		9.0		MHz	
-3dB Bandwidth	f <sub>3dB</sub>	MAX11509	10		IVI□Z		
Stopband Attenuation	AsB	f = 27MHz	35	48		dB	
Low-Frequency Gain	Av	No load	5.6	6.0	6.6	dB	
Differential Gain	dG	All channels		0.7		%	
Differential Phase dø		All channels		0.7		Degrees	
Total Harmonic Distortion	THD	V <sub>OUT</sub> = 1.4V <sub>P-P</sub> , f = 1MHz, all channels		-60		dB	
Signal-to-Noise Ratio SNR		2V <sub>P-P</sub> signal to RMS noise, f = 100kHz to 4.2MHz		71		dB	
Group Delay	tG	f = 4.5MHz		59		ns	
Power-Supply Rejection Ratio	PSRR	DC, all channels		50		dB	

## **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{CC} = +5V, R_{LOAD} = 150\Omega)$  to GND,  $C_{IN} = 0.1 \mu F$ ,  $T_A = 0^{\circ}C$  to  $+85^{\circ}C$ , all frequency responses are relative to 100kHz, unless otherwise noted.) (Note 1)

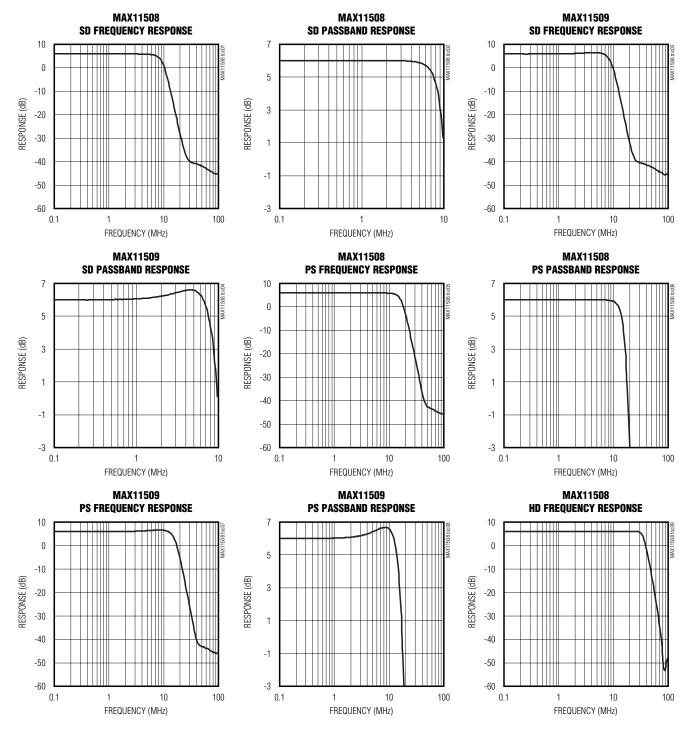
PARAMETER SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS	
PROGRESSIVE-SCAN VIDE	0		·				
1 dD. Donaduui dtla	f	MAX11508, T <sub>A</sub> = +25°C	9	13.5		NAL I-	
-1dB Bandwidth	f <sub>1dB</sub>	MAX11509, T <sub>A</sub> = +25°C	10	14		MHz	
OdD Doody idth	f	MAX11508		16		MHz	
-3dB Bandwidth	f <sub>3dB</sub>	MAX11509		17			
Stopband Attenuation	Asa	f = 54MHz		44		dB	
Low-Frequency Gain	Ay		5.6	6.0	6.6	dB	
Total Harmonic Distortion	THD	$V_{OUT} = 1.4V_{P-P}, f = 7MHz$		-55		dB	
Signal-to-Noise Ratio	SNR	2V <sub>P-P</sub> signal to RMS noise, f = 100kHz to 15MHz		66		dB	
Group Delay	tg	f = 10MHz		47		ns	
HIGH-DEFINITION VIDEO		<u>'</u>					
4 ID D		MAX11508, T <sub>A</sub> = +25°C	21	30		N 41 1	
-1dB Bandwidth	f <sub>1dB</sub>	MAX11509, T <sub>A</sub> = +25°C	22	31		MHz	
0.10.0		MAX11508		33		MHz	
-3dB Bandwidth	f <sub>3dB</sub>	MAX11509		34			
		f = 37.125MHz		6.5		dB	
Stopband Attenuation	ASB	f = 44.25MHz		14.5			
		f = 74.25MHz		45			
Low-Frequency Gain	Ay		5.4	6.0	6.6	dB	
		$V_{OUT} = 1.4V_{P-P}$ , $f = 10MHz$		-55		dB	
Total Harmonic Distortion	THD	$V_{OUT} = 1.4V_{P-P}, f = 15MHz$		-50			
		$V_{OUT} = 1.4V_{P-P}, f = 22MHz$		-40			
Signal-to-Noise Ratio SNR		2V <sub>P-P</sub> signal to RMS noise, f = 100kHz to 30MHz		65		dB	
Group Delay	tg	f = 20MHz		25		ns	
BYPASS VIDEO (Note 2)		·	•				
-3dB Bandwidth	f <sub>3dB</sub>			60		MHz	
Low-Frequency Gain	Av		5.4	6.0	6.6	dB	
Total Harmonic Distortion	THD	V <sub>OUT</sub> = 1.4V <sub>P-P</sub> , f = 22MHz	-40			dB	
Signal-to-Noise Ratio	SNR	2V <sub>P-P</sub> signal to RMS noise, f = 100kHz to 30MHz		65		dB	

**Note 1:** All devices are 100% production tested at  $T_A = +25$ °C.

Note 2: Output AC-coupled with 220 $\mu F$  into 150 $\Omega$  to GND.

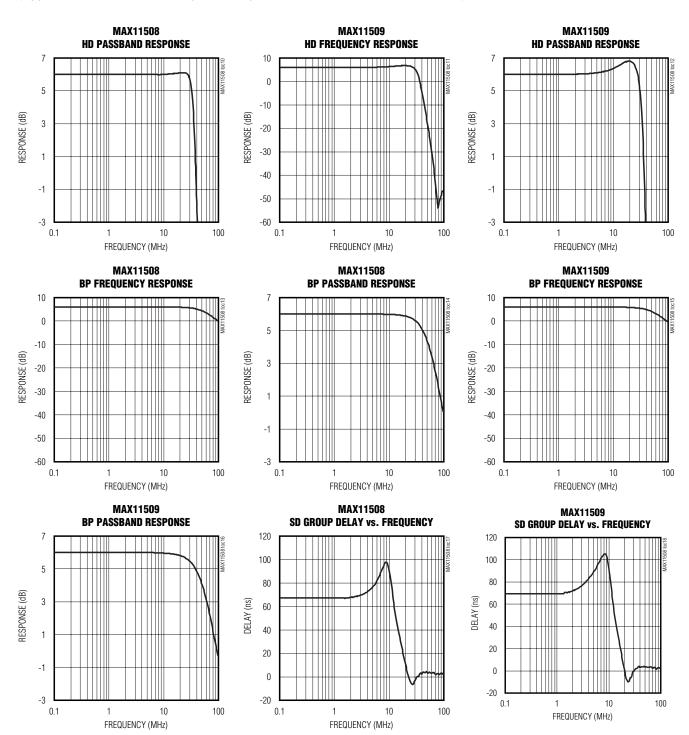
# Typical Operating Characteristics

 $(V_{CC} = 5V, R_L = 150\Omega)$  to GND, output DC-coupled,  $T_A = +25$ °C, unless otherwise noted.)



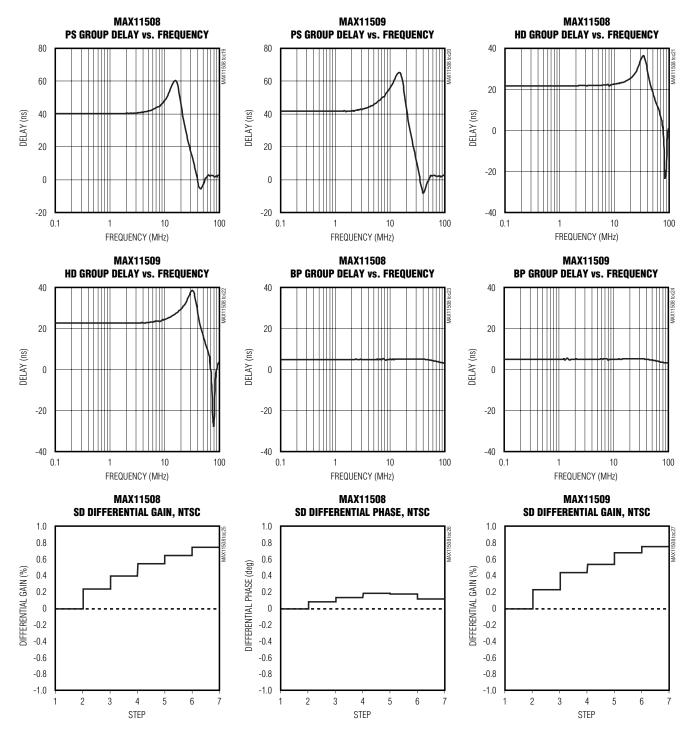
## Typical Operating Characteristics (continued)

 $(V_{CC} = 5V, R_L = 150\Omega \text{ to GND, output DC-coupled, } T_A = +25^{\circ}\text{C, unless otherwise noted.})$ 



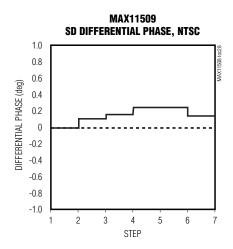
## \_Typical Operating Characteristics (continued)

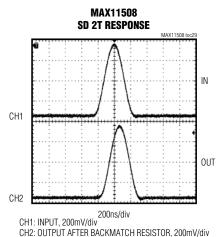
 $(V_{CC} = 5V, R_L = 150\Omega)$  to GND, output DC-coupled,  $T_A = +25$ °C, unless otherwise noted.)

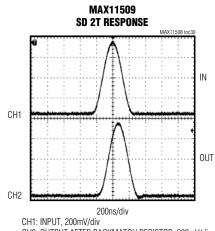


## Typical Operating Characteristics (continued)

 $(V_{CC} = 5V, R_L = 150\Omega)$  to GND, output DC-coupled,  $T_A = +25$ °C, unless otherwise noted.)







CH2: OUTPUT AFTER BACKMATCH RESISTOR, 200mV/div

# **Pin Description**

PIN	NAME	FUNCTION		
1	Vcc	ower-Supply Input		
2	FSEL0	requency-Select Input. LSB for bandwidth selection.		
3	Y/G_IN	Video Input Channel 1		
4	Pb/B_IN	Video Input Channel 2		
5	Pr/R_IN	Video Input Channel 3		
6	BIAS	Bias Control Digital Input. Enables voltage bias for the Pb/B and Pr/R channels. Force BIAS low to enable voltage bias and disable the input clamps. Force BIAS high to disable voltage bias and enable input clamps.		
7, 14	N.C.	No Connection. Not internally connected.		
8	GND	Ground		
9	SHDN	Shutdown Digital Input. SHDN places the device into a powered-down state. Force SHDN low to enable shutdown. Force SHDN high for normal operation.		
10	Pr/R_OUT	Video Output Channel 3		
11	Pb/B_OUT	Video Output Channel 2		
12	Y/G_OUT	Video Output Channel 1		
13	FSEL1	Frequency Select Input. MSB for bandwidth selection.		

## **Detailed Description**

The MAX11508/MAX11509 integrated filters offer 3 channels of SD, PS, or HD video, and include a bypass mode useful for high-bandwidth 1080p and RGB video signals. The MAX11508 6th-order lowpass filters provide -3dB bandwidths of 9MHz (SD), 16MHz (PS), 33MHz (HD), and 60MHz (BP). The MAX11509 6th-order lowpass filters provide -3dB bandwidths of 10MHz (SD), 17MHz (PS), 34MHz (HD), and 60MHz (BP).

Two control inputs, FSEL0 and FSEL1, select the filter mode for all filters (Table 1). Input bias circuitry on 2 filter channels can be enabled through the BIAS input to offer simple connection to bipolar video signals such as C, Pb, and Pr. Shutdown mode reduces device current to  $1\mu A$  (typ) and is enabled by forcing SHDN low. Figure 1 shows a simplified block diagram of the MAX11508/MAX11509.

The MAX11508 provides a flat passband response and the MAX11509 provides a +0.8dB high-frequency boost at 5MHz (SD), 8.5MHz (PS), and 20MHz (HD) to help with system rolloff. No frequency boost is included in bypass mode. Typical voltage waveforms are shown in Figures 2 and 3.

### **Inputs**

#### Transparent Clamps

All inputs feature transparent clamps to allow either AC-or DC-coupling of the inputs. The clamp remains inactive while the input signal is above ground, offering true DC input coupling. When the signal goes below ground, as is the case when it is AC-coupled, the clamp sets the sync tip close to the ground level.

#### **Input Coupling**

The choice of AC- or DC-coupling the input depends on the video source. Many DACs have a current output and are terminated to ground with a resistor; such signals are conveniently DC-coupled. Use AC-coupling when the DC level of the video signal is unknown or outside the specified input range of the MAX11508/MAX11509, such as SCART or VCC-terminated DAC outputs. A bias network within 2 channels of the MAX11508/MAX11509, for use with bipolar signals,

connects to the input node when BIAS is forced low. Figures 4 and 5 show how the bias network operates.

#### **DC-Coupled Inputs**

When the input is DC-coupled, the voltage must remain above zero, but not to exceed 1.4V (typ).

#### **AC-Coupling and BIAS**

When the input is AC-coupled, the transparent clamps are active and set the lowest point of the signal at ground. This is appropriate for unipolar signals such as CVBS, Y, R, G, or B, with sync pulses (Figure 4). Force BIAS high when coupling unipolar signals.

For bipolar signals, such as Pb and Pr, a bias network is provided within the MAX11508/MAX11509 for inputs Pb/B\_IN and Pr/R\_IN. Force BIAS low to connect the bias network to these inputs (Figure 5). The internal network biases AC-coupled inputs to a fixed DC voltage, typically 0.59V, to ensure that the transparent clamp remains off.

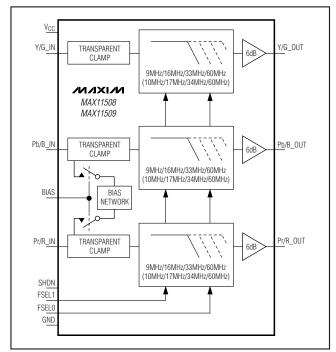


Figure 1. Block Diagram

**Table 1. Frequency-Selection Truth Table** 

CONTRO	CONTROL INPUTS		EQUENCY (MHz)	OPERATING MODE	
FSEL1	FSEL0	MAX11508	MAX11509	OPERATING MODE	
0	0	9	10	Standard definition	
0	1	16	17	Progressive scan	
1	0	33	34	High definition	
1	1	60	60	Bypass	

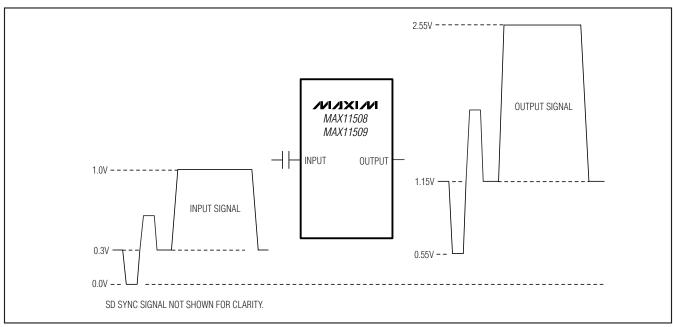


Figure 2. Typical AC-Coupled Signal

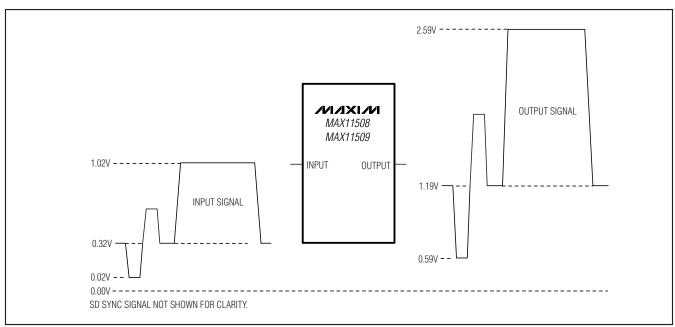


Figure 3. Typical DC-Coupled Signal

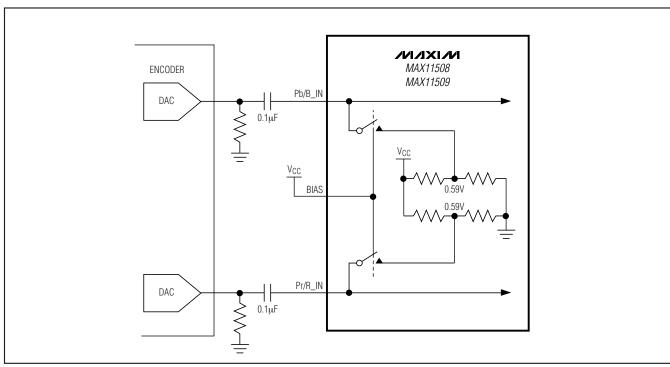


Figure 4. Simple AC-Coupling and BIAS Control for Unipolar Signals (CVBS, Y, R, G, B)

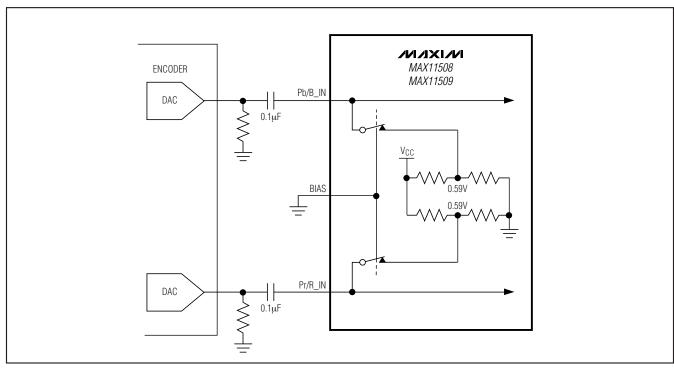


Figure 5. AC-Coupling and BIAS Control for Bipolar Signals (C, Pb, Pr)

#### **Filters**

The MAX11508 filter characteristic provides excellent time-domain response with low overshoot and guarantees minimal attenuation in the passband. The MAX11509 filters offer a small gain peaking response to counter system rolloff.

Select filter frequency with inputs FSEL0 and FSEL1, as shown in Table 1.

#### Standard-Definition (SD) Filters

The MAX11508 SD filters have a 9MHz (typ) -3dB frequency, while the MAX11509 SD filters offer a 10MHz (typ) -3dB frequency and a +0.8dB high-frequency boost at 5MHz (typ). Both devices have a stopband attenuation of +48dB (typ) at 27MHz.

#### Progressive-Scan (PS) Filters

The MAX11508 PS filters have a 16MHz (typ) -3dB frequency, while the MAX11509 PS filters offer a 17MHz (typ) -3dB frequency and a +0.8dB high-frequency boost at 8.5MHz (typ). Both devices have a stopband attenuation of +44dB (typ) at 54MHz.

#### **High-Definition (HD) Filters**

The MAX11508 HD filters have a 33MHz (typ) -3dB frequency, while the MAX11509 HD filters offer a 34MHz (typ) -3dB frequency and a +0.8dB high-frequency boost at 20MHz (typ). Both devices have a stopband attenuation of +45dB (typ) at 74.25MHz.

#### **Bypassing the Filters**

The MAX11508/MAX11509 filter bypass circuitry offers a 60MHz (typ) -3dB frequency. Bypassed filters offer no gain peaking.

#### **Output Buffer**

The MAX11508/MAX11509 feature output buffers with +6dB gain that drive a standard 150 $\Omega$  video load at 2V<sub>P-P</sub>. A typical load consists of a 75 $\Omega$  backmatch resistor, an optional 220 $\mu$ F or larger coupling capacitor, and a 75 $\Omega$  termination resistor.

An offset of 550mV is added at the output. The offset ensures that the blanking level on the output is less than 1V after the backmatch resistor, thus meeting digital TV specifications allowing the devices to drive video loads directly without using costly AC-coupling capacitors. The basic output voltage equation of all filters is:

 $V_{OUT} = (2 \times V_{IN}) + 0.55V$ 

Typical voltage waveforms are shown in Figures 2 and 3.

#### Shutdown

Forcing digital input SHDN low places the MAX11508/ MAX11509 into low-power shutdown mode. In shutdown, the device consumes only 1µA (typ), and the outputs are internally connected to GND through  $1.58k\Omega$  resistors. In shutdown, the input clamps are disabled and the inputs are internally connected to GND through  $350k\Omega$  resistors. When shutdown is forced low while BIAS is low, the bias network remains active, but the bias voltage changes from its nominal 0.59V to 0.475V (typ).

## **Applications Information**

#### **Output Considerations**

DC- or AC-couple the MAX11508/MAX11509 outputs. These devices, with +6dB gain, are typically connected to a 75 $\Omega$  series backmatch resistor followed by a video cable. Choose an AC-coupling capacitor value that ensures that the lowest frequency content in the video signal is passed and the field-time distortion is kept within desired limits when using an AC connection. The selection of this value is a function of the input impedance, and more importantly, the input leakage of the circuit being driven. Common industry practice is to use a 220µF or larger capacitor.

The MAX11508/MAX11509 outputs are fully protected against short circuits to ground. The short-circuit protection circuitry limits the output current to 75mA (typ) per output. Shorting more than one output to ground simultaneously may exceed the maximum package power dissipation.

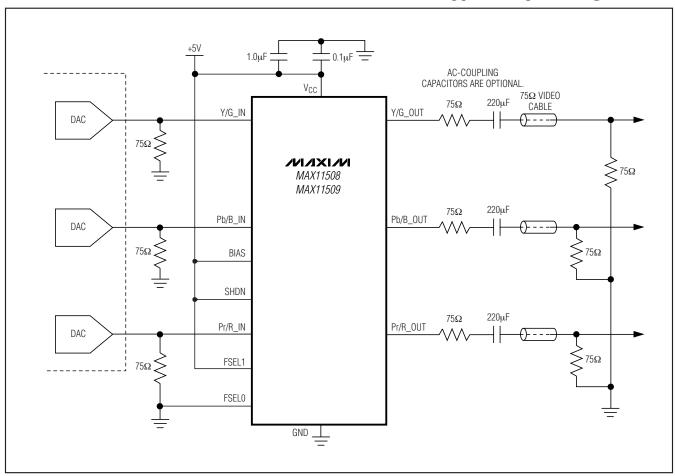
#### **PCB Layout Recommendations**

Connect the power and ground traces to large copper areas to enhance power dissipation. Bypass  $V_{CC}$  to GND with  $0.1\mu F$  and  $1.0\mu F$  capacitors. Place the  $0.1\mu F$  capacitor closest to  $V_{CC}$ . Use surface-mount capacitors for their low inductance. Place traces carrying video signals appropriately to avoid mutual coupling. When inputs are AC-coupled, place the capacitors as close as possible to the device and keep traces short to minimize parasitic capacitance and inductance. Refer to the MAX11508/MAX11509 evaluation kit data sheet for PCB layout.

**Chip Information** 

PROCESS: BICMOS

### **Typical Operating Circuit**



### Package Information

For the latest package outline information and land patterns, go to <a href="https://www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
14 TSSOP	U14+1	<u>21-0066</u>

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