

Tiny 500mA, 4MHz/2MHz Synchronous **Step-Down DC-DC Converters**

General Description

The MAX8640Y/MAX8640Z step-down converters are

optimized for applications where small size, high effi-

ciency, and low output ripple are priorities. They utilize

a proprietary PWM control scheme that optimizes the

switching frequency for high efficiency with small exter-

nal components and maintains low output ripple volt-

age at all loads. The MAX8640Z switches at up to

4MHz to allow a tiny 1µH inductor and 2.2µF output

capacitor. The MAX8640Y switches at up to 2MHz for

higher efficiency while still allowing small 2.2µH and

4.7µF components. Output current is guaranteed up to 500mA, while typical quiescent current is 28µA.

Factory-preset output voltages from 0.8V to 2.5V elimi-

Internal synchronous rectification greatly improves efficiency and replaces the external Schottky diode required in conventional step-down converters. Internal fast soft-start eliminates inrush current so as to reduce

The MAX8640Y/MAX8640Z are available in the tiny 6pin, SC70 (2.0mm x 2.1mm) and µDFN (1.5mm x 1.0mm) packages. Both packages are lead-free.

Microprocessor/DSP Core Power

Cell Phones, PDAs, DSCs, MP3s

Other Handhelds Where Space Is Limited

nate external feedback components.

input capacitor requirements.

I/O Power

INPUT

2.7V TO 5.5V

C1

ON/OFF

2.2µF

IN

GND

SHDN

Features

- Tiny SC70 and µDFN Packages 500mA Guaranteed Output Current
 - ♦ 4MHz or 2MHz PWM Switching Frequency
 - ♦ Tiny External Components: 1µH/2.2µF or
 - 2.2µH/4.7µF
 - 28µA Quiescent Current
 - Factory Preset Outputs from 0.8V to 2.5V
 - ♦ ±1% Initial Accuracy
 - Low Output Ripple at All Loads
 - Ultrasonic Skip Mode Down to 1mA Loads
 - Ultra-Fast Line- and Load-Transient Response
 - Fast Soft-Start Eliminates Inrush Current

Ordering Information

PART*	PIN- PACKAGE	TOP MARK
MAX8640YEXT08+T	6 SC70	ACQ
MAX8640YEXT10+T	6 SC70	ADF
MAX8640YEXT11+T	6 SC70	ACR
MAX8640YEXT12+T	6 SC70	ACS
MAX8640YEXT13+T	6 SC70	ACG
MAX8640YEXT15+T	6 SC70	ADD
MAX8640YEXT16+T	6 SC70	ADB
MAX8640YEXT18+T	6 SC70	ACI
MAX8640YEXT19+T	6 SC70	ACH
MAX8640YEXT82+T	6 SC70	ADJ

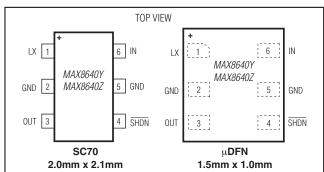
*Contact factory for availability of each version. For 2.85V output (82 version), request application note that includes limitations and typical operating characteristics.

+Denotes a lead(Pb)-free/RoHS-compliant package. T = Tape and reel.

Note: All devices are specified over the -40°C to +85°C operating temperature range.

Ordering Information continued and Selector Guide appears at end of data sheet.

Pin Configurations



For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maximintegrated.com.

OUTPUT

0.8V TO 2.5V

UP TO 500mA

Applications

LX

001

MAX8640Y

MAX8640Z

11

1μH OR 2.2μH

C2

2.2µF OR 4.7µF

Typical Operating Circuit

Tiny 500mA, 4MHz/2MHz Synchronous Step-Down DC-DC Converters

ABSOLUTE MAXIMUM RATINGS

IN to GND	0.3V to +6V
OUT, SHDN to GND	0.3V to (V _{IN} + 0.3V)
LX Current (Note 1)	
OUTPUT Short Circuit to GND	
Continuous Power Dissipation ($T_A = +70$)°C)
6-Pin SC70 (derate 3.1mW/°C above	+70°C)245mW

6-Pin µDFN (derate 2.1mW/°C above +70°C)	167.7mW
Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Note 1: LX has internal clamp diodes to IN and GND. Applications that forward bias these diodes should not exceed the IC's package power-dissipation limit.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

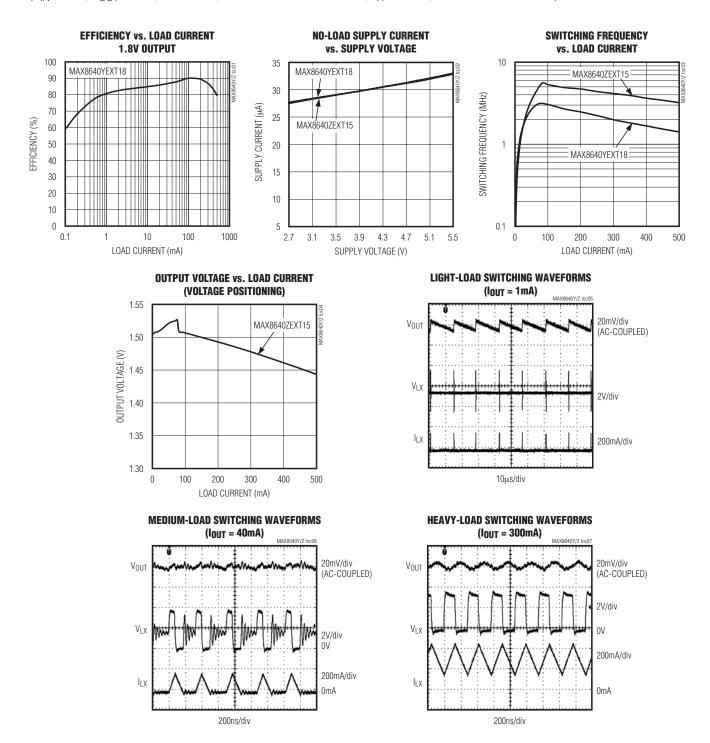
 $(V_{IN} = 3.6V, \overline{SHDN} = IN, T_A = -40^{\circ}C$ to +85°C, typical values are at $T_A = +25^{\circ}C$, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CONE	MIN	ТҮР	МАХ	UNITS		
Supply Range	VIN		2.7		5.5	V		
UVLO Threshold	UVLO	VIN rising, 100mV hyste	eresis	2.44	2.6	2.70	V	
		No load, no switching			28	48		
Supply Current	ICC	SHDN = GND	$T_A = +25^{\circ}C$		0.01	0.1	μA	
		SHDIN = GIND	$T_A = +85^{\circ}C$		0.1			
Output Voltage Range	Vout	Factory preset	0.8		2.5	V		
Output Voltage Accuracy		$I_{LOAD} = 0mA, T_A = +25$	5°C	-1	0	+1	%	
(Falling Edge)		$I_{LOAD} = 0mA, T_A = -40^\circ$	$OAD = 0mA$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$				70	
Output Load Regulation (Voltage Positioning)		Equal to inductor DC re		RL		V/A		
	VIH	V _{IN} = 2.7V to 5.5V	1.4			V		
SHDN Logic Input Level	VIL	V _{IN} = 2.7V to 5.5V			0.4			
SHDN Input Bias Current	lu . u	$V_{IN} = 5.5V,$	$T_A = +25^{\circ}C$		0.001	1		
SHDN IIIput blas Culterit	liH,i∟	$\overline{SHDN} = GND \text{ or } IN$	$T_A = +85^{\circ}C$		0.01		μA	
Minimum Required SHDN Reset Time	t _{SHDN}			10			μs	
Peak Current Limit	ILIMP	pFET switch		590	770	1400	mA	
Valley Current Limit	ILIMN	nFET rectifier		450	650	1300	mA	
Rectifier Off-Current Threshold	ILXOFF	nFET rectifier		10	40	70	mA	
On-Resistance	Ronp	pFET switch, $I_{LX} = -40r$	nA		0.6	1.2		
On-Resistance	R _{ONN}	nFET rectifier, $I_{LX} = 40r$	nA		0.35	0.7	Ω	
LX Lookogo Current	harre	V _{IN} = 5.5V, LX = GND	$T_A = +25^{\circ}C$		0.1	1		
LX Leakage Current	ILXLKG	to IN, $\overline{SHDN} = GND$	$T_A = +85^{\circ}C$		1		μA	
Minimum On and Off Times	ton(MIN)				95		20	
	toff(MIN)				95		ns	
Thermal Shutdown					+160		°C	
Thermal-Shutdown Hysteresis					20		°C	

Note 2: All devices are 100% production tested at $T_A = +25^{\circ}$ C. Limits over the operating temperature range are guaranteed by design.

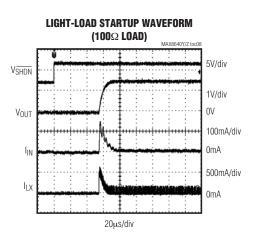
Typical Operating Characteristics

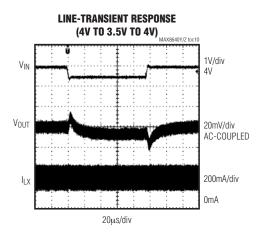
(V_{IN} = 3.6V, V_{OUT} = 1.5V, MAX8640Z, L = Murata LQH32CN series, T_A = +25°C, unless otherwise noted.)

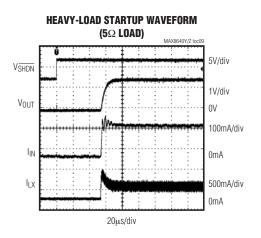


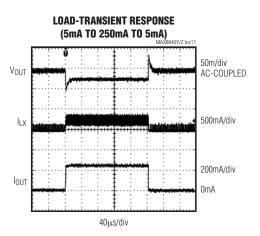
Typical Operating Characteristics (continued)

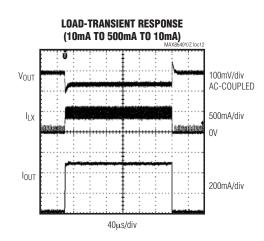
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Pin Description

PIN	NAME	FUNCTION
1	LX	Inductor Connection to the Internal Drains of the p-channel and n-channel MOSFETs. High impedance during shutdown.
2, 5	GND	Ground. Connect these pins together directly under the IC.
3	OUT	Output Sense Input. Bypass with a ceramic capacitor as close as possible to pin 3 (OUT) and pin 2 (GND). OUT is internally connected to the internal feedback network.
4	SHDN	Active-Low Shutdown Input. A logic-low on \overline{SHDN} disables the step-down DC-DC and resets the logic. A logic- high on \overline{SHDN} enables the step-down DC-DC. Ensure that \overline{SHDN} is low for $\geq 10\mu s$ (t _{SHDN}) after V _{IN} rises above its undervoltage lockout threshold (UVLO) to reset the logic. See the <i>Shutdown Mode</i> section for more information.
6	IN	Supply Voltage Input. Input voltage range is 2.7V to 5.5V. Bypass with a ceramic capacitor as close as possible to pin 6 (IN) and pin 5 (GND).

Detailed Description

The MAX8640Y/MAX8640Z step-down converters deliver over 500mA to outputs from 0.8V to 2.5V. They utilize a proprietary hysteretic PWM control scheme that switches at up to 4MHz (MAX8640Z) or 2MHz (MAX8640Y), allowing some trade-off between efficiency and size of external components. At loads below 100mA, the MAX8640Y/MAX8640Z automatically switch to pulse-skipping mode to minimize the typical quiescent current (28 μ A). Output ripple remains low at all loads, while the skip-mode switching frequency remains ultrasonic down to 1mA (typ) loads. Figure 1 is the simplified functional diagram.

Control Scheme

A proprietary hysteretic PWM control scheme ensures high efficiency, fast switching, fast transient response, low output ripple, and physically tiny external components. This control scheme is simple: when the output voltage is below the regulation threshold, the error comparator begins a switching cycle by turning on the high-side switch. This switch remains on until the minimum on-time expires and the output voltage is above the regulation threshold or the inductor current is above the current-limit threshold. Once off, the high-side switch remains off until the minimum off-time expires and the output voltage falls again below the regulation threshold. During the off period, the low-side synchronous rectifier turns on and remains on until either the high-side switch turns on again or the inductor current approaches zero. The internal synchronous rectifier eliminates the need for an external Schottky diode.

Voltage-Positioning Load Regulation

The MAX8640Y/MAX8640Z utilize a unique feedback network. By taking DC feedback from the LX node, the usual phase lag due to the output capacitor is removed, making the loop exceedingly stable and allowing the use of very small ceramic output capacitors. This configuration yields load regulation equal to the inductor's series resistance multiplied by the load current. This voltage-positioning load regulation greatly reduces overshoot during load transients, effectively halving the peak-to-peak output-voltage excursions compared to traditional step-down converters. See the Load-Transient Response in the *Typical Operating Characteristics*.

Shutdown Mode

A logic-low on SHDN places the MAX8640Y/MAX8640Z in shutdown mode by disabling the step-down DC-DC and resetting its logic. In shutdown mode, the supply current (I_{CC}) is reduced to 0.01µA typical. Additionally, the power MOSFETs between IN, LX, and GND (Figure 1) are open such that LX is high impedance.

Ensure that \overline{SHDN} is low for $\geq 10\mu s$ (t_{SHDN}) after V_{IN} rises above its undervoltage lockout threshold (UVLO) to reset the logic. In the majority of systems, this t_{SHDN} requirement is fulfilled naturally because the upstream logic controlling \overline{SHDN} is powered off of the same V_{IN} as the MAX8640Y/MAX8640Z. However, systems that want an always on regulator without the burden of enable/disable logic can use an R and C circuit on \overline{SHDN} as shown in Figure 2.

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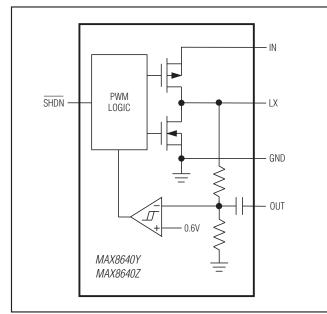


Figure 1. Simplified Functional Diagram

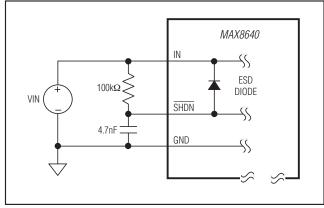


Figure 2. Using an R and C circuit to create an always on regulator

Soft-Start

The MAX8640Y/MAX8640Z include internal soft-start circuitry that eliminates inrush current at startup, reducing transients on the input source. Soft-start is particularly useful for higher impedance input sources, such

as Li+ and alkaline cells. See the Soft-Start Response in the *Typical Operating Characteristics*.

Applications Information

The MAX8640Y/MAX8640Z are optimized for use with a tiny inductor and small ceramic capacitors. The correct selection of external components ensures high efficiency, low output ripple, and fast transient response.

Inductor Selection

A 1 μ H inductor is recommended for use with the MAX8640Z, and 2.2 μ H is recommended for the MAX8640Y. A 1 μ H inductor is physically smaller but requires faster switching, resulting in some efficiency loss. Table 1 lists several recommended inductors.

It is acceptable to use a 1.5 μ H inductor with either the MAX8640Y or MAX8640Z, but efficiency and ripple should be verified. Similarly, it is acceptable to use a 3.3 μ H inductor with the MAX8640Y, but performance should be verified.

For optimum voltage positioning of load transients, choose an inductor with DC series resistance in the $75m\Omega$ to $150m\Omega$ range. For higher efficiency at heavy loads (above 200mA) or minimal load regulation (but some transient overshoot), the resistance should be kept as low as possible. For light-load applications up to 200mA, higher resistance is acceptable with very little impact on performance.

Capacitor Selection

Output Capacitor

The output capacitor, C2, is required to keep the output voltage ripple small and to ensure regulation loop stability. C2 must have low impedance at the switching frequency. Ceramic capacitors are recommended due to their small size and low ESR. Make sure the capacitor maintains its capacitance over temperature and DC bias. Capacitors with X5R or X7R temperature characteristics typically perform well. The output capacitance can be very low; see the *Selector Guide* for recommended capacitance values. For optimum load-transient performance and very low output ripple, the output capacitor value in μ F should be equal to or larger than the inductor value in μ H.

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MANUFACTURER	SERIES	INDUCTANCE (µH)	DC RESISTANCE (Ω typ)	CURRENT RATING (mA)	DIMENSIONS L x W x H (mm)		
	MIPFT2520D	2.0	0.16	900	2.5 x 2.0 x 0.5		
FDK Murata Sumida Taiyo Yuden		1.5	0.07	1500			
	MIPF2520D	2.2	0.08	1300	2.5 x 2.0 x 1.0		
		3.3	0.10	1200			
		1.0	0.12	1200			
Murata	LQM31P	1.5	0.16	1000	3.2 x 1.6 x 0.95		
		2.2	0.22	900			
		1.2	0.08	590			
Sumida	CDRH2D09	1.5	0.09	520	3.0 x 3.0 x 1.0		
		2.2	0.12	440			
		1.0	0.11	1100			
Taiyo Yuden	CKP3216T	1.5	0.13	1000	3.2 x 1.6 x 0.9		
		2.2	0.14	900			
		1.0	0.15	460			
	GLF201208T	2.2	0.36	300	2.0 x 1.25 x 0.9		
		1.0	0.07	400	0.01.051.05		
TDK	GLF2012T	2.2	0.10	300	2.0 x 1.25 x 1.35		
		1.0	0.10	800			
	GLF251812T	2.2	0.20	600	2.5 x 1.8 x 1.35		
		1.0	0.05	1000			
TOKO	MDT2520-CR	2.2	0.08	700	2.5 x 2.0 x 1.0		
ТОКО	D2912C	1.0	0.07	1100			
	D2812C	2.2	0.14	770	2.8 x 2.8 x 1.2		

Table 1. Suggested Inductors

Input Capacitor

The input capacitor, C1, reduces the current peaks drawn from the battery or input power source and reduces switching noise in the IC. The impedance of C1 at the switching frequency should be kept very low. Ceramic capacitors are recommended due to their small size and low ESR. Make sure the capacitor maintains its capacitance over temperature and DC bias. Capacitors with X5R or X7R temperature characteristics typically perform well. Due to the MAX8640Y/MAX8640Z soft-start, the input capacitance can be very low. For optimum noise immunity and low input ripple, choose a capacitor value in μ F that is equal to or larger than the inductor's value in μ H.

PCB Layout and Routing High switching frequencies and large peak currents make PCB layout a very important part of design. Good design minimizes excessive EMI on the feedback paths and voltage gradients in the ground plane, both of which can result in instability or regulation errors. Connect the inductor, input capacitor, and output capacitor as close together as possible, and keep their traces short, direct, and wide. Connect the two GND pins under the IC and directly to the grounds of the input and output capacitors. Keep noisy traces, such as the LX node, as short as possible. Refer to the MAX8640Z evaluation kit for an example PCB layout and routing scheme.

Tiny 500mA, 4MHz/2MHz Synchronous Step-Down DC-DC Converters

Selector Guide

	OUTPUT	FREQUENCY	RECOMMENDE			
PART	VOLTAGE (V)	(MHz)	L1 (µH)	C2 (µF)	TOP MARK	
MAX8640YEXT08	0.8	1.2	2.2	10	ACQ	
MAX8640YEXT10	1.0	1.6	2.2	4.7	ADF	
MAX8640YEXT11	1.1	1.7	2.2	4.7	ACR	
MAX8640YEXT12	1.2	1.8	2.2	4.7	ACS	
MAX8640YEXT13	1.3	1.9	2.2	4.7	ACG	
MAX8640YEXT15	1.5	2.0	2.2	4.7	ADD	
MAX8640YEXT16	1.6	2.0	2.2	4.7	ADB	
MAX8640YEXT18	1.8	2.0	2.2	4.7	ACI	
MAX8640YEXT19	1.9	2.0	2.2	4.7	ACH	
MAX8640YEXT24	2.4	2.0	2.2	4.7	ADM	
MAX8640YEXT25	2.5	1.7	2.2	4.7	ACJ	
MAX8640YEXT82	2.85	1.5	2.2	4.7	ADJ	
MAX8640YELT08	0.8	1.2	2.2	10	NB	
MAX8640YELT11	1.1	1.7	2.2	4.7	NC	
MAX8640YELT12	1.2	1.8	2.2	4.7	ND	
MAX8640YELT13	1.3	1.9	2.2	4.7	NE	
MAX8640YELT15	1.5	2.0	2.2	4.7	NF	
MAX8640YELT16	1.6	2.0	2.2	4.7	NG	
MAX8640YELT18	1.8	2.0	2.2	4.7	NH	
MAX8640YELT19	1.9	2.0	2.2	4.7	NI	
MAX8640YELT25	2.5	1.7	2.2	4.7	NJ	
MAX8640YELT82	2.85	1.5	2.2	4.7	OW	
MAX8640ZEXT08	0.8	2.4	1	4.7	ACL	
MAX8640ZEXT11	1.1	3.4	1	2.2	ACM	
MAX8640ZEXT12	1.2	3.6	1	2.2	ACN	
MAX8640ZEXT13	1.3	3.7	1	2.2	ACO	
MAX8640ZEXT15	1.5	3.9	1	2.2	ACP	
MAX8640ZEXT18	1.8	4.0	1	2.2	ACU	
MAX8640ZELT08	0.8	2.4	1	4.7	NK	
MAX8640ZELT11	1.1	3.4	1	2.2	NL	

Tiny 500mA, 4MHz/2MHz Synchronous **Step-Down DC-DC Converters**

Selector Guide (continued)

PART	OUTPUT	FREQUENCY	RECOMMENDED		
PARI	VOLTAGE (V)	(MHz)	L1 (µH)	C2 (µF)	TOP MARK
MAX8640ZELT12	1.2	3.6	1	2.2	NM
MAX8640ZELT13	1.3	3.7	1	2.2	NN
MAX8640ZELT15	1.5	3.9	1	2.2	NO
MAX8640ZELT18	1.8	4.0	1	2.2	NP

Ordering Information (continued)

PART*	PIN- PACKAGE	TOP MARK
MAX8640YEXT24+T	6 SC70	ADM
MAX8640YEXT25+T	6 SC70	ACJ
MAX8640YELT08+T	6 µDFN	NB
MAX8640YELT11+T	6 µDFN	NC
MAX8640YELT12+T	6 µDFN	ND
MAX8640YELT13+T	6 µDFN	NE
MAX8640YELT15+T	6 µDFN	NF
MAX8640YELT16+T	6 µDFN	NG
MAX8640YELT18+T	6 µDFN	NH
MAX8640YELT19+T	6 µDFN	NI
MAX8640YELT25+T	6 µDFN	NJ
MAX8640YELT82+T	6 µDFN	WO

*Contact factory for availability of each version. For 2.85V output (82 version), request application note that includes limitations and typical operating characteristics.

+Denotes a lead(Pb)-free/RoHS-compliant package. T = Tape and reel.

Note: All devices are specified over the -40°C to +85°C operating temperature range.

PART*	PIN- PACKAGE	TOP MARK
MAX8640ZEXT08+T	6 SC70	ACL
MAX8640ZEXT11+T	6 SC70	ACM
MAX8640ZEXT12+T	6 SC70	ACN
MAX8640ZEXT13+T	6 SC70	ACO
MAX8640ZEXT15+T	6 SC70	ACP
MAX8640ZEXT18+T	6 SC70	ACU
MAX8640ZELT08+T	6 µDFN	NK
MAX8640ZELT11+T	6 µDFN	NL
MAX8640ZELT12+T	6 µDFN	NM
MAX8640ZELT13+T	6 µDFN	NN
MAX8640ZELT15+T	6 µDFN	NO
MAX8640ZELT18+T	6 µDFN	NP

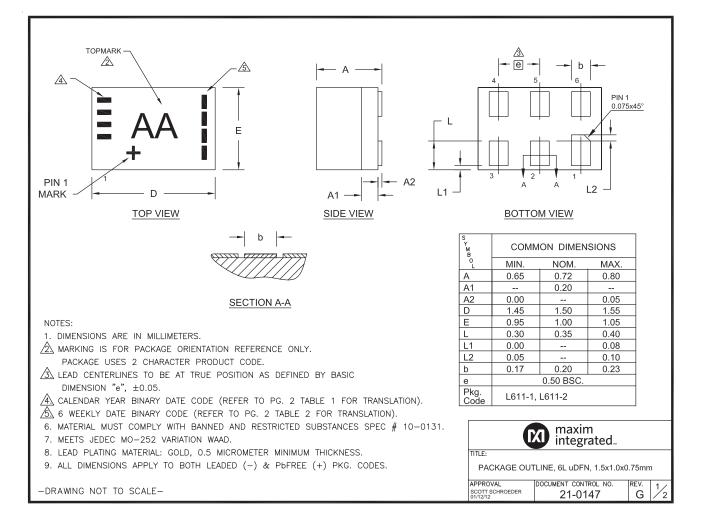
Chip Information

PROCESS: BICMOS

Package Information

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
6 µDFN	L611-1	<u>21-0147</u>
6 SC70	X6S-1	<u>21-0077</u>



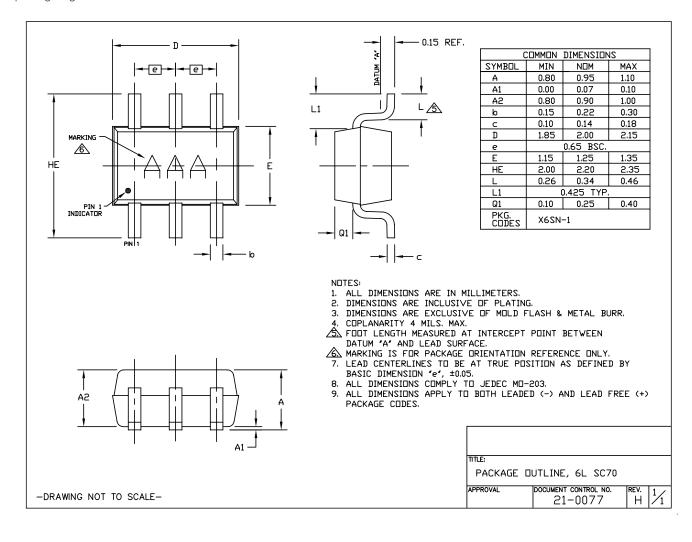
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TABLE 1 Translation	n Table fo	r Calenda	ar Year C	ode								
Calendar Year	2005	2006	2007	2008	2009	2010	2011	2012	2013	2014		
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TABLE 2 Translation		•		-								
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Package Information (continued)

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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
3	6/08	Added MAX8640YEXT10+T voltage option	7
4	2/09	Added MAX8640YEXT24+T voltage option and MAX8640YELT82+MAX8640YEXT82+ (82 = 2.85V), and corrected error	1, 2, 7, 8
5	8/14	Updated <i>Electrical Characteristics</i> table, <i>Pin Description</i> , and <i>Shutdown Mode</i> sections	2, 5



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