# 2-Wire Serial 8-Bit DACs with Rail-to-Rail Outputs 

> General Description

The MAX517/MAX518/MAX519 are 8-bit voltage output digital-to-analog converters (DACs) with a simple 2-wire serial interface that allows communication between multiple devices. They operate from a single 5 V supply and their internal precision buffers allow the DAC outputs to swing rail-to-rail.
The MAX517 is a single DAC and the MAX518/MAX519 are dual DACs. The MAX518 uses the supply voltage as the reference for both DACs. The MAX517 has a reference input for its single DAC and each of the MAX519's two DACs has its own reference input.
The MAX517/MAX518/MAX519 feature a serial interface and internal software protocol, allowing communication at data rates up to 400 kbps . The interface, combined with the double-buffered input configuration, allows the DAC registers of the dual devices to be updated individually or simultaneously. In addition, the devices can be put into a low-power shutdown mode that reduces supply current to $4 \mu \mathrm{~A}$. Power-on reset ensures the DAC outputs are at 0 V when power is initially applied.
The MAX517/MAX518 are available in space-saving 8pin DIP and SO packages. The MAX519 comes in 16pin DIP and SO packages.

Applications
Minimum Component Analog Systems
Digital Offset/Gain Adjustment
Industrial Process Control
Automatic Test Equipment
Programmable Attenuators

Pin Configurations


- Single +5V Supply
- Simple 2-Wire Serial Interface
- $I^{2} C$ Compatible
- Output Buffer Amplifiers Swing Rail-to-Rail
- Space-Saving 8-pin DIP/SO Packages (MAX517/MAX518)
- Reference Input Range Includes Both Supply Rails (MAX517/MAX519)
- Power-On Reset Clears All Latches
- 4 A A Power-Down Mode

Ordering Information

| PART | TEMP. RANGE | PIN-PACKAGE | TUE <br> (LSB) |
| :--- | :--- | :--- | :---: |
| MAX517ACPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Plastic DIP | 1 |
| MAX517BCPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Plastic DIP | 1.5 |
| MAX517ACSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 SO | 1 |
| MAX517BCSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 SO | 1.5 |
| MAX517BC/D | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | Dice $^{*}$ | 1.5 |

Ordering Information continued at end of data sheet.
*Dice are specified at $T_{A}=+25^{\circ} \mathrm{C}, D C$ parameters only.
**Contact factory for availability and processing to MIL-STD-883.
Functional Diagram


Maxim Integrated Products 1

## MAXIM

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Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

$\left(V_{D D}=5 \mathrm{~V} \pm 10 \%, V_{R E F}=4 \mathrm{~V}(M A X 517, M A X 519), R_{L}=10 k \Omega, C_{L}=100 \mathrm{pF}, T_{A}=T_{M I N}\right.$ to $T_{M A X}$, unless otherwise noted Typical values are $\mathrm{T}_{\mathrm{A}}={ }^{-}+25^{\circ} \mathrm{C}$.)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| STATIC ACCURACY |  |  |  |  |  |  |
| Resolution |  |  |  | 8 |  | Bits |
| Total Unadjusted Error (Note 1) | TUE |  | MAX51 _A |  | $\pm 1$ | LSB |
|  |  |  | MAX51 _B |  | $\pm 1.5$ |  |
| Differential Nonlinearity (Note 1) | DNL | Guaranteed monotonic |  |  | $\pm 1$ | LSB |
| Zero-Code Error | ZCE | Code $=00$ hex | MAX51 _C |  | 18 | mV |
|  |  |  | MAX51 _E |  | 20 |  |
|  |  |  | MAX51 BM |  | 20 |  |
| Zero-Code-Error Supply Rejection |  | Code $=00$ hex | MAX51 _C |  | $\pm 1$ | mV |
|  |  |  | MAX51 _E |  | $\pm 1$ |  |
|  |  |  | MAX51 _BM |  | $\pm 1$ |  |
| Zero-Code-Error Temperature Coefficient |  | Code = 00 hex |  |  | $\pm 10$ | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| Full-Scale Error |  | Code $=$ FF hex, MAX518 unloaded | MAX51 _C |  | $\pm 18$ | mV |
|  |  |  | MAX51 _E |  | $\pm 20$ |  |
|  |  |  | MAX51 _BM |  | $\pm 20$ |  |
| Full-Scale-Error Supply Rejection |  | MAX517, MAX519 <br> Code = FF hex <br> $V_{D D}=+5 \mathrm{~V} \pm 10 \%$ | MAX51 _C |  | $\pm 1$ | mV |
|  |  |  | MAX51 _E |  | $\pm 1$ |  |
|  |  |  | MAX51 _BM |  | $\pm 1$ |  |
| Full-Scale-Error Temperature Coefficient |  | Code = FF hex |  | $\pm 10$ |  | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |

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## ELECTRICAL CHARACTERISTICS (continued)

$\left(V D D=5 \mathrm{~V} \pm 10 \%, V_{R E F}=4 \mathrm{~V}(M A X 517, M A X 519), R L=10 k \Omega, C L=100 p F, T A=T M I N\right.$ to $T M A X$, unless otherwise noted. typical values are $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| REFERENCE INPUTS (MAX517, MAX519) |  |  |  |  |  |  |  |
| Input Voltage Range |  |  |  | 0 |  | VDD | V |
| Input Resistance | RIN | Code = 55 hex (Note 2) |  | 16 | 24 |  | $\mathrm{k} \Omega$ |
| Input Current |  | Power-down mode |  |  |  | $\pm 10$ | $\mu \mathrm{A}$ |
| Input Capacitance |  | Code = FF hex (Note 3) |  |  | 30 |  | pF |
| Channel-to-Channel Isolation (MAX519) |  | (Note 4) |  |  | -60 |  | dB |
| AC Feedthrough |  | (Note 5) |  |  | -70 |  | dB |
| DAC OUTPUTS |  |  |  |  |  |  |  |
| Full-Scale Output Voltage |  |  |  | 0 |  | VDD | V |
| Output Load Regulation |  | OUT_= $4 \mathrm{~V}, 0 \mathrm{~mA}$ to 2.5 mA |  |  | 0.25 |  |  |
|  |  | $\begin{aligned} & \text { MAX51 C/E, REF }=V_{D D} \\ & \left(\text { MAX } 517_{7}, \operatorname{MAX519}\right), \text { code }=\text { FF hex, } \\ & 0 \mu A \text { to } 500 \mu \mathrm{~A} \end{aligned}$ |  |  | 1.5 |  | LSB |
|  |  | $\begin{aligned} & \text { MAX51 M, REF_ = VDD } \\ & (\text { MAX517, MAX519), code }=\text { FF hex, } \\ & 0 \mu A \text { to } 500 \mu \mathrm{~A} \end{aligned}$ |  |  | 2.0 |  |  |
| Output Leakage Current |  | OUT_ = OV to VDD, power-down mode |  |  |  | $\pm 10$ | $\mu \mathrm{A}$ |
| DIGITAL INPUTS SCL, SDA |  |  |  |  |  |  |  |
| Input High Voltage | $\mathrm{V}_{\mathrm{IH}}$ |  |  | $0.7 \mathrm{~V}_{\text {DD }}$ |  |  | V |
| Input Low Voltage | VIL |  |  |  |  | $0.3 \mathrm{~V}_{\mathrm{DD}}$ | V |
| Input Leakage Current | IIN | O |  |  |  | $\pm 10$ | $\mu \mathrm{A}$ |
| Input Hysteresis | VHYST |  |  | 0.05 V DD |  |  | V |
| Input Capacitance | CIN | (Note 6) |  |  |  | 10 | pF |
| DIGITAL INPUTS AD0, AD1, AD2, AD3 |  |  |  |  |  |  |  |
| Input High Voltage | $\mathrm{V}_{\mathrm{IH}}$ |  |  | 2.4 |  |  | V |
| Input Low Voltage | $\mathrm{V}_{\text {IL }}$ |  |  |  |  | 0.8 | V |
| Input Leakage Current | IIN | $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$ to $\mathrm{V}_{\mathrm{DD}}$ |  |  |  | $\pm 10$ | $\mu \mathrm{A}$ |
| DIGITAL OUTPUT SDA (Note 7) |  |  |  |  |  |  |  |
| Output Low Voltage | VOL | ISINK $=3 \mathrm{~mA}$ |  |  |  | 0.4 | V |
|  |  | ISINK $=6 \mathrm{~mA}$ |  |  |  | 0.6 |  |
| Three-State Leakage Current | IL | VIN $=0 \mathrm{~V}$ to $\mathrm{V}_{\mathrm{DD}}$ |  |  |  | $\pm 10$ | $\mu \mathrm{A}$ |
| Three-State Output Capacitance | Cout | (Note 6) |  |  |  | 10 | pF |
| DYNAMIC PERFORMANCE |  |  |  |  |  |  |  |
| Voltage Output Slew Rate |  | Positive and negative | MAX51 _C |  | 2.0 |  | $\mathrm{V} / \mathrm{\mu s}$ |
|  |  |  | MAX51 _E |  | 1.4 |  |  |
|  |  |  | MAX51 _M |  | 1.0 |  |  |
| Output Settling Time |  | To $1 / 2 \mathrm{LSB}, 10 \mathrm{k} \Omega$ and 100pF load (Note 8) |  |  | 6 |  | $\mu \mathrm{s}$ |
| Digital Feedthrough |  | Code $=00$ hex, all digital inputs from OV to VDD |  |  | 5 |  | nV -s |

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## ELECTRICAL CHARACTERISTICS (continued)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Digital-Analog Glitch Impulse |  | Code 128 to 127 |  |  | 12 |  | nV-s |
| Signal to Noise + Distortion Ratio (MAX517, MAX519) | SINAD | $\begin{aligned} & \mathrm{V}_{\text {REF }_{-}}=4 \mathrm{Vp}-\mathrm{p} \text { at } 1 \mathrm{kHz}, \mathrm{VDD}=5 \mathrm{~V}, \\ & \text { Code }=\mathrm{FF} \text { hex } \end{aligned}$ |  |  | 87 |  | dB |
| Multiplying Bandwidth (MAX517, MAX519) |  | $\mathrm{VREF}_{-}=4 \mathrm{Vp}-\mathrm{p}, 3 \mathrm{~dB}$ bandwidth |  |  | 1 |  | MHz |
| Wideband Amplifier Noise |  |  |  |  | 60 |  | $\mu \mathrm{V}_{\mathrm{RMS}}$ |
| POWER REQUIREMENTS |  |  |  |  |  |  |  |
| Supply Voltage | VDD |  |  | 4.5 |  | 5.5 | V |
| Supply Current | IDD | Normal mode, output(s) unloaded, all digital inputs at OV or $\mathrm{V}_{\mathrm{DD}}$ | MAX517C |  | 1.5 | 3.0 | mA |
|  |  |  | MAX517E/M |  | 1.5 | 3.5 |  |
|  |  |  | MAX518C, MAX519C |  | 2.5 | 5 |  |
|  |  |  | MAX518E/M, MAX519E/M |  | 2.5 | 6 |  |
|  |  | Power-down mode |  |  | 4 | 20 | $\mu \mathrm{A}$ |

## TIMING CHARACTERISTICS

( $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} \pm 10 \%, \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\text {MIN }}$ to $\mathrm{T}_{\mathrm{MAX}}$, unless otherwise noted. Typical values are $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.)

| PARAMETER | SYMBOL | CONDITIONS | MIN TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Serial Clock Frequency | fSCL |  | 0 | 400 | kHz |
| Bus Free Time Between a STOP and a START Condition | tBUF |  | 1.3 |  | $\mu \mathrm{s}$ |
| Hold Time, (Repeated) Start Condition | thD, STA |  | 0.6 |  | $\mu \mathrm{s}$ |
| Low Period of the SCL Clock | tLow |  | 1.3 |  | $\mu \mathrm{s}$ |
| High Period of the SCL Clock | thigh |  | 0.6 |  | $\mu \mathrm{s}$ |
| Setup Time for a Repeated START Condition | tSU, STA |  | 0.6 |  | $\mu \mathrm{s}$ |
| Data Hold Time | thD, DAT | (Note 9) | 0 | 0.9 | $\mu \mathrm{S}$ |
| Data Setup Time | tSU, DAT |  | 100 |  | ns |
| Rise Time of Both SDA and SCL Signals, Receiving | tR | (Note 10) | $20+0.1 \mathrm{Cb}$ | 300 | ns |
| Fall Time of Both SDA and SCL Signals, Receiving | $\mathrm{t}_{\mathrm{F}}$ | (Note 10) | $20+0.1 \mathrm{Cb}$ | 300 | ns |
| Fall Time of SDA Transmitting (Note 7) | tF | ISINK $\leq 6 \mathrm{~mA}$ (Note 10) | $20+0.1 \mathrm{Cb}$ | 250 | ns |
| Setup Time for STOP Condition | tsu, STO |  | 0.6 |  | $\mu \mathrm{s}$ |
| Capacitive Load for Each Bus Line | Cb |  |  | 400 | pF |
| Pulse Width of Spike Suppressed | tsp | (Notes 6, 11) | 0 | 50 | ns |

Note 1: For the MAX518 (full-scale = $V_{D D}$ ) the last three codes are excluded from the TUE and DNL specifications, due to the limited output swing when loaded with $10 \mathrm{k} \Omega$ to GND
Note 2: Input resistance is code dependent. The lowest input resistance occurs at code $=55$ hex.
Note 3: Input capacitance is code dependent. The highest input capacitance occurs at code FF hex.
Note 4: $\mathrm{V}_{\text {REF }}=4 \mathrm{Vp}-\mathrm{p}, 10 \mathrm{kHz}$. Channel-to-channel isolation is measured by setting the code of one DAC to FF hex and setting the code of all other DACs to 00 hex
Note 5: $\operatorname{VREF}=4 \mathrm{Vp}-\mathrm{p}, 10 \mathrm{kHz}, \mathrm{DAC}$ code $=00$ hex.
Note 6: Guaranteed by design.
Note 7: $1^{2} \mathrm{C}$ compatible mode.
Note 8: Output settling time is measured by taking the code from 00 hex to FF hex, and from FF hex to 00 hex.
Note 9: A master device must provide a hold time of at least 300ns for the SDA signal (referred to VIL of the SCL signal) in order to bridge the undefined region of SCL's falling edge.
Note 10: $\mathrm{Cb}=$ total capacitance of one bus line in pF . $\mathrm{t}_{\mathrm{R}}$ and $\mathrm{tF}_{\mathrm{F}}$ measured between $0.3 \mathrm{~V}_{\mathrm{DD}}$ and $0.7 \mathrm{~V}_{\mathrm{DD}}$.
Note 11: Input filters on the SDA and SCL inputs suppress noise spikes less than 50 ns .

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| PIN |  |  |  | MAX519 |
| :---: | :---: | :---: | :---: | :--- |
|  | MAX518 | NAME |  |  |
| MAX517 | 1 |  |  |  |
| 1 | 2 | 4 | FUT0 | DAC0 Voltage Output |
| 2 | - | 5 | GND | Ground |
| - | 3 | 6 | AD3 | Address Input 3; sets IC's slave address |
| 3 | 4 | 8 | SCL | Serial Clock Input |
| 4 | - | 9 | SDA | Serial Data Input |
| - | 5 | 10 | AD2 | Address Input 2; sets IC's slave address |
| 5 | 6 | 11 | AD0 | Address Input 1; sets IC's slave address |
| 6 | 7 | 12 | VDD | Power Supply, +5V; used as reference for MAX518 |
| 7 | - | 13 | REF1 | Reference Voltage Input for DAC1 |
| - | 8 | 15 | REF0 | Reference Voltage Input for DAC0 |
| 8 | - | $2,3,7,14$ | N.C. | No Connect—not internally connected. |
| - |  |  |  |  |
| - |  |  |  |  |



Figure 1. MAX517/MAX519 Functional Diagram

Detailed Description

## Serial Interface

The MAX517/MAX518/MAX519 use a simple 2-wire serial interface requiring only two I/O lines (2-wire bus) of a standard microprocessor ( $\mu \mathrm{P}$ ) port. Figure 2 shows the timing diagram for signals on the 2 -wire bus. Figure 3 shows a typical application. The 2 -wire bus can have several devices (in addition to the MAX517/ MAX518/MAX519) attached. The two bus lines (SDA and SCL) must be high when the bus is not in use. When in use, the port bits are toggled to generate the appropriate signals for SDA and SCL. External pull-up resistors are not required on these lines. The MAX517/MAX518/ MAX519 can be used in applications where pull-up resistors are required (such as in $1^{2} \mathrm{C}$ systems) to maintain compatibility with existing circuitry.
The MAX517/MAX518/MAX519 are receive-only devices and must be controlled by a bus master device. They operate at SCL rates up to 400 kHz . A master device sends information to the devices by transmitting their address over the bus and then transmitting the desired information. Each transmission consists of a START condition, the MAX517/MAX518/MAX519's programmable slave-address, one or more command-byte/out-put-byte pairs (or a command byte alone, if it is the last byte in the transmission), and finally, a STOP condition (Figure 4).

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MAX517/MAX518/MAX519


Figure 2. Two-Wire Serial Interface Timing Diagram


Figure 3. MAX517/MAX518/MAX519 Application Circuit

The address byte and pairs of command and output bytes are transmitted between the START and STOP conditions. The SDA state is allowed to change only while SCL is low, with the exception of START and STOP conditions. SDA's state is sampled, and therefore must remain stable while SCL is high. Data is transmitted in 8-bit bytes. Nine clock cycles are required to transfer the data bits to the MAX517/MAX518/MAX519. Set SDA low during the 9th clock cycle as the MAX517/MAX518/MAX519 pull SDA low during this time. Rc (see Figure 3) limits the current that flows during this time if SDA stays high for short periods of time.

The START and STOP Conditions
When the bus is not in use, both SCL and SDA must be high. A bus master signals the beginning of a transmission with a START condition by transitioning SDA from high to low while SCL is high (Figure 5). When the master has finished communicating with the slave, it issues a STOP condition by transitioning SDA from low to high while SCL is high. The bus is then free for another transmission.

The Slave Address
The MAX517/MAX518/MAX519 each have a 7-bit long slave address (Figure 6). The first three bits (MSBs) of the slave address have been factory programmed and are always 010. In addition, the MAX517 and MAX518 have the next two bits factory programmed to 1 s . The logic state of the address inputs (AD0 and AD1 on the MAX517/MAX518; AD0, AD1, AD2, and AD3 on the MAX519) determine the LSB bits of the 7-bit slave address. These input pins may be connected to VDD or DGND, or they may be actively driven by TTL or CMOS logic levels. The MAX517/MAX518 have four possible slave addresses and therefore a maximum of four of

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Figure 4. A Complete Serial Transmission


Figure 5. All communications begin with a START condition and end with a STOP condition, both generated by a bus master.


Figure 6. Address Byte
these devices may share the bus. The MAX519 has 16 possible slave addresses. The eighth bit (LSB) in the slave address byte should be low when writing to the MAX517/MAX518/MAX519.
The MAX517/MAX518/MAX519 monitor the bus continuously, waiting for a START condition followed by their slave address. When a device recognizes its slave address, it is ready to accept data.

The Command Byte and Output Byte A command byte follows the slave address. Figure 7 shows the format for the command byte. A command byte is usually followed by an output byte unless it is the last byte in the transmission. If it is the last byte, all bits except PD (power-down) and RST (reset) are


Figure 7. Command Byte
ignored. If an output byte follows the command byte, AO of the command byte indicates the digital address of the DAC whose input data latch receives the digital output data. Set this bit to 0 when writing to the MAX517. The data is transferred to the DAC's output latch during the STOP condition following the transmission. This allows both DACs of the MAX518/MAX519 to be updated simultaneously (Figure 8).
Setting the PD bit high powers down the MAX517/ MAX518/MAX519 following a STOP condition (Figure $9 \mathrm{a})$. If a command byte with PD set high is followed by an output byte, the addressed DAC's input latch will be updated and the data will be transferred to the DAC's output latch following the STOP condition (Figure 9b).

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Figure 8b. Setting Both DAC Outputs (MAX518/MAX519)


Figure 9. Entering the Power-Down State

## 2-Wire Serial 8-Bit DACs with Rail-to-Rail Outputs

Furthermore if the transmission's last command byte has PD high, the output latches are updated, but voltage outputs will not reflect the newly entered data because the DAC enters power-down mode when the STOP condition is detected. When in power-down, the DAC outputs float. In this mode, the supply current is a maximum of $20 \mu \mathrm{~A}$. A command byte with the PD bit low returns the MAX517/MAX518/MAX519 to normal operation following a STOP condition, with the voltage outputs reflecting the output-latch contents (Figures 10a and 10b). Because each subsequent command byte overwrites the previous PD bit, only the last command byte of a transmission affects the power-down state.
Setting the RST bit high clears the DAC input latches. The DAC outputs remain unchanged until a STOP condition is detected (Figure 11a). If a reset is issued, the
following output byte is ignored. Subsequent pairs of command/output bytes overwrite the input latches (Figure 11b).
All changes made during a transmission affect the MAX517/MAX518/MAX519's outputs only when the transmission ends and a STOP has been recognized.
The R0, R1, and R2 bits are reserved and must be set to zero.

I2C Compatibility
The MAX517/MAX518/MAX519 are fully compatible with existing ${ }^{12} \mathrm{C}$ systems. SCL and SDA are highimpedance inputs; SDA has an open drain that pulls the data line low during the 9th clock pulse. Figure 12 shows a typical I2C application.


Figure 10. Returning to Normal Operation from Power-Down


Figure 11. Resetting DAC Outputs

## 2-Wire Serial 8-Bit DACs with Rail-to-Rail Outputs



Figure 12. MAX517/MAX518/MAX519 Used in a Typical ${ }^{2} C$ Application Circuit

Additional START Conditions
It is possible to interrupt a transmission to a device with a new START (repeated start) condition (perhaps addressing another device), which leaves the input
latches with data that has not been transferred to the output latches (Figure 13). Only the currently addressed device will recognize a STOP condition and transfer data to its output latches. If the device is left with data in its input latches, the data can be transferred to the output latches the next time the device is addressed, as long as it receives at least one command byte and a STOP condition.

Early STOP Conditions
The addressed device recognizes a STOP condition at any point in a transmission. If the STOP occurs during a command byte, all previous uninterrupted command and output byte pairs are accepted, the interrupted command byte is ignored, and the transmission ends (Figure 14a). If the STOP occurs during an output byte, all previous uninterrupted command and output byte pairs are accepted, the final command byte's PD and RST bits are accepted, the interrupted output byte is ignored, and the transmission ends (Figure 14b).

## Analog Section

DAC Operation
The MAX518 and MAX519 contain two matched volt-age-output DACs. The MAX517 contains a single DAC. The DACs are inverted R-2R ladder networks that convert 8-bit digital words into equivalent analog output voltages in proportion to the applied reference voltages. The MAX518 has both DAC's reference inputs connected to VDD. Figure 15 shows a simplified diagram of one DAC.

MAX517/MAX519 Reference Inputs The MAX517 and MAX519 can be used for multiplying applications. The reference accepts a OV to VDD volt-


Figure 13. Repeated START Conditions

## 2-Wire Serial 8-Bit DACs with Rail-to-Rail Outputs



Figure 14. Early STOP Conditions

Table 1. Unipolar Code Table

| DAC CONTENTS | ANALOG OUTPUT |
| :---: | :---: |
| 11111111 | $+V_{\text {REF }}\left(\frac{255}{256}\right)$ |
| 10000001 | $+\operatorname{VREF}\left(\frac{129}{256}\right)$ |
| 10000000 | $+\operatorname{VREF}\left(\frac{128}{256}\right)=\frac{V_{\text {REF }}}{2}$ |
| 01111111 | $+\operatorname{VREF}\left(\frac{127}{256}\right)$ |
| 00000001 | $+V_{\text {REF }}\left(\frac{1}{256}\right)$ |
| 00000000 | $0 V$ |

age, both DC and AC signals. The voltage at each REF input sets the full-scale output voltage for its respective DAC. The reference voltage must be positive. The DAC's input impedance is code dependent, with the lowest value occurring when the input code is 55 hex or 0101 0101, and the maximum value occurring when the input code is 00 hex. Since the REF input resistance (RIN) is code dependent, it must be driven by a circuit with low output impedance (no more than RIN $\div 2000$ ) to maintain output linearity. The REF input capacitance is also code dependent, with the maximum value occurring at code FF hex (typically 30pF). The output voltage for any DAC can be represented by a digitally programmable voltage source as: VOUT $=(\mathrm{N} \times$ VREF $) /$ 256 , where $N$ is the numerical value of the DAC's binary input code.


Figure 15. DAC Simplified Circuit Diagram

## Output Buffer Amplifiers

The DAC voltage outputs are internally buffered precision unity-gain followers that slew up to $1 \mathrm{~V} / \mu \mathrm{s}$. The outputs can swing from 0 V to $\mathrm{V}_{\mathrm{DD}}$. With a 0 V to 4 V (or 4 V to 0 V ) output transition, the amplifier outputs typically settle to $1 / 2 \mathrm{LSB}$ in $6 \mu \mathrm{~s}$ when loaded with $10 \mathrm{k} \Omega$ in parallel with 100 pF . The buffer amplifiers are stable with any combination of resistive loads $\geq 2 \mathrm{k} \Omega$ and capacitive loads $\leq 300 \mathrm{pF}$.
The MAX517/MAX518/MAX519 are designed for unipo-lar-output, single-quadrant multiplication where the output voltages and the reference inputs are positive with respect to AGND. Table 1 shows the unipolar code.

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Figure 16. PC Board Layout for Minimizing MAX519 Crosstalk (bottom view)

## Applications Information

Power-Supply Bypassing and Ground Management
Bypass VDD with a $0.1 \mu \mathrm{~F}$ capacitor, located as close to VDD and GND as possible. Careful PC board layout minimizes crosstalk among DAC outputs, reference inputs, and digital inputs. Figure 16 shows the suggested PC board layout to minimize crosstalk.
When using the MAX518 (or the MAX517/MAX519 with $V_{D D}$ as the reference), you may want to add a noise filter to the VDD supply (Figure 17) or to the reference input(s) (Figure 18), especially in noisy environments. The reference input's bandwidth exceeds 1 MHz for AC signals, so disturbances on the reference input can easily affect the DAC output(s).
The maximum input current for a single reference input is $\mathrm{V}_{\mathrm{REF}} / 16 \mathrm{k} \Omega=\operatorname{lREF}$ (max). In Figure 17, choose RF so that changes in the reference input current will have little effect on the reference voltage. For example, with $R_{F}$ $=6 \Omega$, the maximum output error due to $R_{F}$ is given by:

$$
6 \Omega \times \operatorname{IREF}(\max )=1.9 \mathrm{mV} \text { or } 0.1 \mathrm{LSB}
$$

In Figure 18, there is a voltage drop across RF that adds to the TUE. This voltage drop is due to the sum of the reference input current ( $\mathrm{V}_{\mathrm{REF}} / 16 \mathrm{k} \Omega$ maximum), supply current ( 6 mA maximum), and the amplifier output current (VREF/RLOAD). Choose RF to limit this voltage drop to an acceptable value. For example, with a $10 \mathrm{k} \Omega$ load, you can limit the error due to $\mathrm{RF}_{\mathrm{F}}$ to 0.5 LSB ( 9.8 mV ) by selecting RF so that:

$$
\begin{aligned}
\mathrm{R}_{\mathrm{F}}= & \mathrm{V}_{\mathrm{R}_{\mathrm{F}}} / \mathrm{IR}_{\mathrm{R}} \leq 9.8 \mathrm{mV} /(5 \mathrm{~V} / 16 \mathrm{k} \Omega+6 \mathrm{~mA}+ \\
& 5 \mathrm{~V} / 10 \mathrm{k} \Omega)
\end{aligned}
$$

$R_{F} \leq 1.4 \Omega$


Figure 17. Reference Filter When Using $V_{D D}$ as a Reference


Figure 18. $V_{D D}$ Filter When Using $V_{D D}$ as a Reference

## 2-Wire Serial 8-Bit DACs with Rail-to-Rail Outputs



Ordering Information (continued)

| PART | TEMP. RANGE | PIN-PACKAGE | TUE <br> (LSB) |
| :--- | :--- | :--- | :---: |
| MAX517AEPA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 Plastic DIP | 1 |
| MAX517BEPA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 Plastic DIP | 1.5 |
| MAX517AESA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SO | 1 |
| MAX517BESA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SO | 1.5 |
| MAX517BMJA | $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 8 CERDIP** | 1.5 |
| MAX518ACPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Plastic DIP | 1 |
| MAX518BCPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Plastic DIP | 1.5 |
| MAX518ACSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 SO | 1 |
| MAX518BCSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 SO | 1.5 |
| MAX518BC/D | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | Dice | 1.5 |
| MAX518AEPA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 Plastic DIP | 1 |
| MAX518BEPA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 Plastic DIP | 1.5 |
| MAX518AESA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SO | 1 |
| MAX518BESA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SO | 1.5 |
| MAX518BMJA | $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 8 CERDIP** | 1.5 |
| MAX519ACPE | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 16 Plastic DIP | 1 |
| MAX519BCPE | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 16 Plastic DIP | 1.5 |
| MAX519ACSE | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 16 Narrow SO | 1 |
| MAX519BCSE | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 16 Narrow SO | 1.5 |
| MAX519BC/D | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | Dice ${ }^{*}$ | 1.5 |
| MAX519AEPE | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 Plastic DIP | 1 |
| MAX519BEPE | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 Plastic DIP | 1.5 |
| MAX519AESE | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 Narrow SO | 1 |
| MAX519BESE | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 Narrow SO | 1.5 |
| MAX519BMJE | $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 16 CERDIP** | 1.5 |

*Dice are specified at $T_{A}=+25^{\circ} \mathrm{C}, D C$ parameters only
**Contact factory for availability and processing to MIL-STD-883.


6TGXVW/8IGXVW/LLGXVW

TRANSISTOR COUNT: 1797
SUBSTRATE CONNECTED TO VDD

## 2-Wire Serial 8-Bit DACs with Rail-to-Rail Outputs



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