

General Description

The MAX9867 is an ultra-low power stereo audio codec designed for portable consumer devices such as mobile phones and portable gaming consoles.

The device features stereo differential microphone inputs that can be connected to either analog or digital microphones. The single-ended line inputs, with configurable preamplifier, can be sent to the ADC for record or routed directly to the headphone amplifier for playback. An auxiliary ADC path can be used to track any DC voltage.

The stereo headphone amplifiers support differential, single-ended, and capacitorless output configurations. Using the capacitorless output configuration, the device can output 10mW into 32Ω headphones. Comprehensive click-and-pop circuitry suppresses audible clicks and pops during volume changes and startup or shutdown.

Utilizing Maxim's proprietary digital circuitry, the device can accept any available 10MHz to 60MHz system clock. This architecture eliminates the need for an external PLL and multiple crystal oscillators. The stereo ADC and DAC paths provide user-configurable voiceband or audioband digital filters. Voiceband filters provide extra attenuation at the GSM packet frequency and greater than 70dB stopband attenuation at fs/2.

The MAX9867 operates from a single 1.8V supply, and supports a 1.65V to 3.6V logic level. An I²C 2-wire serial interface provides control for volume levels, signal mixing, and general operating modes.

The MAX9867 is available in a tiny 2.2mm x 2.7mm, 0.4mm-ball-pitch, WLP package. A 32-pin 5mm x 5mm TQFN package is also available.

Features

- ♦ 1.8V Single-Supply Operation
- 6.7mW Playback Power Consumption
- ♦ 90dB Stereo DAC, 8kHz ≤ fs ≤ 48kHz
- ♦ 85dB Stereo ADC, 8kHz ≤ fs ≤ 48kHz
- Battery-Measurement Auxiliary ADC
- Support for Any Master Clock Between 10MHz to 60MHz
- Stereo Digital Microphone Input Support
- Stereo Analog Differential Microphone Inputs
- Stereo Headphone Amplifiers: Differential, Single-Ended, or Capacitorless
- Stereo Line Inputs
- Voiceband Filter with a Stopband Attenuation Greater than 70dB
- ♦ 1.65V to 3.6V Digital Interface Supply Voltage
- ♦ I²S/TDM-Compatible Digital Audio Bus
- ♦ 30-Bump, 2.2mm x 2.7mm 0.4mm-Pitch WLP

Applications

Cell Phones Portable Gaming Devices Portable Navigation Devices Portable Multimedia Players Wireless Headsets

Ordering Information

DADT				
FARI	I EIVIF HANGE	PIN-PACKAGE		
MAX9867EWV+	-40°C to +85°C	30 WLP		
MAX9867ETJ+	-40°C to +85°C	32 TQFN-EP*		

+Denotes lead(Pb)-free/RoHS-compliant package.

*EP = Exposed pad.

Simplified Block Diagram



M/IXI/M

_ Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

MAX9867

ABSOLUTE MAXIMUM RATINGS

(Voltages with respect to AGND.)

DVDD, AVDD, and PVDD	-0.3V to +2V
DVDDIO	-0.3V to +3.6V
DGND and PGND	0.1V to +0.1V
PREG, REF, REG, MICBIAS	-0.3V to (AVDD + 0.3V)
MCLK, LRCLK, BCLK	
SDOUT, SDIN	$\dots -0.3V$ to (DVDDIO + 0.3V)
SDA, SCL, IRQ	0.3V to +3.6V
LOUTP, LOUTN, ROUTP,	
ROUTN	(PGND - 0.3V) to $(PVDD + 0.3V)$

LINL, LINR, JACKSNS/AUX, MICLP/DIGMICDATA,
MICLN/DIGMICCLK, MICRP, MICRN0.3V to (AVDD + 0.3V)
Continuous Power Dissipation ($T_A = +70^{\circ}C$)
30-Bump WLP (derate 12.5mW/°C above +70°C) 1000mW
32-Pin TQFN-EP (derate 34.5mW/°C above +70°C).2759mW
Junction-to-Ambient Thermal Resistance (θ_{JA}) (Note 1)
30-Bump WLP80°C/W
32-Pin TQFN-EP
Operating Temp Range40°C to +85°C
Storage Temp Range65°C to +150°C

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to <u>www.maxim-ic.com/thermal-tutorial</u>.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty$, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, $C_{REF} = 2.2\mu$ F, $C_{MICBIAS} = C_{PREG} = C_{REG} = 1\mu$ F, $AV_{PRE} = +20$ dB, $AV_{PGAM} = 0$ dB, $AV_{DAC} = 0$ dB, $AV_{LINE} = +20$ dB, $AV_{VOL} = 0$ dB, MCLK = 13MHz, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}$ C.) (Note 2)

PARAMETER	SYMBOL	CONDI	TIONS	MIN	ТҮР	MAX	UNITS
Supply Voltage Dange		PVDD, DVDD, AVDD		1.65	1.8	1.95	V
Supply Voltage Hange		DVDDIO		1.65	1.8	3.6	v
		Full-duplex 8kHz	Analog (AVDD + PVDD)		4.65	7	
		(Note 3)	Digital (DVDD + DVDDIO)		0.96	1.5	
		DAC playback 48kHz	Analog (AVDD + PVDD)		3.28	5	
Total Supply Current	Ivdd	(Note 3)	Digital (DVDD + DVDDIO)		1.40	2	
		Full-duplex 48kHz stereo (audio mode) (Note 3)	Analog (AVDD + PVDD)		8.0	12	ΠA
			Digital (DVDD + DVDDIO)		2.0	3	
		Stereo line-in only	Analog (AVDD + PVDD)		3.8	6	
			Digital (DVDD + DVDDIO)		0.004	0.05	
Shutdown Supply Current		T 0500	Analog (AVDD + PVDD)		1	5	
		T _A = +25°C	Digital (DVDD + DVDDIO)		1	5	μΑ

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty$, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, C_{REF} = 2.2µF, C_{MICBIAS} = C_{PREG} = C_{REG} = 1µF, AV_{PRE} = +20dB, AV_{PGAM} = 0dB, AV_{DAC} = 0dB, AV_{LINE} = +20dB, AV_{VOL} = 0dB, MCLK = 13MHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

PARAMETER	SYMBOL	COND	ITIONS	MIN	ТҮР	MAX	UNITS
Shutdown to Full Operation		Excludes PLL lock tim	Ie		10		ms
Soft-Start/-Stop Time					10		ms
DAC (Note 4)							
Dynamic Range (Note 5)	DR	$f_S = 48$ kHz, AV _{VOL} =	Master or slave mode		90		dB
		+00D, 1A = +20 O	Slave mode	84			
			Differential mode		1		
Full-Scale Output		$V_{OLL}/V_{OLR} = 0 \times 09$	Capacitorless and single-ended modes		0.56		VRMS
Gain Error		DC accuracy, measur full-scale output	red with respect to		1	5	%
Voice Dath Phase Dalay	Daux	f = 1kHz, 0dBFS, HP	f _S = 8kHz		1.2		
voice Path Phase Delay	PDLY	input to analog output	f _S = 16kHz		0.59		ms
Total Harmonic Distortion	THD	MCLK = 12.288 MHz, fs = 48 kHz, 0dBFS, measured at headphone outputs			-80		dB
DAC Attenuation Range	AV _{DAC}	DACA = 0xF to 0x0		-15		0	dB
DAC Gain Adjust	AVGAIN	DACG = 00 to 11		0		+18	dB
		$V_{AVDD} = V_{PVDD} = 1.65V$ to 1.95V		60	78		
		f = 217Hz, V _{RIPPLE} = 100mV _{P-P} , AV _{VOL} = 0dB			78		
Power-Supply Rejection Ratio	PSRR	$f = 1 kHz$, $V_{RIPPLE} = 100 mV_{P-P}$, $AV_{VOL} = 0 dB$			75		dB
		$f = 10kHz$, $V_{RIPPLE} = 100mV_{P-P}$, $AV_{VOL} = 0dB$			62		
DAC VOICE MODE DIGITAL IIR	LOWPASS FI	LTER					
		With respect to fs with 48kHz	nin ripple; $f_S = 8$ kHz to		0.448 x f _S		
Passband Cutoff	†PLP	-3dB cutoff			0.451 x f _S		Hz
Passband Ripple		f < fplp			±0.1		dB
Stopband Cutoff	f _{SLP}	With respect to f_S ; f_S :	= 8kHz to 48kHz		0.476 x f _S		Hz
Stopband Attenuation		$f > f_{SLP}, f = 20Hz to 2$	0kHz	75			dB

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, C_{REF} = 2.2 \mu F, C_{MICBIAS} = C_{PREG} = C_{REG} = 1 \mu F, AV_{PRE} = +20 dB, AV_{PGAM} = 0 dB, AV_{DAC} = 0 dB, AV_{LINE} = +20 dB, AV_{VOL} = 0 dB, MCLK = 13 MHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)$

PARAMETER	SYMBOL	CONDITIONS	MIN TYP MAX	UNITS
DAC VOICE MODE DIGITAL 5th 0	ORDER IIR H	IGHPASS FILTER		-
		DVFLT = 0x1 (elliptical tuned for 16kHz GSM + 217Hz notch)	0.0161 × fs	
5th Order Passband Cutoff (-3dB from Peak, I ² C Register Programmable)		DVFLT = 0x2 (500Hz Butterworth tuned for 16kHz)	0.0312 x fs	
	fdhppb	DVFLT = 0x3 (elliptical tuned for 8kHz GSM + 217Hz notch)	0.0321 x fs	Hz
		DVFLT = 0x4 (500Hz Butterworth tuned for 8kHz)	0.0625 x fs	
		DVFLT = 0x5 (f _S /240 Butterworth)	0.0042 x f _S	
5th Order Stopband Cutoff (-30dB from Peak, I ² C Register Programmable)		DVFLT = 0x1 (elliptical tuned for 16kHz GSM + 217Hz notch)	0.0139 × fs	
	fdhpsb	DVFLT = 0x2 (500Hz Butterworth tuned for 16kHz)	0.0156 x fs	
		DVFLT = 0x3 (elliptical tuned for 8kHz GSM + 217Hz notch)	0.0279 × fs	Hz
		DVFLT = 0x4 (500Hz Butterworth tuned for 8kHz)	0.0312 x f _S	
		DVFLT = 0x5 (f _S /240 Butterworth)	0.0021 x f _S	
DC Attenuation	DCATTEN	DVFLT ≠ 000	90	dB
DAC STEREO AUDIO MODE DIG	ITAL FIR LO	WPASS FILTER	1	
		With respect to f_S within ripple; $f_S = 8kHz$ to $48kHz$	0.43 x fs	
Passband Cutoff	fplp	-3dB cutoff	0.47 x f _S	Hz
		-6.02dB cutoff	0.50 x f _S	
Passband Ripple		f < fpLP	±0.1	dB
Stopband Cutoff	fSLP	With respect to f_S ; $f_S = 8kHz$ to $48kHz$	0.58 x f _S	Hz
Stopband Attenuation			60	dB

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty$, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, C_{REF} = 2.2µF, C_{MICBIAS} = C_{PREG} = C_{REG} = 1µF, AV_{PRE} = +20dB, AV_{PGAM} = 0dB, AV_{DAC} = 0dB, AV_{LINE} = +20dB, AV_{VOL} = 0dB, MCLK = 13MHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

PARAMETER	SYMBOL	COND	ITIONS	MIN	ТҮР	MAX	UNITS
DAC STEREO AUDIO MODE DIG	ITAL DC BLC	OCKING HIGHPASS FI	LTER				•
Passband Cutoff (-3dB from Peak)	fdhppb	DVFLT = 0x1			0.000625 x f _S		Hz
DC Attenuation	DCATTEN	DVFLT = 0x1			90		dB
ADC (Note 6)							
Dynamia Ranga (Nota 5)	ΠP	$f_S = 8kHz$, MODE = 0	(IIR voice)	75	84		dP
	Dh	$f_{\rm S}$ = 8kHz to 48kHz, N	10DE = 1 (FIR audio)		85		uВ
Full-Scale Input		Differential MIC input AVPRE = 0dB, AVPG	or stereo-line inputs, A = 0dB		1		Vp-p
Gain Error (Note 7)		DC accuracy, measur 80% of full-scale outp	DC accuracy, measured with respect to 80% of full-scale output		1	5	%
Voice Path Phase Delay	Deve	f = 1kHz, 0dBFS, HP filter disabled,	f _S = 8kHz		1.2		
	PDLY	analog input to digital output	f _S = 16kHz		0.61		ms
Total Harmonic Distortion	THD	f = 1kHz, f _S = 8kHz, T	A = +25°C, 0dBFs		-81	-70	dB
ADC Level Adjust Range	AVADC	AVL/VR = 0xF to $0x0$		-12		+3	dB
		V_{AVDD} = 1.65V to 1.95V, input referred		60	85		
		f = 217Hz, V _{RIPPLE} = 100mV, AV _{ADC} = 0dB, input referred			85		
Power-Supply Rejection Ratio	PSRR	f = 1kHz, V _{RIPPLE} = 100mV, AV _{ADC} = 0dB, input referred			80		dB
		$f = 10kHz$, $V_{RIPPLE} = 100mV$, $AV_{ADC} = 0dB$, input referred			80		
ADC VOICE MODE DIGITAL IIR L	OWPASS FI	LTER					
		With respect to fs with	nin ripple;		0.445 x		
Passhand Cutoff	foun	$f_S = 8$ kHz to 48kHz		fs			
	IFLF	-3dB cutoff			0.449 x f _S		112
Passband Ripple		f < fplp			±0.1		dB
Stopband Cutoff	fSLP	With respect to fs; fs :	= 8kHz to 48kHz		0.469 x f _S		Hz
Stopband Attenuation		$f > f_{SLP}, f = 20Hz to 2$	0kHz	74			dB

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, C_{REF} = 2.2 \mu F, C_{MICBIAS} = C_{PREG} = C_{REG} = 1 \mu F, AV_{PRE} = +20 dB, AV_{PGAM} = 0 dB, AV_{DAC} = 0 dB, AV_{LINE} = +20 dB, AV_{VOL} = 0 dB, MCLK = 13 MHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)$

PARAMETER	SYMBOL	CONDITIONS	CONDITIONS MIN TYP MAX						
ADC VOICE MODE DIGITAL 5th ORDER IIR HIGHPASS FILTER									
		AVFLT = 0x1 (elliptical tuned for 16kHz GSM + 217Hz notch)	0.0161 x fs						
5th Order Passband Cutoff (-3dB from Peak, I ² C Register Programmable)		AVFLT = 0x2 (500Hz Butterworth tuned for 16kHz)	0.0312 x fs						
	fahppb	AVFLT = 0x3 (elliptical tuned for 8kHz GSM + 217Hz notch)	0.0321 x f _S		Hz				
		AVFLT = 0x4 (500Hz Butterworth tuned for 8kHz)	0.0625 x fs						
		AVFLT = 0x5 (f _S /240 Butterworth)	0.0042 x f _S						
Stopband Cutoff (-30dB from Peak)	fahpsb	AVFLT = 0x1 (elliptical tuned for 16kHz GSM + 217Hz notch)	0.0139 × fs 0.0156 × fs		Hz				
		AVFLT = 0x2 (500Hz Butterworth tuned for 16kHz)							
		f_{AHPSB} AVFLT = 0x3 (elliptical tuned for 8kHz GSM + 217Hz notch) 0.0 x							
		AVFLT = 0x4 (500Hz Butterworth tuned for 8kHz)	FLT = 0x4 0.0312 0.0312 0.0312 0.0312						
		AVFLT = 0x5 (f _S /240 Butterworth)	0.0021 x fs						
DC Attenuation	DCATTEN	AVFLT ≠ 000	90		dB				
ADC STEREO AUDIO MODE DIGI	TAL FIR LO	WPASS FILTER							
		With respect to f_S within ripple; $f_S = 8kHz$ to $48kHz$	0.43 x f _S						
Passband Cutoff	fplp	-3dB cutoff	0.48 x f _S		Hz				
		-6.02dB cutoff 0.5 x f _S			1				
Passband Ripple		f < fpLp	±0.1		dB				
Stopband Cutoff	fSLP	With respect to f_S ; $f_S = 8kHz$ to $48kHz$	0.58 x fs		Hz				
Stopband Attenuation		$f > f_{SLP}, f = 20Hz$ to 20kHz	60		dB				

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty$, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, C_{REF} = 2.2µF, C_{MICBIAS} = C_{PREG} = C_{REG} = 1µF, AV_{PRE} = +20dB, AV_{PGAM} = 0dB, AV_{DAC} = 0dB, AV_{LINE} = +20dB, AV_{VOL} = 0dB, MCLK = 13MHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

PARAMETER	SYMBOL	CON	MIN	ТҮР	МАХ	UNITS	
ADC STEREO AUDIO MODE DIGI	TAL DC BLO	OCKING HIGHPASS	FILTER				
Passband Cutoff (-3dB from Peak)	fанррв	AVFLT = 0x1			0.000625 x fs	1	Hz
DC Attenuation	DCATTEN	AVFLT = 0x1			90		dB
OUTPUT VOLUME CONTROL							
		VOLL/VOLR = 0x00		14.55	14.9	15.15	
		VOLL/VOLR = 0x01		14.1	14.4	14.6	
		VOLL/VOLR = 0x02		13.6	13.9	14.1	
Line Input to Output Volume	AVVOL	VOLL/VOLR = 0x04		12.6	12.9	13.1	dB
		VOLL/VOLR = 0x08		9.35	9.9	10.35	
		VOLL/VOLR = 0x10		0.35	0.9	1.35	
		VOLL/VOLR = 0x20		-50.15	-49.2	-48.15	
		VOLL/VOLR = 0x00	to 0x06 (+6dB to +3dB)		0.5		
Output Volume Control Stop Size		VOLL/VOLR = 0x06	to 0x0F (+3dB to -6dB)		1		٩D
Output volume Control Step Size		VOLL/VOLR = $0x0F$ to $0x17$ (-6dB to -22dB)		2		uв	
		VOLL/VOLR = 0x17 to 0x3F (-22dB to mute)		4			
Output Volume Control Mute Attenuation		f = 1kHz		100		dB	
HEADPHONE AMPLIFIER (Note 8	5)						
Output Power per Channel	Pour	f = 1kHz, THD <	$R_L = 16\Omega$	30	52		m\\/
(Differential Mode)	FOUT	1%, $T_A = +25^{\circ}C$	$R_L = 32\Omega$		32		11100
Output Power per Channel	Pour	$f = 1kHz$, THD < $R_L =$	$R_L = 16\Omega$		19		m\\/
(Capacitorless Mode)	1001	1%, T _A = +25°C	$R_L = 32\Omega$	8	10		11100
		$R_L = 16\Omega$, $P_{OUT} = 2$	5mW, f = 1kHz		-76		
Total Harmonic Distortion + Noise	THD+N	$R_L = 32\Omega$, $P_{OUT} =$	MCLK = 13MHz, f _S = 8kHz		-77	-70	dB
		25mW, f = 1kHz	$\label{eq:MCLK} \begin{array}{l} MCLK = 12.288 MHz, \\ f_{S} = 48 kHz \end{array}$		-80		
		$R_L = 16\Omega, P_{OUT} = 6$	6.25mW, f = 1kHz		-72		
Total Harmonic Distortion + Noise	THD+N	$R_L = 32\Omega$, $P_{OUT} =$	$\begin{array}{l} \text{MCLK} = 13 \text{MHz}, \\ \text{f}_{\text{S}} = 8 \text{kHz} \end{array}$		-74	-65	dB
		6.25mW, f = 1kHz	$\begin{array}{l} \text{MCLK} = 12.288 \text{MHz}, \\ \text{f}_{\text{S}} = 48 \text{kHz} \end{array}$		-74		
		$R_L = 16\Omega, P_{OUT} = 6$	6.25mW, f = 1kHz		-74		
Total Harmonic Distortion + Noise	THD+N	$R_L = 32\Omega, P_{OUT} =$	MCLK = 13MHz, f _S = 8kHz		-74	-65	dB
(SE Mode)		6.25mW, f = 1kHz	$\begin{array}{l} \text{MCLK} = 12.288 \text{MHz}, \\ \text{f}_{\text{S}} = 48 \text{kHz} \end{array}$		-76		
Dynamic Range	DR	$AV_{VOL} = +6dB$ (Not	es 5, 7)	76	90		dB



ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, C_{REF} = 2.2 \mu F, C_{MICBIAS} = C_{PREG} = C_{REG} = 1 \mu F, AV_{PRE} = +20 dB, AV_{PGAM} = 0 dB, AV_{DAC} = 0 dB, AV_{LINE} = +20 dB, AV_{VOL} = 0 dB, MCLK = 13 MHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)$

PARAMETER	SYMBOL	CON	IDITIONS	MIN	ТҮР	МАХ	UNITS
		$V_{AVDD} = V_{PVDD} = 1.65V$ to 1.95V		60	78		
		$f = 217Hz, V_{RIPPLE}$ AV _{VOL} = 0dB	= 100mV _{P-P} ,		78		
Power-Supply Rejection Ratio (Note 7)	PSRR	$f = 1kHz, V_{RIPPLE} = AV_{VOL} = 0dB$	100mV _{P-P} ,		75		dB
		$f = 10kHz, V_{RIPPLE} = AV_{VOL} = 0dB$	= 100mV _{P-P} ,		62		
Output Offset Voltage	Vee	AV _{VOL} = -84dB differential mode	(LOUTP–LOUTN, ROUTP–ROUTN), T _A = +25°C		±0.2	±1.3	m\/
	VOS	AV _{VOL} = -84dB capacitorless mode	(LOUTP–LOUTN, ROUTP–LOUTN), T _A = +25°C		±0.8	±3	IIIV
		Differential mode, P _{OUT} = 5mW, f = 1kHz			87		
Crosstalk	X _{TALK}	ALK Capacitorless mode, $P_{OUT} = 5mW$, f = 1kHz	TQFN		55		dB
			WLP		60		
		No sustained	$R_L = 32\Omega$		500		ρE
		oscillations	RL = ∞		100		ρr
Click-and-Pop Level		Peak voltage, A-weighted, 32	Into shutdown		-80		dBV
Modes)		samples per second	Out of shutdown		-69		άΒν
Click-and-Pop Level		Peak voltage, A-weighted, 32	Into shutdown		-75		
(SE Mode)		samples per second	Out of shutdown		-75		aBV
MICROPHONE AMPLIFIER							
		$PALEN/PAREN = 0^{-1}$	1	-0.5	0	+0.5	
Preamplifier Gain	AVPRE	PALEN/PAREN = 10)	19.5	20	20.5	dB
		$PALEN/PAREN = 1^{-1}$	1	29.5	30	30.5	
MIC PGA Gain		PGAML/PGAMR = 0)x1F	-0.6	-0.1	+0.4	dB
	AVPGAM	PGAML/PGAMR = 0)x00	19.3	19.75	20.3	uD
Common-Mode Rejection Ratio	CMRR	$V_{IN} = 100mV_{P-P}, f =$	= 217Hz		50		dB
MIC Input Resistance	RIN_MIC	All gain settings		30	50		kΩ

ELECTRICAL CHARACTERISTICS (continued)

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PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
		AV _{PRE} = 0dB, V _{IN} = 1V _{P-P} , f = 1kHz		-80		
Total Harmonic Distortion + Noise	THD+N	$AV_{PRE} = +30dB,$ $V_{IN} = 32mV_{P-P}, f = 1kHz,$ $(1V_{P-P} at ADC input)$		-67		dB
		V _{AVDD} = 1.65V to 1.95V, input referred	60	85		
		$f = 217Hz$, $V_{RIPPLE} = 100mV$, $AV_{ADC} = 0dB$, input referred		85		
Power-Supply Rejection Ratio	PSRR	$f = 1 kHz$, $V_{RIPPLE} = 100 mV$, $AV_{ADC} = 0 dB$, input referred		80		dB
		$f = 10kHz$, $V_{RIPPLE} = 100mV$, $AV_{ADC} = 0dB$, input referred		80		
MICROPHONE BIAS						
Output Voltage	V _{MICBIAS}	$V_{AVDD} = 1.8V, I_{LOAD} = 1mA$	1.5	1.525	1.55	V
Load Regulation		$I_{LOAD} = 1 mA$ to $2 mA$		0.2	10	V/A
Line Regulation		V _{AVDD} = 1.65V to 1.95V		10		μV/V
Power-Supply Rejection Ratio	DCDD	$f = 217Hz$, $V_{RIPPLE} = 100mV_{P-P}$		85		dB
	1 3111	$f = 10kHz, V_{RIPPLE} = 100mV_{P-P}$		81		
Noise Voltage		A-weighted		9.1		μV _{RMS}
LINE INPUT	-		-			
Full-Scale Input	VIN	$AV_{LINE} = 0 dB$		1.0		VP-P
Line Input Level Adjust Range	AVLINE	LIGL/LIGR = 0xF to $0x0$	-6.5		+24.5	dB
Line Input Mute Attenuation		f = 1 kHz		100		dB
Input Resistance	RIN_LINE	$AV_{LINE} = +24 dB$	20			kΩ
Total Harmonic Distortion + Noise	THD+N	$V_{IN} = 0.1V_{P-P}$, f = 1kHz, differential output		-83		dB
AUXIN INPUT	-		-			
Input DC Voltage Range		AUXEN = 1	0		0.738	V
AUXIN Input Resistance	RIN	AUXEN = 1, $0V \le AUXIN \le 0.738$	10	40		MΩ
JACK SENSE OPERATION						
Thrachold	ЛТП	JDETEN = 1, SHDN = 1, JACKSNS	0.92 x MICBIAS	0.95 x MICBIAS	0.98 x MICBIAS	V
	VILL	JDETEN = 1, SHDN = 0, JACKSNS, LOUTP	AVDD - 0.8	AVDD - 0.4	AVDD - 0.15	v
Pullup Current	IDI I	JDETEN = 1, SHDN = 1, JACKSNS = GND		4		μA
Pullup Current		JDETEN = 1, \overline{SHDN} = 0, JACKSNS = LOUTP = GND		4	20	μΑ
Pullup Voltage		JDETEN = 1, JACKSNS, LOUTP		AVDD		V



ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, C_{REF} = 2.2 \mu F, C_{MICBIAS} = C_{PREG} = C_{REG} = 1 \mu F, AV_{PRE} = +20 dB, AV_{PGAM} = 0 dB, AV_{DAC} = 0 dB, AV_{LINE} = +20 dB, AV_{VOL} = 0 dB, MCLK = 13 MHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)$

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS
DIGITAL SIDETONE							
Sidetone Gain Adjust Range	AV _{STGA}	Differential output mode DVST = 0x1F to 0x01	9,	-60		0	dB
Voice Path Phase Delay	PDLY	MIC input to headphone HP filter disabled, $f_S = 8$	e output, f = 1kHz, BkHz		2.2		ms
INPUT CLOCK CHARACTERISTIC	CS	-					
MCLK Input Frequency	f MCLK	For any LRCLK sample	rate	10		60	MHz
MCLK Input Duty Cycle		Prescaler = /1 mode		40		60	%
		/2 or /4 modes		30		70	70
Maximum MCLK Input Jitter		Maximum allowable RM limits	S for performance		100		psRMS
LRCLK Sample Rate Range				8		48	kHz
		Any allowable LRCLK	Rapid lock mode		2	7	
LRCLK PLL Lock Time		and PCLK rate, slave mode	Nonrapid lock mode		12	25	ms
LRCLK Acceptable Jitter for Maintaining PLL Lock		Allowable LRCLK period change from nominal for slave PLL mode at any allowable LRCLK and PCLK rates				±100	ns
		FREQ = 0x8 through 0x	F	0		0	%
LRCLK Average Frequency Error (Master and Slave Modes)		PCLK = 192xf _S , 256xf _S , 768xf _S , and 1024xf _S	384xf _S , 512xf _S ,	0		0	
		All other modes		-0.025		+0.025	
DIGITAL INPUT (MCLK)							
Input High Voltage	VIH			1.2			V
Input Low Voltage	VIL					0.6	V
Input Leakage Current	I _{IH} , I _{IL}	$T_A = +25^{\circ}C$				±1	μA
Input Capacitance					10		рF
DIGITAL INPUTS (SDIN, BCLK, L	RCLK)	1					
Input High Voltage	VIH			0.7 x DVDDIO			V
Input Low Voltage	VIL					0.3 x DVDDIO	V
Input Hysteresis					200		mV
Input Leakage Current	I _{IH} , I _{IL}	$T_A = +25^{\circ}C$				±1	μA
Input Capacitance					10		pF

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty$, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, C_{REF} = 2.2µF, C_{MICBIAS} = C_{PREG} = C_{REG} = 1µF, AV_{PRE} = +20dB, AV_{PGAM} = 0dB, AV_{DAC} = 0dB, AV_{LINE} = +20dB, AV_{VOL} = 0dB, MCLK = 13MHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
DIGITAL INPUTS (SDA, SCL)						
Input High Voltage	VIH		0.7 x DVDD			V
Input Low Voltage	VIL				0.3 x DVDD	V
Input Hysteresis				200		mV
Input Leakage Current	I _{IH} , I _{IL}	$T_A = +25^{\circ}C$			±1	μA
Input Capacitance				10		рF
DIGITAL INPUT (DIGMICDATA)						
Input High Voltage	VIH		0.65 x DVDD			V
Input Low Voltage	V _{IL}				0.35 x DVDD	V
Input Hysteresis				100		mV
Input Leakage Current	I _{IH} , I _{IL}	$T_{A} = +25^{\circ}C$			±35	μΑ
Input Capacitance				10		рF
CMOS DIGITAL OUTPUTS (BCL)	, LRCLK, SE	OUT)				
Output Low Voltage	VOL	I _{OL} = 3mA			0.4	V
Output High Voltage	V _{OH}	I _{OH} = 3mA	DVDDIO - 0.4			V
CMOS DIGITAL OUTPUT (DIGMI	CCLK)		•			•
Output Low Voltage	V _{OL}	I _{OL} = 1mA			0.4	V
Output High Voltage	V _{OH}	I _{OH} = 1mA	DVDD - 0.4			V
OPEN-DRAIN DIGITAL OUTPUTS	(SDA, IRQ)					
Output High Current	Iон	$V_{OUT} = DVDD, T_A = +25^{\circ}C$			1	μA
Output Low Voltage	Vol	I _{OL} = 3mA			0.2 x DVDD	V
DIGITAL MICROPHONE TIMING	CHARACTER	RISTICS (V _{DVDD} = 1.65V)				
DICMICCI // Divide Batic	function	MICCLK = 00		PCLK/8		M⊔⇒
DIGIMICCER DIVIDE RALIO	IMICCLK	MICCLK = 01		PCLK/6		IVINZ
DIGMICDATA to DIGMICCLK Setup Time	tsu,MIC	Either clock edge	20			ns
DIGMICDATA to DIGMICCLK Hold Time	thd,mic	Either clock edge	0			ns
DIGITAL AUDIO INTERFACE TIM	ING CHARA	CTERISTICS (V _{DVDD} = 1.65V)				
	t BCLKS	Slave operation		75		ns
	^t BCLKM	Master operation		325		ns



ELECTRICAL CHARACTERISTICS (continued)

 $(V_{AVDD} = V_{PVDD} = V_{DVDD} = V_{DVDDIO} = +1.8V, R_L = \infty$, headphone load (R_L) connected between _OUTP and _OUTN in differential mode, $C_{REF} = 2.2\mu$ F, $C_{MICBIAS} = C_{PREG} = C_{REG} = 1\mu$ F, $AV_{PRE} = +20$ dB, $AV_{PGAM} = 0$ dB, $AV_{DAC} = 0$ dB, $AV_{LINE} = +20$ dB, $AV_{VOL} = 0$ dB, MCLK = 13MHz, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}$ C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	МАХ	UNITS
Minimum BCLK High Time	^t BCLKH	Slave operation		30		ns
Minimum BCLK Low Time	^t BCLKL	Slave operation		30		ns
BCLK or LRCLK Rise and Fall	t _R , t _F	Master operation, $C_L = 15 pF$		7		ns
SDIN or LRCLK to BCLK Setup Time	ts∪		20			ns
SDIN or LRCLK to BCLK Hold Time	thd		0			ns
SDOUT Delay Time from BCLK Rising Edge	tDLY	C _L = 30pF	0		40	ns
I ² C TIMING CHARACTERISTICS (V _{DVDD} = 1.6	5V)				
Serial-Clock Frequency	fscl		0		400	kHz
Bus Free Time Between STOP and START Conditions	^t BUF		1.3			μs
Hold Time (REPEATED) START Condition	^t HD,STA		0.6			μs
SCL Pulse-Width Low	tLOW		1.3			μs
SCL Pulse-Width High	thigh		0.6			μs
Setup Time for a REPEATED START Condition	^t SU,STA		0.6			μs
Data Hold Time	thd,dat	$R_{PU,SDA} = 475\Omega$	0		900	ns
Data Setup Time	tsu,dat		100			ns
SDA and SCL Receiving Rise Time	t _R	(Note 10)	20 + 0.1C _B		300	ns
SDA and SCL Receiving Fall Time	tF	(Note 10)	20 + 0.1C _B		300	ns
SDA Transmitting Fall Time	tF	$R_{PU,SDA} = 475\Omega$ (Note 10)	20 + 0.1C _B		250	ns
Setup Time for STOP Condition	tsu,sto		0.6			μs
Bus Capacitance	CB				400	pF
Pulse Width of Suppressed Spike	tsp		0		50	ns

Note 2: The MAX9867 is 100% production tested at $T_A = +25^{\circ}C$. Specifications over temperature limits are guaranteed by design.

Note 3: Clocking all zeros into the DAC, master mode, and differential headphone mode.

Note 4: DAC performance measured at the headphone outputs.

Note 5: Dynamic range measured using the EIAJ method. -60dBFS 1kHz output signal, A-weighted, and normalized to 0dBFS. f = 20Hz to 20kHz.

Note 6: Performance measured using microphone inputs, unless otherwise stated.

Note 7: Performance measured using line inputs.

Note 8: Performance measured using DAC, unless otherwise stated. LRCLK = 8kHz, unless otherwise stated.

Note 9: In master-mode operation, the accuracy of the MCLK input proportionally determines the accuracy of the sample clock rate.

Note 10: C_B is in pF.





_ 13

Typical Operating Characteristics

 $(V_{AVDD} = V_{DVDD} = V_{PVDD} = +1.8V, C_{REF} = 2.2\mu$ F, $C_{MICBIAS} = C_{PREG} = C_{REG} = 1\mu$ F, $AV_{MICPGA} = 0$ dB, MCLK = 13MHz, LRCLK = 8kHz, BW = 20Hz to fs/2, T_A = +25°C, unless otherwise noted.)

(HD+N (dB)



_Typical Operating Characteristics (continued)

 $(V_{AVDD} = V_{DVDD} = V_{PVDD} = +1.8V, C_{REF} = 2.2\mu$ F, $C_{MICBIAS} = C_{PREG} = C_{REG} = 1\mu$ F, $AV_{MICPGA} = 0$ dB, MCLK = 13MHz, LRCLK = 8kHz, BW = 20Hz to f_S/2, T_A = +25°C, unless otherwise noted.)

TOTAL HARMONIC DISTORTION + NOISE vs. Power out (dac to headphone)



TOTAL HARMONIC DISTORTION + NOISE vs. Power out (dac to headphone)



TOTAL HARMONIC DISTORTION + NOISE vs. Frequency (dac to headphone)





TOTAL HARMONIC DISTORTION + NOISE vs. Power out (dac to headphone)



TOTAL HARMONIC DISTORTION + NOISE vs. Power out (line in to headphone)



TOTAL HARMONIC DISTORTION + NOISE vs. Frequency (dac to headphone)



TOTAL HARMONIC DISTORTION + NOISE vs. Frequency (dac to headphone)



TOTAL HARMONIC DISTORTION + NOISE vs. Power out (line in to headphone)







Typical Operating Characteristics (continued) MAX986

M /X / M



(VAVDD = VDVDD = VPVDD = +1.8V, CREF = 2.2µF, CMICBIAS = CPREG = CREG = 1µF, AVMICPGA = 0dB, MCLK = 13MHz, LRCLK = 8kHz, BW = 20Hz to $f_S/2$, $T_A = +25^{\circ}C$, unless otherwise noted.)



/N/IXI/N

Typical Operating Characteristics (continued)



MAX986

Typical Operating Characteristics (continued)

 $(V_{AVDD} = V_{DVDD} = V_{PVDD} = +1.8V, C_{REF} = 2.2\mu$ F, $C_{MICBIAS} = C_{PREG} = C_{REG} = 1\mu$ F, $AV_{MICPGA} = 0$ dB, MCLK = 13MHz, LRCLK = 8kHz, BW = 20Hz to f_S/2, T_A = +25°C, unless otherwise noted.)





SCL (2V/div)

ADC OUT (500mV/div)

Typical Operating Characteristics (continued)

 $(V_{AVDD} = V_{DVDD} = V_{PVDD} = +1.8V, C_{REF} = 2.2\mu$ F, $C_{MICBIAS} = C_{PREG} = C_{REG} = 1\mu$ F, $AV_{MICPGA} = 0$ dB, MCLK = 13MHz, LRCLK = 8kHz, BW = 20Hz to f_S/2, T_A = +25°C, unless otherwise noted.)















TIME (4ms/div)



AUX CODE vs. INPUT VOLTAGE



____Pin Description

PIN/BUMP			EUNCTION				
TQFN-EP	WLP	NAME	FUNCTION				
1	A2	DGND	Digital Ground				
2	B3	SCL	I ² C Serial-Clock Input. Connect a pullup resistor to a 1.7V to 3.3V supply.				
3	A3	SDA	I ² C Serial-Data Input/Output. Connect a pullup resistor to a 1.7V to 3.3V supply.				
4	C3	ĪRQ	Hardware Interrupt Output. \overline{IRQ} can be programmed to pull low when bits in status register 0x00 are set. Read status register 0x00 to clear \overline{IRQ} once set. Repeat faults have no effect on \overline{IRQ} until it is cleared by reading register 0x00. Connect a 10k Ω pullup resistor to a 1.7V to 3.3V supply.				
5	A4	AVDD	Analog Power Supply. Bypass to AGND with a 1µF capacitor.				
6	B4	REF	Converter Reference. Bypass to AGND with a 2.2µF capacitor (1.23V nominal).				
7	A5	PREG	Positive Internal Regulated Supply. Bypass to AGND with a $1\mu F$ capacitor (1.6V nominal).				
8	B5	REG	PREG/2 Voltage Reference. Bypass to AGND with a 1µF capacitor (0.8V nominal).				
9	A6	AGND	Analog Ground				
10	B6	MICBIAS	Low-Noise Microphone Bias. Connect a $2.2k\Omega$ to 470Ω resistor to the positive output of a microphone (1.525V nominal). Bypass to AGND with a 1μ F capacitor.				
11	C5	MICLN/ DIGMICCLK	Left Negative Differential Microphone Input or Digital Microphone Clock Output. For analog microphones, AC-couple to the negative output of a microphone with a 1μ F capacitor. For digital microphones, connect to the clock input of the microphone.				
12	C6	MICLP/ DIGMICDATA	Left Positive Differential Microphone Input or Digital Microphone Data Input. For analog microphones, AC-couple to the positive output of a microphone with a 1μ F capacitor. For digital microphones, connect to the data output of the microphone(s). Up to two digital microphones can be connected.				
13	C4	MICRP	Right Positive Differential Microphone Input. AC-couple to the positive output of a microphone with a $1\mu F$ capacitor.				
14	D6	MICRN	Right Negative Differential Microphone Input. AC-couple to the negative output of a microphone with a 1μ F capacitor.				
15	D5	LINL	Left-Line Input. AC-couple analog audio signal to LINL with a 1µF capacitor.				
16	E6	LINR	Right-Line Input. AC-couple analog audio signal to LINR with a 1µF capacitor.				
17	D4	JACKSNS/AUX	Jack Sense or Auxiliary ADC Input. When configured for jack detection, JACKSNS detects the presence or absence of a jack. See the <i>Mode Configuration</i> section for details. When configured as an auxiliary ADC input, AUX is used to measure DC voltages.				
18	E5	PGND	Headphone Power Ground				
19	D3	ROUTP	Positive Right-Channel Headphone Output. Connect directly to the load in differential and capacitorless mode. AC-couple to the load in single-ended mode.				
20	E4	ROUTN	Negative Right-Channel Headphone Output. Inverting output in differential mode. Leave unconnected in capacitorless and fast turn-on single-ended mode. Bypass with a 1μ F capacitor to AGND in clickless, single-ended mode.				
21	D2	LOUTN	Negative Left-Channel Headphone Output. Noninverting output in differential mode. Common headphone return in capacitorless mode. Leave unconnected in fast turn-on single-ended mode. Bypass with a 1μ F capacitor to AGND in clickless single-ended mode.				

Master Clock Input. Acceptable input frequency range: 10MHz to 60MHz. Digital Power Supply. Supply for the digital circuitry and I²C interface. Bypass to

Exposed Pad. Connect the exposed thermal pad to AGND.

PIN/E	UMP		FUNCTION
TQFN-EP	WLP	NAME	FUNCTION
22	E3	LOUTP	Positive Left-Channel Headphone Output. Connect directly to the load in differential and capacitorless mode. AC-couple to the load in single-ended mode
23	E2	PVDD	Headphone Power Supply. Bypass to PGND with a 1µF capacitor.
24, 25		N.C.	No Connection
26	E1	DVDDIO	Digital Audio Interface Power Supply. Bypass to DGND with a 1µF capacitor.
27	D1	SDOUT	Digital Audio Serial-Data ADC Output
28	C2	SDIN	Digital Audio Serial-Data DAC Input
29	C1	LRCLK	Digital Audio Left-Right Clock Input/Output. LRCLK is the audio sample rate clock and determines whether the audio data on SDIN is routed to the left or right channel. In TDM mode, LRCLK is a frame synchronization pulse. LRCLK is an input when the MAX9867 is in slave mode and an output when in master mode.
30	B1	BCLK	Digital Audio Bit Clock Input/Output. BCLK is an input when the MAX9867 is in slave mode and an output when in master mode.

DGND with a 1µF capacitor.

Pin Description (continued)

Detailed Description

MCLK

DVDD

EΡ

The MAX9867 is a low-power stereo audio codec designed for portable applications requiring minimum power consumption.

B2

Α1

The stereo playback path accepts digital audio through a flexible interface compatible with I2S, TDM, and leftjustified signals. An oversampling sigma-delta DAC converts the incoming digital data stream to analog audio and outputs the audio through the stereo headphone amplifier. The headphone amplifier can be configured in differential, single-ended, and capacitorless output modes.

The stereo record path has two analog microphone inputs with selectable gain. An integrated microphone bias can be used to power the microphones. The left analog microphone inputs can also accept data from up to two digital microphones. An oversampling sigmadelta ADC converts the microphone signals and outputs the digital bit stream over the digital audio interface.

Integrated digital filtering provides a range of notch and highpass filters for both the playback and record paths to limit undesirable low-frequency signals and GSM transmission noise. The digital filtering provides attenuation of out-of-band energy by over 70dB, eliminating audible aliasing. A digital sidetone function allows audio from the record path to be summed into the plavback path after digital filtering.

The MAX9867 also includes two stereo, single-ended line inputs with gain adjustment, which can be recorded by the ADCs and/or output by the headphone amplifiers. An auxiliary ADC accurately measures a DC voltage by utilizing the right audio ADC and reporting the DC voltage through the I²C interface. A jack detection function allows the detection of headphone, microphone, and headset jacks. Insertion and removal events can be programmed to trigger a hardware interrupt and flag an I²C register bit.

The MAX9867's flexible clock circuitry utilizes a programmable clock divider and a digital PLL, allowing the DAC and ADC to operate at maximum dynamic range for all combinations of master clock (MCLK) and sample rate (LRCLK) without consuming extra supply current. Any master clock between 10MHz and 60MHz is supported as are all sample rates from 8kHz to 48kHz. Master and slave modes are supported for maximum flexibility.

MAX986

30 31

32

I²C Registers

I²C Slave Address

The MAX9867 responds to the slave address 0x30 for all write commands and 0x31 for all read operations.

Table 1. I²C Register Map

The MAX9867 audio codec is completely controlled

through software using an I²C interface. The power-on default setting is complete shutdown, requiring that the internal registers be programmed to activate the device. See Table 1 for the device's complete register map.

REGISTER	B7	B6	B5	В4	B3	B2	B1	В0	REGISTER ADDRESS	POWER- ON RESET STATE
STATUS										
Status (Read Only)	CLD	SLD	ULK	0	0	0	JDET	0	0x00	_
Jack Sense (Read Only)	LSNS	JKSNS	JKMIC	0	0	0	0	0	0x01	_
AUX High (Read Only)				AUX[15:8]				0x02	_
AUX Low (Read Only)				AUX	[7:0]				0x03	—
Interrupt Enable	ICLD	ISLD	IULK	0	0	SDODLY	IJDET	0	0x04	0x00
CLOCK CONTROL										
System Clock	0	0	PSC	CLK		FRI	EQ		0x05	0x00
Stereo Audio Clock Control High	PLL				NI[14:8]				0x06	0x00
Stereo Audio Clock Control Low				NI[7:1]				RLK/ NI[0]	0x07	0x00
DIGITAL AUDIO INTERF	ACE									
Interface Mode	MAS	WCI	BCI	DLY	HIZOFF	TDM	0	0	0x08	0x00
Interface Mode	0	0	0	LVOLFIX	DMONO		BSEL		0x09	0x00
DIGITAL FILTERING										
Codec Filters	MODE		AVFLT		0	DVFLT			0x0A	0x00
LEVEL CONTROL										
Sidetone	DS	TS	0		DVST			0x0B	0x00	
DAC Level	0	DACM	DA	CG		DA	СА		0x0C	0x00
ADC Level		A١	/L			A٧	/R		0x0D	0x00
Left-Line Input Level	0	LILM	0	0		LIC	GL		0x0E	0x00
Right-Line Input Level	0	LIRM	0	0		LIC	GR		0x0F	0x00
Left Volume Control	0	VOLLM			VC)LL			0x10	0x00
Right Volume Control	0	VOLRM			VC	LR			0x11	0x00
Left Microphone Gain	0	PAL	EN			PGAML			0x12	0x00
Right Microphone Gain	0	PAF	REN			PGAMR			0x13	0x00
CONFIGURATION										
ADC Input	MX	INL	MX	INR	AUXCAP	AUXGAIN	AUXCAL	AUXEN	0x14	0x00
Microphone	MIC	CLK	DIGMICL	DIGMICR	0	0	0	0	0x15	0x00
Mode	DSLEW	VSEN	ZDEN	0	JDETEN		HPMODE		0x16	0x00
POWER MANAGEMENT									r	
System Shutdown	SHDN	LNLEN	LNREN	0	DALEN	DAREN	ADLEN	ADREN	0x17	0x00
Revision				RE	EV				0xFF	0x42



Device Status

Status registers 0x00 and 0x01 are read-only registers that report the status of various device functions. The status register bits are cleared upon reading the status register and are set the next time the event occurs. Registers 0x02 and 0x03 report the DC level applied to AUX. See the *ADC* section for more details and Table 2.

Table 2. Status Registers

REGISTER	B7	B6	В5	B4	В3	B2	B1	В0	REGISTER ADDRESS	
Status (Read Only)	CLD	SLD	ULK	0	0	0	JDET	0	0x00	
Jack Sense (Read Only)	LSNS	JKSNS	JKMIC	0	0	0	0	0	0x01	
AUX High (Read Only)		AUX[15:8]								
AUX Low (Read Only)				AUX	[7:0]				0x03	

BITS	FUNCTION
CLD	Clip Detect Flag Indicates that a signal has reached or exceeded full scale in the ADC or DAC.
SLD	Slew Level Detect Flag When volume or gain changes are made, the slewing circuitry smoothly steps through all intermediate settings. When SLD is set high, all slewing has completed and the volume or gain is at its final value. SLD is also set when soft-start or stop is complete.
ULK	Digital PLL Unlock Flag Indicates that the digital audio PLL has become unlocked and digital signal data is not reliable.
JDET	Headset Configuration Change Flag JDET is set whenever there is a change in register 0x01, indicating that the headset configuration has changed.
LSNS	LOUTP State (Valid if SHDN = 0, JDETEN = 1) LSNS is set when the voltage at LOUTP exceeds AVDD - 0.4V. An internal pullup from AVDD to LOUTP causes this condition whenever there is no load on LOUTP. LSNS is only valid in differential and capacitorless output modes.
JKSNS	JACKSNS State (Valid if JDETEN = 1) JKSNS is set when the voltage at JACKSNS exceeds AVDD - 0.4V. An internal pullup from AVDD to JACKSNS causes this condition whenever there is no load on JACKSNS.
JKMIC	Microphone Detection (Valid if PALEN or PAREN \neq 00 and JDETEN = 1) JKMIC is set when JACKSNS exceeds 0.95 x V _{MICBIAS} .
AUX	Auxiliary Input MeasurementAUX is a 16-bit signed two's complement number representing the voltage measured at JACKSNS/AUX.Before reading a value from AUX, set AUXCAP to 1 to ensure a stable reading. After reading the value, set AUXCAP to 0.Use the following formula to convert the AUX value into an equivalent JACKSNS/AUX voltage:Voltage = $0.738V \times \left(\frac{AUX}{K}\right)$ K = AUX value when AUXGAIN = 1. See the ADC section for complete details.

Hardware Interrupts

Hardware interrupts are reported on the open-drain IRQ pin. When an interrupt occurs, IRQ remains low until the interrupt is serviced by reading the status register 0x00. If a flag is set, it is reported as a hardware interrupt only if the corresponding interrupt enable is set. Each bit enables interrupts for the status flag in the respective bit location in register 0x00. See Table 3.

SDODLY is used to control the SDOUT timing. See the *Digital Audio Interface* section for a detailed description.

Clock Control

The MAX9867 can work with a master clock (MCLK) supplied from any system clock within the 10MHz-to-60MHz range. Internally, the MAX9867 requires a 10MHz-to-20MHz clock. A prescaler divides MCLK by 1, 2, or 4 to create the internal clock (PCLK). PCLK is used to clock all portions of the MAX9867. See Table 4.

The MAX9867 is capable of supporting any sample rate from 8kHz to 48kHz, including all common sample rates (8kHz, 16kHz, 24kHz, 32kHz, 44.1kHz, and 48kHz). To

accommodate a wide range of system architectures, the MAX9867 supports three main clocking modes:

- **Normal:** This mode uses a 15-bit clock divider coefficient to set the sample rate relative to the prescaled MCLK input (PCLK). This allows high flexibility in both the MCLK and LRCLK frequencies and can be used in either master or slave mode.
- Exact Integer: In both master and slave mode, common MCLK frequencies (12MHz, 13MHz, 16MHz, and 19.2MHz) can be programmed to operate in exact integer mode for both 8kHz and 16kHz sample rates. In these modes, the MCLK and LRCLK rates are selected by using the FREQ bits instead of the NI and PLL control bits.
- PLL: When operating in slave mode, a PLL can be enabled to lock onto externally generated LRCLK signals that are not integer related to PCLK. Prior to enabling the interface, program NI to the nearest desired ratio and set the NI[0] = 1 to enable the PLL's rapid lock mode. If NI[0] = 0, then NI is ignored and PLL lock time is slower.

REGISTER	B7	B6	B5	B4	В3	B2	B1	В0	REGISTER ADDRESS
Interrupt Enable	ICLD	ISLD	IULK	0	0	SDODLY	IJDET	0	0x04

Table 3. Interrupt Registers

Table 4. Clock Control Registers

REGISTER	B7	B6	B5	B4	B3	B2	B1	B0	REGISTER
System Clock	0	0	PSC	PSCLK FREQ				0x05	
Stereo Audio Clock Control High	PLL				NI[14:8]		0x06		
Stereo Audio Clock Control Low			NI[7:1] NI[0]						

BITS	FUNCTION
PSCLK	MCLK PrescalerDivides MCLK to generate a PCLK between 10MHz and 20MHz.00 = Disable clock for low-power shutdown.01 = Select if MCLK is between 10MHz and 20MHz.10 = Select if MCLK is between 20MHz and 40MHz.11 = Select if MCLK is between 40MHz and 60MHz.

Table 4. Clock Control Registers (continued)

BITS	FUNCTION									
	Exact Integer Modes									
	Allows integer sampling for s	pecific PCLK (prescaled M	CLK) frequencies and 8kHz c	or 16kHz sample rates.						
	FREQ[3:0]	PCLK (MHz)	LRCLK (kHz)	PCLK/LRCLK						
	0x00		Normal or PLL mode							
	0x1–0x7	Reserved	Reserved	Reserved						
	0x8	12	8	1500						
	0x9	12	16	750						
FREQ	0xA	13	8	1625						
	0xB	13	16	812.5						
	0xC	16	8	2000						
	0xD	16	16	1000						
	0xE	19.2	8	2400						
	0xF	19.2	16	1200						
	Modes 0x8–0xF are available	Modes UX8-UXF are available in either master or slave mode. In slave mode, if the indicated PCLK/LRCLK ratio								
	cannot be guaranteed, use PLL mode instead.									
PLL	 0 = Valid for slave and master MAX9867 generates LRC LRCLK as specified by the specified by the slave mode only Rapid Lock Mode To enable rapid lock mode. 	er mode. The frequency of L CLK using the specified divi- ne divide ratio. 7. A digital PLL locks on to a set NI to the nearest desired	RCLK is set by the NI divider de ratio. In slave mode, the M any externally supplied LRCLk d ratio and set NI[0] = 1 befor	bits. In master mode, the IAX9867 expects an < signal. e enabling the interface.						
	Normal Mode LRCLK Divid	er								
NI	When PLL = 0, the frequency of LRCLK is determined by NI. See Table 5 for common NI values. NI = $(65536 \times 96 \times f_{LRCLK})/f_{PCLK}$ $f_{LRCLK} = LRCLK$ frequency $f_{PCLK} = Prescaled$ MCLK internal clock frequency (PCLK) LRCLK > 24kHz is only valid for MODE = 0 (stereo audio mode). MODE = 1 (voice mode) requires LRCLK ≤ 24kHz.									

Table 5. Common NI Values

				LRCLK (kHz)			
	PSCLK	8	16	24	32	44.1	48
11.2896	01	0x116A	0x22D4	0x343F	0x45A9	0x6000	0x687D
12	01	0x1062	0x20C5	0x3127	0x4189	0x5A51	0x624E
12.288	01	0x1000	0x2000	0x3000	0x4000	0x5833	0x6000
13	01	0x0F20	0x1E3F	0x2D5F	0x3C7F	0x535F	0x5ABE
19.2	01	0x0A3D	0x147B	0x1EB8	0x28F6	0x3873	0x3D71
24	10	0x1062	0x20C5	0x1893	0x4189	0x5A51	0x624E
26	10	0x0F20	0x1E3F	0x16AF	0x3C7F	0x535F	0x5ABE
27	10	0x0E90	0x1D21	0x15D8	0x3A41	0x5048	0x5762

Note: Bolded values are exact integers that provide maximum full-scale performance.

MAX9867

Digital Audio Interface

The MAX9867's digital audio interface supports a wide range of operating modes to ensure maximum compatibility. See Figures 1–4 for timing diagrams. In master mode, the MAX9867 outputs LRCLK and BCLK, while in slave mode they are inputs. When operating in master mode, BCLK can be configured in a number of ways to ensure compatiblity with other audio devices.

LVOLFIX is used to fix the line input playback volume to 0dB regardless of VOLL and VOLR. See the *Line Inputs* section for complete details and Table 6.

Table 6. Digital Audio Interface Registers

REGISTER	B7	B6	B5	B4	В3	B2	B1	В0	REGISTER ADDRESS
Interface Mode	MAS	WCI	BCI	DLY	HIZOFF	TDM	0	0	0x08
Interface Mode	0	0	0	LVOLFIX	DMONO		BSEL		0x09

BITS	FUNCTION
	Master Mode
MAS	0 = The MAX9867 operates in slave mode with LRCLK and BCLK configured as inputs.
	1 = The MAX9867 operates in master mode with LRCLK and BCLK configured as outputs.
	LRCLK Invert
WCI	0 = Left-channel data is input and output while LRCLK is low.
WOI	1 = Right-channel data is input and output while LRCLK is low.
	Note: WCI is ignored when TDM = 1.
	BCLK Invert
	In master and slave modes:
	0 = SDIN is latched into the part on the rising edge of BCLK.
5.01	SDOUT transitions after the rising edge of BCLK as determined by SDODLY.
BCI	1 = SDIN is latched into the part on the falling edge of BCLK.
	SDOUT transitions after the falling edge of BCLK as determined by SDODLY.
	In master mode:
	0 = LRCLK changes state immediately after the rising edge of BCLK.
	I = LRULK changes state immediately after the failing edge of BULK.
	SDOUT belay
SDODLY	U = SDOUT transitions one hall bound cycle after SDIN is latched into the part.
	T = SDOOT (transitions on the same BCLK edge as SDIN is fatched into the part.
	See Figures 1–4 for complete details. See Register 0x04 (interrupt registers).
	Delay mode 0 - SDIN/SDOLIT data is latebod on the first BCLK edge following an LPCLK edge
	1 = SDIN/SDOUT data is accurate to be deleved one BCLK cycle so that it is latehold on the 2nd BCLK edge.
DET	following an LRCLK edge (I ² S-compatible mode)
	Note: DLY is ignored when TDM -1
	SDOUT High-Impedance Mode
	$\Omega = SDOLIT$ goes to a high-impedance state after all data bits have been transferred out of the MAX9867
HIZOFE	allowing SDOUT to be shared by other devices
	1 = SDOUT is set either high or low after all data bits have been transferred out of the MAX9867
	Note: High-impedance mode is intended for use when TDM = 1.
LVOLFIX	See the Line Inputs section.

Table 6. Digital Audio Interface Registers (continued)

BITS	FUNCTION
TDM	 TDM Mode Select 0 = LRCLK signal polarity indicates left and right audio. 1 = LRCLK is a framing pulse that transitions polarity to indicate the start of a frame of audio data consisting of multiple channels. When operating in TDM mode, the left channel is output immediately following the frame sync pulse. If right-channel data is being transmitted, the 2nd channel of data immediately follows the 1st channel data.
DMONO	 Mono Playback Mode 0 = Stereo data input on SDIN is processed separately. 1 = Stereo data input on SDIN is mixed to a single channel and routed to both the left and right DAC.
BSEL	BCLK Select Configures BCLK when operating in master mode. BSEL has no effect in slave mode. Set BSEL = 010, unless sharing the bus with multiple devices: 000 = Off 001 = 64x LRCLK (192x internal clock divided by 3) 010 = 48x LRCLK (192x internal clock divided by 4) 011 = Reserved for future use. 100 = PCLK/2 101 = PCLK/4 110 = PCLK/8 111 = PCLK/16





M/IXI/M



Figure 1. Digital Audio Interface Audio Master Mode Example (Sheet 2 of 2)

VOICE (TDM, PCM) MASTER MODES: BCI = 0, HIZOFF = 0, SDODLY = 0 - 7ns (typ) - 7ns (typ) I RCI K 1/fs RELATIVE TO PCLK (SEE NOTE) XL15XL14XL13XL12XL11XL10XL9XL8XL7XL6XL5XL4XL3XL2XL1XL0XR15XR14XR13XR12XR11XR10XR9XR8XR7XR6XR5XR4XR3XR2XR1XR0//// '//X// SDOUT 40ns (max) Ons (min) ► | < - 7ns (typ) BCI K - CONFIGURED BY BSEL 25ns (min) - Ons (min) 🧤 χμτ5 χμτ4 χμτ3 χμτ2 χμτ1 χμτ0 χμ9 χμ8 χμτ7 χμ6 χμ5 χμ4 χμ3 χμ2 χμτ χμ0 χμτ5 χμτ4 χμτ3 χμτ2 χμτ1 χμτ0 χμ7 χμ6 χμ5 χμ4 χμ3 χμ2 χμτ1 χμ0 SDIN BCI = 1, HIZOFF = 0, SDODLY = 0 → - 7ns (typ) ► - 7ns (typ) I RCI K - 1/fs RELATIVE TO PCLK (SEE NOTE) SDOUT 40ns (max) Ons (min) ► - 7ns (typ) BCLK CONFIGURED BY BSEL 25ns (min) - Ons (min) SDIN BCI = 0, HIZOFF = 1, SDODLY = 0 ► - 7ns (typ) 7ns (typ) I RCI K 1/fs RELATIVE TO PCLK (SEE NOTE) XL15XL14XL13XL12XL11XL10XL9XL8XL7XL6XL5XL4XL3XL2XL1XL0XR15XR14XR13XR12XR11XR10XR9XR8XR7XR6XR5XR4XR3XR2XR1XR0 SDOUT 40ns (max) Ons (min) -**►**I**-** 7ns (typ) BCI K - CONFIGURED BY BSEI 25ns (min) - Ons (min) SDIN 🧤 🕼 μτο μια ματαγματικά ματαγμα BCI = 0, HIZOFF = 0, SDODLY = 1 🗲 7ns (typ) ► - 7ns (typ) I RCI K - 1/fs -RELATIVE TO PCLK (SEE NOTE) SDOUT 40ns (max) -Ons (min) ► I < - 7ns (typ) BCI K - CONFIGURED BY BSEL --**- -**- Ons (min) 25ns (min) -> 🧤 🕼 ματαγματικά SDIN



/N/IXI/N

AX9867



Figure 3. Digital Audio Interface Audio Slave Mode Examples (Sheet 1 of 2)



Figure 3. Digital Audio Interface Audio Slave Mode Examples (Sheet 2 of 2)



Figure 4. Digital Audio Interface Voice Slave Mode Examples

MAX9867

Digital Filtering

The MAX9867 incorporates both IIR (voice) and FIR (audio) digital filters to accomodate a wide range of audio sources. The IIR fiilters provide over 70dB of

stopband attenuation as well as selectable highpass filters. The FIR filters provide low-power consumption and are linear phase to maintain stereo imaging. Table 7 is the digital filtering register.

Table 7. Digital Filtering Register

REGISTER	B7	B6	B 5	B4	B3	B2	B1	В0	REGISTER ADDRESS
Codec Filters	MODE		AVFLT		0		DVFLT		0x0A

BITS	FUNCTION
MODE	Digital Audio Filter Mode 0 = IIR Voice Filters 1 = FIR Audio Filters
AVFLT	ADC Digital Audio Filter MODE = 0 Select the desired digital filter response from Table 8. See the Frequency Response graph in the <i>Typical</i> <i>Operating Characteristics</i> section for details on each filter. MODE = 1 0x0 = DC-blocking filter is disabled. Any other setting = DC-blocking filter is enabled.
DVFLT	DAC Digital Audio FilterMODE = 0Select the desired digital filter response from Table 8. See the Frequency Response graph in the TypicalOperating Characteristics section for details on each filter.MODE = 10x0 = DC-blocking filter is disabled. Any other setting = DC-blocking filter is enabled.

Table 8. IIR Highpass Digital Filters

CODE	FILTER TYPE	INTENDED SAMPLE RATE (kHz)	HIGHPASS CORNER FREQUENCY (Hz)	217Hz NOTCH
0x0		Disa	bled	
0x1	Elliptical	16	256	Yes
0x2	Butterworth	16	500	No
0x3	Elliptical	8	256	Yes
0x4	Butterworth	8	500	No
0x5	Butterworth	8 to 24	f _S /240	No
0x6 to 0x7		Rese	erved	

Digital Gain Control

The MAX9867 includes digital gain adjustment for the playback and record paths. Independent gain adjustment is provided for the two record channels. Sidetone

Table 9. Digital Gain Registers

gain adjustment is also provided to set the sidetone level relative to the playback level. Table 9 is the digital gain registers.

REGISTER	B7	B6	B5	B4	В3	B2	B1	В0	REGISTER ADDRESS	
Sidetone	DS	STS	0			0x0B				
DAC Level	0	DACM	DA	DACG		DACA				
ADC Level		A١	/L		AVR				0x0D	

BITS	FUNCTION										
DSTS	Digital Sidetone 00 = No sid 01 = Left AE 10 = Right A 11 = Left +	Digital Sidetone Source Mixer00=01=Left ADC10=Right ADC11=Left + right ADC									
	Digital Sidetone All gain settings Differential Hea	E Level Control are relative to the A dphone Output M	ADC input voltage ode								
	SETTING	GAIN (dB)	SETTING	GAIN (dB)	SETTING	GAIN (dB)					
	0x00	Off	0x0B	-20	0x16	-42					
	0x01	0	0x0C	-22	0x17	-44					
	0x02	-2	0x0D	-24	0x18	-46					
	0x03	-4	0x0E	-26	0x19	-48					
	0x04	-6	0x0F	-28	0x1A	-50					
	0x05	-8	0x10	-30	0x1B	-52					
	0x06	-10	0x11	-32	0x1C	-54					
	0x07	-12	0x12	-34	0x1D	-56					
	0x08	-14	0x13	-36	0x1E	-58					
DVST	0x09	-16	0x14	-38	0x1F	-60					
5101	0x0A	-18	0x15	-40		—					
	Capacitorless a	nd Single-Ended	Headphone Outp	out Mode							
	SETTING	GAIN (dB)	SETTING	GAIN (dB)	SETTING	GAIN (dB)					
	0x00	Off	0x0B	-25	0x16	-47					
	0x01	-5	0x0C	-27	0x17	-49					
	0x02	-7	0x0D	-29	0x18	-51					
	0x03	-9	0x0E	-31	0x19	-53					
	0x04	-11	0x0F	-33	0x1A	-55					
	0x05	-13	0x10	-35	0x1B	-57					
	0x06	-15	0x11	-37	0x1C	-59					
	0x07	-17	0x12	-39	0x1D	-61					
	0x08	-19	0x13	-41	0x1E	-63					
	0x09	-21	0x14	-43	0x1F	-65					
	0x0A	-23	0x15	-45		_					
DACM	DAC Mute Enab 0 = No mute 1 = Mute	le									

Table 9. Digital Gain Registers (continued)

MAX9867

BITS		FUN	CTION	
DACG	DAC Gain 00 = 0dB 01 = +6dB 10 = +12dB 11 = +18dB Note: DACG is only used DAC Level Control	when MODE = 0. If MODE	= 1, the DAC level is only se	et by DACA.
	SETTING	GAIN (dB)	SETTING	GAIN (dB)
				-8
	0x1	-1	0x9	-9
DACA	0x2	-2	0xA	-10
	0x3	-3	0xB	-11
	0x4	-4	0xC	-12
	0x5	-5	0xD	-13
	0x6	-6	0xE	-14
	0x7	-7	0xF	-15
	ADC Left/Right Level Co	ntrol	+	
	SETTING	GAIN (dB)	SETTING	GAIN (dB)
	0x0	+3	0x8	-5
	0x1	+2	0x9	-6
	0x2	+1	0xA	-7
	0x3	0	0xB	-8
	0x4	-1	0xC	-9
	0x5	-2	0xD	-10
	0x6	-3	0xE	-11
	0x7	-4	0xF	-12

Line Inputs

The MAX9867 includes one pair of single-ended line inputs. When enabled, the line inputs connect directly

Table 10. Line Input Registers

to the headphone amplifier and can be optionally connected to the ADC for recording. Table 10 lists the line input registers.

REGISTER	B7	B6	B5	B4	B3	B2	B1	В0	REGISTER ADDRESS
Left-Line Input Level	0	LILM	0	0	LIGL				0x0E
Right-Line Input Level	0	LIRM	0	0		LIC	GR		0x0F

Table 10. Line Input Registers (continued)

BITS		FUNC	TION						
LILM/LIRM	Line-Input Left/Right Playback Mute 0 = Line input is connected to the headphone amplifiers. 1 = Line input is disconnected from the headphone amplifiers.								
	Line-Input Left/Right Gai	n							
	SETTING	GAIN (dB)	SETTING	GAIN (dB)					
	0x0	+24	0x8	+8					
	0x1	+22	0x9	+6					
	0x2	+20	0xA	+4					
LIGL/LIGR	0x3	+18	0xB	+2					
	0x4	+16	0xC	0					
	0x5	+14	0xD	-2					
	0x6	+12	0xE	-4					
	0x7	+10	0xF	-6					
LVOLFIX	Fix Line Input Volume 0 = Line input to headphone output volume tracks VOLL and VOLR bits. 1 = Line input to headphone output volume fixed at VOLL and VOLR bits. See the Digital Audio Interface section.								

MAX9867

Playback Volume

The MAX9867 incorporates volume and mute control to allow level control for the playback audio path. Program

registers 0x10 and 0x11 to set the desired volume. See Table 11.

Table 11. Playback Volume Registers

REGISTER	B7	B6	В5	В4	В3	B2	B1	В0	REGISTER ADDRESS
Left Volume Control	0	VOLLM		VOLL					
Right Volume Control	0	VOLRM			VC	LR			0x11

Table 11. Playback Volume Registers (continued)

MAX9867

BITS	FUNCTION									
VOLLM/VOLRM	Left/Right Playb VOLLM and VOL 0 = Audio playba 1 = Audio playba Note: VSEN has immediately (ZD	VOLLM and VOLRM mute both the DAC and line input audio signals. 0 = Audio playback is unmuted. 1 = Audio playback is muted Note: VSEN has no effect on the mute function. When VOLLM or VOLRM is set, the output is muted immediately (ZDEN = 1) or at the next zero-crossing (ZDEN = 0).								
	Left/Right Playb VOLL and VOLR	eack Volume	ack volume for bot	h the DAC and line	e input audio signa	ls.				
	SETTING	GAIN (dB)	SETTING	GAIN (dB)	SETTING	GAIN (dB)				
	0x00	+6	0x0E	-5	0x1C	-42				
	0x01	+5.5	0x0F	-6	0x1D	-46				
	0x02	+5	0x10	-8	0x1E	-50				
	0x03	+4.5	0x11	-10	0x1F	-54				
	0x04	+4	0x12	-12	0x20	-58				
	0x05	+3.5	0x13	-14	0x21	-62				
VOLL/VOLR	0x06	+3	0x14	-16	0x22	-66				
	0x07	+2	0x15	-18	0x23	-70				
	0x08	+1	0x16	-20	0x24	-74				
	0x09	0	0x17	-22	0x25	-78				
	0x0A	-1	0x18	-26	0x26	-82				
	0x0B	-2	0x19	-30	0x27	-84				
	0x0C	-3	0x1A	-34	0,229 to 0,225					
	0x0D	-4	0x1B	-38	0x28 10 0x3F	MUTE				
	Note: Gain settin ended and capa	igs apply when the citorless modes, th	e headphone amp ne actual gain is 5	lifier is configured dB lower for each	in differential mode setting.	e. In the single-				

Microphone Inputs

Two differential microphone inputs and a low-noise microphone bias for powering the microphones are provided by the MAX9867. In typical applications, the left microphone records a voice signal and the right microphone records a background noise signal. In applications that require only one microphone, use the left microphone input and disable the right ADC. The microphone signals are amplified by two stages of gain and then routed to the ADCs. The first stage offers selectable 0dB, 20dB, or 30dB settings. The second stage is a programmable gain amplifier (PGA) adjustable from 0dB to 20dB in 1dB steps. Zero-crossing detection is included on the PGA to minimize zipper noise while making gain changes. See Figure 5 for a detailed diagram of the microphone input structure. Table 12 is the microphone input register.

Table 12. Microphone Input Register

REGISTER	B7	B6	B5	B4	B3	B2	B1	В0	REGISTER ADDRESS	
Left Microphone Gain	0	PA	PALEN		PGAML					
Right Microphone Gain	0	PAI	PAREN		PGAMR					

Table 12. Microphone Input Register (continued)

BITS		FUNC	TION								
PALEN/PAREN	Left/Right Microphone Preamplifier Gain Enables the microphone circuitry and sets the preamplifier gain. 00 = Disabled 01 = 0dB 10 = +20dB 11 = +30dB										
	Left/Right Microphone F	Left/Right Microphone Programmable Gain Amplifier									
	SETTING	GAIN (dB)	SETTING	GAIN (dB)							
	0x00	+20	0x0B	+9							
	0x01	0x01 +19		+8							
	0x02	+18	0x0D	+7							
	0x03	+17	0x0E	+6							
PGAML/PGAMR	0x04	+16	0x0F	+5							
	0x05	+15	0x10	+4							
	0x06	+14	0x11	+3							
	0x07	+13	0x12	+2							
	0x08	0x08 +12		+1							
	0x09	+10	0x14								
	0x0A	+11	to 0x1F	0							



Figure 5. Microphone Input Signal Path

ADC

The MAX9867 includes two 16-bit ADCs. The first ADC is used to record left-channel microphone and line-input audio signals. The second ADC can be used to record right-channel microphone and line-input signals, or it can be configured to accurately measure DC voltages.

When measuring DC voltages, both the left and right ADCs must be enabled by setting ADLEN and ADREN in register 0x17. The input to the second ADC is JACK-SNS/AUX and the output is reported in AUX (registers 0x02 and 0x03). Since the audio ADC is used to perform the measurement, the digital audio interface must be properly configured. If the left ADC is being used to convert audio, the DC measurement is performed at the same sample rate. When not using the left ADC, configure the digital interface for a 48kHz sample rate to ensure the fastest possible settling time.

To ensure accurate results, the MAX9867 includes two calibration routines. Calibrate the ADC each time the MAX9867 is powered on. Calibration settings are not lost if the MAX9867 is placed in shutdown. When making a measurement, set AUXCAP to 1 to prevent AUX from changing while reading the registers.

Setup Procedure

- 1) Ensure a valid MCLK signal is provided and configure PSCLK appropriately.
- 2) Choose a clocking mode. The following options are possible:
 - Slave mode with LRCLK and BCLK signals provided. The measurement sample rate is determined by the external clocks.
 - Slave mode with no LRCLK and BCLK signals provided. Configure the device for normal clock mode using the NI ratio. Select f_S = 48kHz to allow for the fastest settling times.
 - Master mode with audio. Configure the device in normal mode using the NI ratio or exact integer mode using FREQ as required by the audio signal.
 - Master mode without audio. Configure the device in normal mode using the NI ratio. Select f_S = 48kHz to allow for the fastest settling times.
- 3) Ensure JACKSNS is disabled.
- 4) Enable the left and right ADC; take the MAX9867 out of shutdown.

Offset Calibration Procedure

Perform the following steps before the first DC measurement is taken after applying power to the MAX9867:

Table 13. AUX ADC Wait Times

WAIT TIMES								
LRCLK (kHz)	WAIT TIME (ms)							
48	40							
44.1	44							
32	60							
24	80							
22.05	90							
16	120							
12	160							
11.025	175							
8	240							

- 1) Enable the AUX input (AUXEN = 1).
- 2) Enable the offset calibration (AUXCAL = 1).
- 3) Wait the appropriate time (see Table 13).
- 4) Complete calibration (AUXCAL = 0).

Gain Calibration Procedure

Perform the following steps the first time a DC measurement is taken after applying power to the MAX9867 or if the temperature changes significantly:

- 1) Enable the AUX input (AUXEN = 1).
- 2) Start gain calibration (AUXGAIN = 1).
- 3) Wait the appropriate time (see Table 13).
- 4) Freeze the measurement results (AUXCAP = 1).
- 5) Read AUX and store the value in memory to correct all future measurements (k = (AUX[15:0], k is typically 19500).
- 6) Complete calibration (AUXGAIN = AUXCAP = 0).

DC Measurement Procedure

Perform the following steps after offset and gain calibration are complete:

- 1) Enable the AUX input (AUXEN = 1).
- 2) Wait the appropriate time (see Table 13).
- 3) Freeze the measurement results (AUXCAP = 1).
- 4) Read AUX and correct with the gain calibration value:

$$V_{AUX} = 0.738 \left(\frac{AUX[15:0]}{k}\right)$$

5) Complete measurement (AUXCAP = 0).



Complete DC Measurement Example

MCLK = 13MHz, slave mode, BCLK and LRCLK not externally supplied:

- 1) Configure the digital audio interface for $f_S = 48$ kHz (PSCLK = 01, FREQ = 0x0, PLL = 0, NI = 0x5ABE, MAS = 0).
- 2) Disable JACKSNS (JDETEN = 0).
- 3) Enable the left and right ADC; take the MAX9867 out of shutdown (ADLEN = ADREN = SHDN = 1).
- 4) Calibrate the offset:
 - a. Enable the AUX input (AUXEN = 1).
 - b. Enable the offset calibration (AUXCAL = 1).
 - c. Wait 40ms.
 - d. Complete calibration (AUXCAL = 0).
- 5) Calibrate the gain:
 - a. Start gain calibration (AUXGAIN = 1).

- b. Wait 40ms.
- c. Freeze the measurement results (AUXCAP = 1).
- Read AUX and store the value in memory to correct all future measurements (k = (AUX[15:0]).
- e. Complete calibration (AUXGAIN = AUXCAP = AUXEN = 0).
- 6) Measure the voltage on JACKSNS/AUX:
 - a. Enable the AUX input (AUXEN = 1).
 - b. Wait 40ms.
 - c. Freeze the measurement results (AUXCAP = 1).
 - d. Read AUX and correct with the gain calibration value.
 - e. Complete measurement (AUXCAP = 0).
- 7) DC measurement complete.

Table 14. ADC Input Register

REGISTER	B7	B6	B5	B4	B3	B2	B1	В0	REGISTER ADDRESS
ADC Input	MX	INL	MX	INR	AUXCAP	AUXGAIN	AUXCAL	AUXEN	0x14

BITS	FUNCTION
MXINL/MXINR	Left/Right ADC Audio Input Mixer 00 = No input is selected. 01 = Left/right analog microphone 10 = Left/right line input 11 = Left/right analog microphone + line input Note: If the right-line input is disabled, then the left-line input is connected to both mixers. Enabling the left and right digital microphones disables the left and right audio mixers, respectively. See DIGMICL/ DIGMICR in Table 15 for more details.
AUXCAP	Auxiliary Input Capture 0 = Update AUX with the voltage at JACKSNS/AUX. 1 = Hold AUX for reading.
AUXGAIN	 Auxiliary Input Gain Calibration 0 = Normal operation 1 = The input buffer is disconnected from JACKSNS/AUX and connected to an internal voltage reference. While in this mode, read the AUX register and store the value. Use the stored value as a gain calibration factor, K, on subsequent readings.
AUXCAL	 Auxiliary Input Offset Calibration 0 = Normal operation 1 = JACKSNS/AUX is disconnected from the input and the ADC automatically calibrates out any internal offsets.
AUXEN	Auxiliary Input Enable 0 = Use JACKSNS/AUX for jack detection. 1 = Use JACKSNS/AUX for DC measurements. Note: For AUXEN = 1, set MXINR = 00, ADLEN = 1, and ADREN = 1.



MAX9867

Digital Microphone Input

The MAX9867 can accept audio from up to two digital microphones. When using digital microphones, the left analog microphone input is retasked as a digital micro-

phone input. The right analog microphone input is still available to allow a combination of analog and digital microphones to be used. Figure 6 shows the digital microphone interface timing diagram. See Table 15.

Table 15. Digital Microphone Input Register

REGISTER	B7	B6	B5	B4	B3	B2	B1	B0	REGISTER ADDRESS
Microphone	MICCLK		DIGMICL	DIGMICR	0	0	0	0	0x15

BITS		FU	NCTION							
	Digital Microphone Clock	K								
	00 = PCLK/8									
MICCLK	01 = PCLK/6									
	10 = Reserved									
	11 = Reserved									
	Digital Left/Right Microphone Enable									
	DIGMICL	DIGMICR	Left ADC Input	Right ADC Input						
	0	0	ADC input mixer	ADC input mixer						
DIGMICL/DIGMICR	0	1	Line input (left analog microphone unavailable)	Right digital microphone						
	1	0	Left digital microphone	ADC input mixer						
	1	1	Left digital microphone	Right digital microphone						
	Note: The left analog micro	ophone input is never av	ailable when DIGMICL or DIG	AMICR = 1.						



Figure 6. Digital Microphone Timing Diagram

Mode Configuration

The MAX9867 includes circuitry to minimize click-andpop during volume changes, detect headsets, and configure the headphone amplifier mode. Both volume slewing and zero-crossing detection are included to ensure click-and-pop free volume transitions. Table 16 is the mode configuration register.

Headset Detection Overview

The MAX9867 features headset detection that can detect the insertion and removal of a jack as well as the load type. When a jack is detected, an interrupt on \overline{IRQ} can be triggered to alert the microcontroller of the event. Figure 7 shows the typical configuration for jack detection.

Sleep-Mode Headset Detection

When the MAX9867 is in shutdown and the power supply is available, sleep-mode headset detection can be enabled to detect jack insertion. Sleep mode applies a 4µA pullup current to JACKSNS/AUX and LOUTP that forces the voltage on JACKSNS/AUX and LOUTP to AVDD when no load is applied. When a jack is inserted, either JACKSNS, LOUTP (assuming the headphone amplifier is not configured in single-ended mode), or both are loaded sufficiently to reduce the output voltage to nearly 0V and clear the JKSNS or LSNS bits, respectively. The change in the LSNS and JKSNS bits sets JDET and triggers an interrupt on \overline{IRQ} if IJDET is set. The interrupt signals the microcontroller that a jack has been inserted, allowing the microcontroller to respond as desired.

Powered-On Headset Detection

When the MAX9867 is in normal operation and the microphone interface is enabled, jack insertion and removal can be detected through the JACKSNS/AUX pin. As shown in Figure 7, V_{MIC} is pulled up by MICBIAS. When a microphone is connected, V_{MIC} is assumed to be between 0V and 95% of V_{MICBIAS}. If the jack is removed, V_{MIC} increases to V_{MICBIAS}. This event causes JKMIC to be set, alerting the system that the headset has been removed. Alternatively, if the jack is inserted, V_{MIC} decreases to below 95% of V_{MICBIAS} and JKMIC is cleared, alerting the system that a jack has been inserted. The JKMIC bit can be configured to create a hardware interrupt that alerts the microcontroller of jack removal and insertion events.

Headphone Modes

The headphone amplifier supports differential, singleended, and capacitorless output modes, as shown in Figure 8. In each mode, the amplifier can be configured for stereo or mono operation. The differential and capacitorless modes are inherently click and pop free. The single-ended mode optionally includes click-andpop reduction to eliminate the click and pop that would normally be caused by the output coupling capacitor. When click-and-pop reduction is not required in the single-ended configuration, leave LOUTN and ROUTN unconnected.



Figure 7. Typical Configuration for Headset Detection



Figure 8. Headphone Amplifier Modes



43

REGISTER	B7	B6	B5	В4	В3	B2	B1	В0	REGISTER ADDRESS			
Mode	DSLEW	VSEN	ZDEN	0	JDETEN		HPMODE		0x16			
BITS					FUNCTION	N						
DSLEW	Digital Vo 0 = Digital 1 = Digital	lume Slew volume cha volume cha	Speed anges are sle anges are sle	ewed over * ewed over {	10ms. 30ms.							
VSEN	Volume C 0 = Volume 1 = Volume	Volume Change Smoothing 0 = Volume changes slew through all intermediate values. 1 = Volume changes occur in one step.										
ZDEN	Line Input 0 = Line in crossir 1 = Line-in	t Zero-Cros put volume ng occurs. Iput volume	changes oc	cur at zero	crossings in iately.	the audio v	waveform or	after 62ms	if no zero			
JDETEN	Jack Dete SHDN = 0: Enables pu LOUTP de SHDN = 1: Enables th microphon Note: AUX	ction Enab Sleep Moo ullups on LC tection is or Normal M e comparat e circuitry is EN must be	le de DUTP and JA nly valid in d ode cor circuitry c s enabled. e set to 0 for	ACKSNS/AU ifferential a on JACKSN jack detect	JX to detect j nd capacitorl S/AUX to det tion to functio	ack inserti ess output ect voltage n.	on. LSNS and modes. e changes. Jl	d JKSNS ar KMIC is val	e valid. id if the			
	Headphon	ne Amplifie	r Mode									
		HPMO	DE				Mode					
		000				Stereo d	ifferential (cli	ckless)				
		001			I	Mono (left)	differential (clickless)				
		010				Stereo cap	pacitorless (c	clickless)				
HPMODE		011			М	ono (left) c	apacitorless	(clickless)				
		100 Stereo single-ended (clickless)										
		101			М	ono (left) s	single-ended	(clickless)				
		110			S	Stereo sing	le-ended (fas	st turn-on)				
		111			Мо	no (left) sir	ngle-ended (fast turn-on)			
	Note: In m	iono operati	ion, the right	amplifier is	disabled.							

Table 16. Mode Configuration Register

MAX9867

power-management register.

required circuitry is active. Toggle the SHDN bit when-

ever a configuration change is made. Table 17 is the

Power Management

The MAX9867 includes complete power management control to minimize power usage. The DAC and both ADC can be independently enabled so that only the

Table 17. Power-Management Register

REGISTER	B7	B6	B5	B4	В3	B2	B1	В0	REGISTER ADDRESS
System Shutdown	SHDN	LNLEN	LNREN	0	DALEN	DAREN	ADLEN	ADREN	0x17

BITS	FUNCTION
SHDN	Shutdown
	Places the device in low-power shutdown mode.
	Left-Line Input Enable
	Enables the left-line input preamp and automatically enables the left and right headphone amplifiers.
LNLEN	If LNREN = 0, the left-line input signal is also routed to the right ADC input mixer and right headphone
	amplifier.
	Note: Control of the right headphone amplifier can be overridden by HPMODE.
	Right-Line Input Enable
LNREN	Enables the right-line input preamp and automatically enables the right headphone amplifier.
	Note: Control of the right headphone amplifier can be overridden by HPMODE.
	Left DAC Enable
	Enables the left DAC and automatically enables the left and right headphone amplifiers. If DAREN = 0, the
DALEN	left DAC signal is also routed to the right headphone amplifier.
	Note: Control of the right headphone amplifier can be overridden by HPMODE.
	Right DAC Enable
DAREN	Enabling the right DAC must be done in the same I ² C write operation that enables the left DAC. Right
	DAC operation requires DALEN = 1.
ADLEN	Left ADC Enable
	Right ADC Enable
	Enabling the right ADC must be done in the same I2C write operation that enables the left ADC. The right
ADREN	ADC can be enabled while the left ADC is running if used for DC measurements. SHDN must be toggled
	to disable the right ADC in this case. Right ADC operation requires ADLEN = 1.

Revision Code

The MAX9867 includes a revision code to allow easy identification of the device revision. The revision code is 0x42. See Table 18 for the revision code register.

Table 18. Revision Code Register

REGISTER	B7	B6	B5	B4	B3	B2	B1	В0	REGISTER ADDRESS
Revision	REV								0xFF

I²C Serial Interface

The MAX9867 features an I²C/SMBus-compatible, 2-wire serial interface consisting of a serial-data line (SDA) and a serial-clock line (SCL). SDA and SCL facilitate communication between the MAX9867 and the master at clock rates up to 400kHz. Figure 9 shows the 2-wire interface timing diagram. The master generates SCL and initiates data transfer on the bus. The master device writes data to the MAX9867 by transmitting the proper slave address followed by the register address and then the data word. Each transmit sequence is framed by a START (S) or REPEATED START (Sr) condition and a STOP (P) condition. Each word transmitted to the MAX9867 is 8 bits long and is followed by an acknowledge clock pulse. A master reading data from the MAX9867 transmits the proper slave address followed by a series of nine SCL pulses. The MAX9867 transmits data on SDA in sync with the master-generated SCL pulses. The master acknowledges receipt of each byte of data. Each read sequence is framed by a START or REPEATED START condition, a not acknowledge, and a STOP condition. SDA operates as both an input and an open-drain output. A pullup resistor, typically greater than 500 Ω is required on SDA. SCL operates only as an input. A pullup resistor, typically greater than 500Ω , is required on SCL if there are multiple masters on the bus, or if the single master has an open-drain SCL output. Series resistors in line with SDA and SCL are optional. Series resistors protect the digital inputs of the MAX9867 from high-voltage spikes on the bus lines, and minimize crosstalk, and undershoot of the bus signals.

Bit Transfer

One data bit is transferred during each SCL cycle. The data on SDA must remain stable during the high period of the SCL pulse. Changes in SDA while SCL is high are control signals. See the *START and STOP Conditions* section.

START and STOP Conditions

SDA and SCL idle high when the bus is not in use. A master initiates communication by issuing a START condition. A START condition is a high-to-low transition on SDA with SCL high. A STOP condition is a low-to-high transition on SDA while SCL is high (Figure 10). A START condition from the master signals the beginning of a transmission to the MAX9867. The master terminates transmission, and frees the bus, by issuing a STOP condition. The bus remains active if a REPEATED START condition is generated instead of a STOP condition.



Figure 9. 2-Wire Interface Timing Diagram



Figure 10. START, STOP, and REPEATED START Conditions

Early STOP Conditions

The MAX9867 recognizes a STOP condition at any point during data transmission except if the STOP condition occurs in the same high pulse as a START condition. For proper operation, do not send a STOP condition during the same SCL high pulse as the START condition.

Slave Address

The slave address is defined as the 7 most significant bits (MSBs) followed by the read/write bit. For the MAX9867, the 7 most significant bits are 0011000. Setting the read/write bit to 1 (slave address = 0x31) configures the MAX9867 for read mode. Setting the read/write bit to 0 (slave address = 0x30) configures the MAX9867 for write mode. The address is the first byte of information sent to the MAX9867 after the START condition.

Acknowledge The acknowledge bit (ACK) is a clocked 9th bit that the MAX9867 uses to handshake receipt each byte of data when in write mode (see Figure 11). The MAX9867 pulls down SDA during the entire master-generated 9th clock pulse if the previous byte is successfully received. Monitoring ACK allows for detection of unsuccessful data transfers. An unsuccessful data transfer occurs if a receiving device is busy or if a system fault has occurred. In the event of an unsuccessful data transfer, the bus master retries communication. The master pulls down SDA during the 9th clock cycle to acknowledge receipt of data when the MAX9867 is in read mode. An acknowledge is sent by the master after each read byte to allow data transfer to continue. A not acknowledge is sent when the master reads the final byte of data from the MAX9867, followed by a STOP condition.

Write Data Format

A write to the MAX9867 includes transmission of a START condition, the slave address with the R/W bit set to 0, 1 byte of data to configure the internal register address pointer, 1 or more bytes of data, and a STOP condition. Figure 12 illustrates the proper frame format for writing 1 byte of data to the MAX9867. Figure 10 illustrates the frame format for writing n bytes of data to the MAX9867.



Figure 11. Acknowledge



Figure 12. Writing 1 Byte of Data to the MAX9867

The slave address with the R/W bit set to 0 indicates that the master intends to write data to the MAX9867. The MAX9867 acknowledges receipt of the address byte during the master-generated 9th SCL pulse.

The second byte transmitted from the master configures the MAX9867's internal register address pointer. The pointer tells the MAX9867 where to write the next byte of data. An acknowledge pulse is sent by the MAX9867 upon receipt of the address pointer data.

The third byte sent to the MAX9867 contains the data that is written to the chosen register. An acknowledge pulse from the MAX9867 signals receipt of the data byte. The address pointer autoincrements to the next register address after each received data byte. This autoincrement feature allows a master to write to sequential registers within one continuous frame. Figure 13 illustrates how to write to multiple registers with one frame. The master signals the end of transmission by issuing a STOP (P) condition. Register addresses greater than 0x17 are reserved. Do not write to these addresses.

Read Data Format Send the slave address with the R/W bit set to 1 to initiate a read operation. The MAX9867 acknowledges receipt of its slave address by pulling SDA low during the 9th SCL clock pulse. A START (S) command followed by a read command resets the address pointer to register 0x00. The first byte transmitted from the MAX9867 is the content of register 0x00. Transmitted data is valid on the rising edge of SCL. The address pointer autoincrements after each read data byte. This autoincrement feature allows all registers to be read sequentially within one continuous frame. A STOP condition can be issued after any number of read data bytes. If a STOP condition is issued followed by another read operation, the first data byte to be read is from register 0x00.

The address pointer can be preset to a specific register before a read command is issued. The master presets the address pointer by first sending the MAX9867's slave address with the R/W bit set to 0 followed by the register address. A REPEATED START (Sr) condition is then sent followed by the slave address with the R/W bit set to 1. The MAX9867 then transmits the contents of the specified register. The address pointer autoincrements after transmitting the first byte.

The master acknowledges receipt of each read byte during the acknowledge clock pulse. The master must acknowledge all correctly received bytes except the last byte. The final byte must be followed by a not acknowledge from the master and then a STOP condition. Figure 14 illustrates the frame format for reading 1 byte from the MAX9867. Figure 15 illustrates the frame format for reading multiple bytes from the MAX9867.



Figure 13. Writing n Bytes of Data to the MAX9867



Figure 14. Reading 1 Byte of Data from the MAX9867





Figure 15. Reading n Bytes of Data from the MAX9867

Applications Information

Proper layout and grounding are essential for optimum performance. When designing a PCB for the MAX9867, partition the circuitry so that the analog sections of the MAX9867 are separated from the digital sections. This ensures that the analog audio traces are not routed near digital traces.

Use a large continuous ground plane on a dedicated layer of the PCB to minimize loop areas. Connect AGND and DGND directly to the ground plane using the shortest trace length possible. Proper grounding improves audio performance, minimizes crosstalk between channels, and prevents any digital noise from coupling into the analog audio signals.

Ground the bypass capacitors on MICBIAS, REG, PREG, and REF directly to the ground plane with minimum trace length. Also be sure to minimize the path length to AGND. Bypass AVDD directly to AGND. Connect all digital I/O termination to the ground plane with minimum path length to DGND. Bypass DVDD and DVDDIO directly to DGND.

Route microphone signals from the microphone to the MAX9867 as a differential pair, ensuring that the positive and negative signals follow the same path as closely as possible with equal trace length. When using single-ended microphones or other single-ended audio sources, ground the negative microphone input as near as possible to the audio source and then treat the positive and negative traces as differential pairs.

The MAX9867 TQFN package features an exposed thermal pad on its underside. Connect the exposed thermal pad to AGND.

An evaluation kit (EV Kit) is available to provide an example layout for the MAX9867. The EV kit allows quick setup of the MAX9867 and includes easy-to-use software, allowing all internal registers to be controlled.



Functional Diagram/Typical Operating Circuit

_Pin Configurations



Package Information

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
30 WLP	W302A2+3	<u>21-0211</u>
32 TQFN-EP	T3255+4	<u>21-0140</u>



Package Information (continued)

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages.



Package Information (continued)

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages.

COMMON DIMENSIONS PKG Isil 5x5 20L 5x5 <th></th>																							
PKG. 16L 5x5 20L						COM	10N I	IMEN	SIONS	:							EXPOSED PAD VARIATIONS						
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A2 0.20 REF. 0.20 REF. 0.20 0.20 0.23 0.23 0.23 0.23 0.23 0.23 0.20 0.25 0.20 0.25 0.20 0.25 0.20 0.25 0.20 0.25 0.20 0.25 0.20 0.25 0.20 0.21 0.23 0.20 <	A1	0	0.02 0.05	5 0	0.02	0.05	0	0.02	0.05	0	0.02	2 0.05	0	0.02	0.05		T1655-3	3.00	3.10	3.20	3.00	3.10	3.20
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N 16 20 28 32 40 ND 4 5 7 8 10 NE 4 5 7 8 10 JEDEC VHB VHC VHD-1 VHD-2 NOTES I. DIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M-1994. NOTES I. DIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M-1994.	L_L	0.30	0.40 0.5	0 0.45	0.55	0.65	0.45	0.55	0.65	0.30	0.40	0.50	0.30	0.40	0.50		T2055MN-5	3.15	3.25	3.35	3.15	3.25	3.35
ND457810JEDECVHHBVHHCVHHD-1VHHD-2NOTES:I. DIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M-1994.2. ALL DIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M-1994.2. ALL DIMENSIONS ARE IN MILLIMETERS. ANGLES ARE IN DEGREES.3. N IS THE TOTAL NUMBER OF TERMINALS. M THE TERMINAL #1 IDENTIFIER AND TERMINAL SUBGERING CONVENTION SHALLCONFORM TO JESD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER AREOPTIONAL, BUT MUST BE LOCATED VITHIN THE ZONE INDICATED. THE TERMINAL #1DIMENSION & APPLIES TO METALLIZED TERMINAL MID IS MEASURED BETWEEN $Q25$ mm AND 0.30 mm FRDM TERMINAL TIP. $Q25$ mm AND 0.30 mm FRDM TERMINAL SUG AS VELL AS THE TERMINALS. $Q25$ mm AND 0.30 mm FRDM TERMINAL SUG AS VELL AS THE TERMINALS. $Q25$ mm AND 0.30 mm FRDM TERMINAL TIP. $Q25$ mm AND 0.30 mm FRDM TERMINAL TIP. $Q25$ mm AND 0.30 mm FRDM TERMINAL AND IS MEASURED BETWEEN $Q25$ mm AND 0.30 mm FRDM TERMINAL SUG AS VELL AS THE TERMINALS. $Q26$ mm AND 0.30 mm FRDM TERMINAL TIP. $Q28$ mm AND 0.30 mm FRDM TERMINAL TIP. $Q28$ mm AND 0.30 mm FRDM TERMINAL TIP. $Q10$ MARKED FERENCE DINLY. 11 MARKING CONFIGNES TO LECCEPT EXPOSED PAD DIMENSION FOR $12855-3$, $12855-6$, T4055-1 AND T4055-2. $MMRKING S FOR PACKAGE CREENTATION REFERENCE DINLY.14 ALL DIMENSIONS APPLY TO BOTH LEADED (->) AND PbFREE (+) PKG. CODES.14 ALL DIMENSIONS APPLY TO BOTH LEA$	N	16			20		28		32			40			T2855-3	3.15	3.25	3.35	3.15	3.25	3.35		
NE 4 5 7 8 10 JEDEC VHHB VHHC VHHD-1 VHHD-2 NOTES I. DIMENSIDNING & TOLERANCING CONFORM TO ASME Y14.5M-1994. 280 280 2.70 2.80	ND	<u> </u>	4 5 7			8		<u> </u>	10			T2855-4	2.60	2.70	2.80	2.60	2.70	2.80					
LEEE WHRD Were 23.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.15 3.25 3.35 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20 3.00 3.10	NE	4		-	5		7		8		<u> </u>	10			T2855-5	2.60	2.70	2.80	2.60	2.70	2.80		
NDTES: 1. DIMENSIDNING & TOLERANCING CONFORM TO ASME Y14.5M-1994. 2. ALL DIMENSIDNIS ARE IN MILLIMETERS. ANGLES ARE IN DEGREES. 3. N IS THE TOTAL NUMBER OF TERMINALS. \bigcirc THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO LSD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE OPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE. \bigcirc DIMENSION & APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.25 mm AND 0.30 mm FROM TERMINAL TIP. \bigcirc ND AND NE REFER TO THE NUMBER OF TERMINAL SON EACH D AND E SIDE RESPECTIVELY. 7. DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION. \bigcirc DRAWING CONFORMS TO JEDEC MO220, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, T2855-6, TA055-1 AND T4055-2. \bigcirc WARPAGE SHALL NOT EXCEPT DIATE STORE THEAT SINK SLUG AS WELL AS THE TERMINALS. 9. DRAWING CONFORMS TO JEDEC MO220, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, T2855-6, TA055-1 AND T4055-2. \bigcirc WARPAGE SHALL NOT EXCEPT DIATE PREFERENCE DNLY. 12. NUMBER OF LEADS SHOWN ARE FOR REFERENCE DNLY. 13. MARKING IS FOR PACKAGE DIRIENTATION REFERENCE DNLY. 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PHOREE (+) PKG. CODES. -DRAWING NOT TO SCALE-	JEDEC									w w	ппр	-2				l l	T2855-6	3.15	3.25	3.35	3.15	3.25	3.35
NUTES: 1 DIMENSIDINING & TOLERANCING CONFORM TO ASME Y14.5M-1994. 2 3.5 3.25 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.15 3.22 3.35 3.10 3.20 3.10 3.20 3.10 3.20 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20 3.00 3.10 3.20																	T2855-7	2.60	2.70	2.80	2.60	2.70	2.80
1. DIMENSIDINING & TOLERANCING CONFORM TO ASME Y14.5M-1994. 2. ALL DIMENSIDINS ARE IN MULLIMETERS. ANGLES ARE IN DEGREES. 3. N IS THE TOTAL NUMBER OF TERMINALS. ▲ THE TERMINAL #1 IDENTIFIER AND TERMINALS. ▲ THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO JESD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE DPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD DR MARKED FEATURE. ▲ DIMENSION & APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.25 mm AND 0.30 mm FROM TERMINAL TIP. ▲ ND AND NE REFER TO THE NUMBER OF TERMINAL SON EACH D AND E SIDE RESPECTIVELY. 7. DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION. ▲ COPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS. 9. DRAWING CONFORMS TO JEDEC MIZEO, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, 72855-6, T4055-1 AND T4055-2. ▲ LEAD CENTERLINES TO BE AT TRUE POSITION AS DEFINED BY BASIC DIMENSION "e", ±005. 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PhOFREE (+) PKG. CODES. • DRAWING NOT TO SCALE-	NOTES	NDTES															T2855-8	3.15	3.25	3.35	3.15	3.25	3.35
 2. ALL DIMENSIONS ARE IN MILLIMETERS. ANGLES ARE IN DEGREES. 3. N IS THE TOTAL NUMBER OF TERMINALS. A THE TERMINAL #1 IDENTIFIER AND TERMINALS. A THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO USED 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE OPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE. A DIMENSION & APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.25 mm AND 0.30 mm FRUM TERMINAL TIP. M DAND NE REFER TO THE NUMBER OF TERMINALS ON EACH D AND E SIDE RESPECTIVELY. T DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION. COLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS VELL AS THE TERMINALS. DRAVING CONFORMS TO JEDEC MO20, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, T2855-6, T4055-1 AND T4055-2. VARPAGE SHALL NOT EXCEED 0.10 mm. MARKING IS FOR PACKAGE CRIENTATION REFERENCE DNLY. NUMBER OF LEADS SHOWN ARE FOR REFERENCE DNLY. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND POFREE (+) PKG. CODES. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND POFREE (+) PKG. CODES. DRAWING NOT TO SCALE- 	1. D	JIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M-1994.															T2855N-1	3.15	3.25	3.35	3.15	3.25	3.35
3. N IS THE TOTAL NUMBER OF TERMINALS. ▲ THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO JESD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE OPTIDNAL, BUT MUST BE LOCATED WITHIN THE ZDNE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE. T3255-4 3.00 3.10 3.20 3.00 3.10 3.20 M. DENTIFIER MAY BE CONTENTION SHALL CONFORM TO JESD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE DPTIDNAL, BUT MUST BE LOCATED WITHIN THE ZDNE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE. T3255-5 3.00 3.10 3.20 3.00 3.10	2. A	ALL DIMENSIONS ARE IN MILLIMETERS, ANGLES ARE IN DEGREES,														T3255-3	3.00	3.10	3.20	3.00	3.10	3.20	
A: THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO JESD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE DPTIONAL, BUT MUST BE LOCATED WITHIN THE ZINE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE LITHER A MOLD OR MARKED FEATURE. T3255-5 3.00 3.10 3.20 3.00 3.10 3.20 A: DIMENSION & APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.25 mm AND 0.30 mm FROM TERMINAL TIP. T4055-1 3.40 3.50 3.60 3.40 3.50 3.60 A: DIMENSION & APPLIES TO THE ALLIZED TERMINAL AND IS MEASURED BETWEEN 0.25 mm AND 0.30 mm FROM TERMINAL TIP. T4055-2 3.40 3.50 3.60 3.40 3.50 3.60 M: ND AND NE REFER TO THE NUMBER OF TERMINALS ON EACH D AND E SIDE RESPECTIVELY. T4055-2 3.40 3.50 3.60 3.40 3.50 3.60 M: COPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS. THE TERMINAL MID TEXCEED 0.10 mm. T4055-1 3.40 3.50 3.60 3.40 3.50 3.60 MARKING IS FOR PACKAGE DRIENTATION REFERENCE DINLY. NARKING IS FOR PACKAGE DRIENTATION REFERENCE DINLY. THE: PACKAGE DUTLINE, 16,20,28,3,2,40L THIN QFN, 5x5x0.75m 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-> AND PWFREE (+) PKG. CODES.	3. N	IS TH	ie total	NUM	BER D	F TE	RMIN/	ALS.									T3255-4	3.00	3.10	3.20	3.00	3.10	3.20
CONFORM TO JESD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE DPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE. DIMENSION & APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.25 mm AND 0.30 mm FRUM TERMINAL TIP. ND AND NE REFER TO THE NUMBER OF TERMINALS ON EACH D AND E SIDE RESPECTIVELY. 7. DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION. COPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS. 9. DRAVING CONFORMS TO JEDEC MO220, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, T2855-6, T4055-1 AND T4055-2. 11. MARKING IS FOR PACKAGE DRIENTATION REFERENCE ONLY. 12. NUMBER OF LEADS SHOLVN ARE FOR REFERENCE ONLY. 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PbFREE (+) PKG. CODES. -DRAWING NOT TO SCALE- DRAWING NOT TO SCALE-	<u>4</u> . T	THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CON∨ENTION SHALL															T3255M-4	3.00	3.10	3.20	3.00	3.10	3.20
$\begin{array}{c} 10^{11} \text{IIMAL}, \text{ BUT MUST BE LUCATED WITHIN THE ZURE INDICATED. THE TERMINAL #1}\\ \text{IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE.}\\ \hline \text{DIMENSION 6 APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN}\\ 0.25 mm AND 0.30 mm FROM TERMINAL TIP.\\ \hline \text{MOLD AND NE REFER TO THE NUMBER OF TERMINALS ON EACH D AND E SIDE RESPECTIVELY.}\\ \hline \text{DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION.}\\ \hline \text{MOLD COMPULATION IS POSSIBLE IN A SYMMETRICAL FASHION.}\\ \hline \text{MOLD ANDING CONFORMS TO JEDEC MOZO, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, T2855-6, T4055-1 AND T4055-2.\\ \hline \text{MARKING IS FOR PACKAGE DRIENTATION REFERENCE ONLY.}\\ 11. MARKING IS FOR PACKAGE DRIENTATION REFERENCE ONLY.\\ 12. NUMBER OF LEADS SHOW ARE FOR REFERENCE ONLY.\\ 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND POFREE (+) PKG. CODES.\\ \hline \text{MARKING NOT TO SCALE-}\\ \hline \text{DRAWING NOT TO SCALE-}\\ \hline \begin{array}{c} 111 \\ 112 \\ 112 \\ 123 \\ 124 \\ 124 \\ 124 \\ 124 \\ 124 \\ 124 \\ 124 \\ 124 \\ 124 \\ 124 \\ 124 \\ 125 \\ 124 \\ $	C	INFOR	MITOLJES	SD 95	-1 SP	P-012	2. DE	ETAIL	S OF	TER	MINA	AL #1	IDEN	TIFIE	RAF	Ε	T3255-5	3.00	3.10	3.20	3.00	3.10	3.20
$\begin{array}{c c c c c c c c c c c c c c c c c c c $	11		AL, BUI I	102.1	BF LL		נע עב				L IN		ED, I	HE I	ERMI	NAL #1	T3255N-1	3.00	3.10	3.20	3.00	3.10	3.20
TAUSSING DEPTICIES TO METRULAL TP. 0.25 m AND 0.30 mm FRUM TERMINAL TP. \therefore ND AND NE REFER TO THE NUMBER OF TERMINALS ON EACH D AND E SIDE RESPECTIVELY. $7.$ DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION. \bigcirc COPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS.9. DRAVING CONFORMS TO JEDEC MO220, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, T2855-6, T4055-1 AND T4055-2. \bigcirc VARPAGE SHALL NOT EXCEED 0.10 mm.11. MARKING IS FOR PACKAGE DRIENTATION REFERENCE DNLY.12. NUMBER OF LEADS SHOWN ARE FOR REFERENCE DNLY.14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND POFREE (+) PKG. CODESDRAWING NOT TO SCALE-	An	MENS1	TER MAT					שוא ויו חיד בו			אוטוא די די	(E.) D MEA	CUDE	ה פר		м	14055-1	3.40	3.50	3.60	3,40	3.50	3.60
Image: Second State Processing State Proces	<u>737</u> 1)	25 mm	ראס אום. ביח תאם	CIES	FRUM	TEPN	INAI		KMIINA	ar win	D 13	SMEM	SURE		IWEE	IN	T4055-2	3.40	3.50	3.60	3.40	3.50	3.60
 T. DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION. COPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS. DRAVING CONFORMS TO JEDEC M0220, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, T2855-6, T4055-1 AND T4055-2. WARPAGE SHALL NOT EXCEED 0.10 mm. MARKING IS FOR PACKAGE DITENTION REFERENCE ONLY. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PhFREE (+) PKG. CODES. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PhFREE (+) PKG. CODES. 	A N		NF RFF	TR TT	THE	NUMP		FTF	RMIN		IN F	ACH		ודי	THE		T4055N-1	3.40	3.50	3.60	3.40	3.50	3.60
CUPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS. DRAWING CONFORMS TO JEDEC M0220, EXCEPT EXPOSED PAD DIMENSION FOR T2855-3, T2855-6, T4055-1 AND T4055-2. WARPAGE SHALL NOT EXCEED 0.10 mm. MARKING IS FOR PACKAGE DITENTION REFERENCE ONLY. LE NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY. LEAD CENTERLINES TO BE AT TRUE POSITION AS DEFINED BY BASIC DIMENSION 'e', ±0.05. 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PbFREE (+) PKG. CODES. DRAWING NOT TO SCALE- DRAWING NOT TO SCALE-	7. D		LATION I	S POS	SIBLE	IN	A SY	MMETI	RICAL	FAS	HID	N.					T4055MN-1	3.40	3.50	3.60	3.40	3.50	3.60
P. DRAWING CUNFURMS TU JEDEC MU220, EXCEPT EXPOSED PAD DIMENSION FUR T2855-3, T2855-6, T4055-1 AND T4055-2. WARPAGE SHALL NOT EXCEED 0.10 mm. II. MARKING IS FUR PACKAGE DIRIENTATION REFERENCE UNLY. LAUN WHER OF LEADS SHOWN ARE FUR REFERENCE UNLY. LEAD CENTERLINES TU BE AT TRUE POSITION AS DEFINED BY BASIC DIMENSION 'e', ±0.05. I4. ALL DIMENSIONS APPLY TU BOTH LEADED (-) AND PbFREE (+) PKG. CUDES. DRAWING NOT TO SCALE- DRAWING NOT TO SCALE-	/8. c	IPLAN	ARITY AP	PLIES	TO	THE E	XPD	SED H	IEAT	SINK	SLI	 Jg as	WEL	LAS	THE	TERMINALS.							
T2855-3, T2855-6, T4055-1 AND T4055-2. Image: Construction of the state of the stat	9. D	RAWIN	G CONFOR	RMS T	o jei	DEC M	10220	, EXC	EPT	EXPO	SED	PAD	DIME	NSID	N FOF	2							
ID. WARPAGE SHALL NOT EXCEED 0.10 mm. 11. MARKING IS FOR PACKAGE DRIENTATION REFERENCE DNLY. 12. NUMBER OF LEADS SHOWN ARE FOR REFERENCE DNLY. 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND POFREE (+) PKG. CODES. -DRAWING NOT TO SCALE- 21 - 014.0	, Ti	2855-	3, T2855	-6, T	4055-	1 AN	D T40)55-2	2.														
11. MARKING IS FOR PACKAGE DRIENTATION REFERENCE DNLY. 12. NUMBER OF LEADS SHOWN ARE FOR REFERENCE DNLY. 13. LEAD CENTERLINES TO BE AT TRUE POSITION AS DEFINED BY BASIC DIMENSION 'e', ±0.05. 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND POFREE (+) PKG. CODES. -DRAWING NOT TO SCALE- DRAWING NOT TO SCALE-	<u>⊿0⊾</u> w	ARPAG	ie shall	NDT	EXCE	ED 0.:	l0 mm	•															
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223 LEAD CENTERLINES TO BE AT TRUE POSITION AS DEFINED BY BASIC DIMENSION "P", ±0.05. 14. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND POFREE (+) PKG. CODES. -DRAWING NOT TO SCALE- DRAWING NOT TO SCALE- 16.20,28,32,40L THIN QFN, 5x5x0.75m APPROVAL DOCUMENT CONTROL NO. REV. 2 21-0140 M 22	12. N	NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.																					
-DRAWING NOT TO SCALE-	14 A	וט עראב אזרד ו			J BE			1021 I		42 DI 42 DI		15 L B			TWEN	210N -6-, ±0.02	TITLE:						
-DRAWING NOT TO SCALE-	TH HE SIMENSIONS HITEL TO SUM LEASES () HAS IN REL () TRO, CODES,											PACK	AGE	JUTLI	NE,								
-DRAWING NOT TO SCALE-																	16,20	,28,30	40L	I HIN		, JX:	
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54

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