461

(315) 701-6751

# **HIGH SPEED/VOLTAGE OP AMP**

## M.S.KENNEDY CORP.

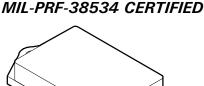
4707 Dey Road Liverpool, N.Y. 13088

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### FEATURES:

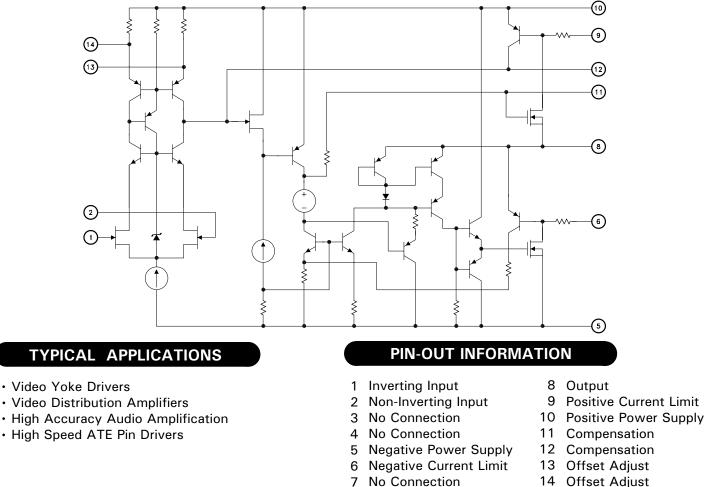
- Extremely Fast 500v/µS
- Wide Supply Range  $\pm 15V$  to  $\pm 45V$
- VMOS Output, No S.O.A. Restrictions .
- Large Gain-Bandwidth Product
- FET Input
- **Electrically Isolated Case**
- 800mA Typical Output Current

### DESCRIPTION:



#### The MSK 1461 is a state of the art high speed FET input operational amplifier. The distinguishing characteristic of the MSK 1461 is its unique VMOS output stage which completely eliminates the safe operating area restrictions associated with secondary breakdown of bipolar transistor output stage op-amps. Freedom from secondary breakdown allows the 1461 to handle large output currents at any voltage level limited only by transistor junction temperature. 115 dB of open loop gain gives the 1461 high closed loop gain accuracy and the typical $\pm$ 1.0mV of input offset voltage will fit well in any error budget. A 500 V/ $\mu$ S slew rate and 1200 MHz gain bandwidth product make the 1461 an outstanding high-speed op-amp. A single external capacitor is used for compensation and output current limiting is user programmable through the selection of two external resistors.

#### EQUIVALENT SCHEMATIC



- - Compensation
- 12 Compensation
- Offset Adjust
- 14 Offset Adjust

### **ABSOLUTE MAXIMUM RATINGS**

±Vсс	Supply Voltage ±45V
Ιουτ	Output Current
VIN	Differential Input Voltage
	Thermal Resistance
	Junction to Case
	(Output Devices Only)

- T<sub>ST</sub> Storage Temperature Range -65°C to +150°C

	(10 Seconds)
Tс	Case Operating Temperature
	(MSK 1461B)
	(MSK 1461)

Junction Temperature +175°C ТJ

### **ELECTRICAL SPECIFICATIONS**

<b>n</b> /	Tool Oran Hitlerin	Group A	MSK 1461B			MSK 1461			
Parameter	Test Conditions	Subgroup	Min.	Тур.	Max.	Min.	Тур.	Max.	Units
STATIC									
Supply Voltage Range ③		-	±15	-	±45	±15	-	±45	V
Quiescent Current	VIN = 0V	1	-	±19	± 25	-	±19	±28	mA
	VIN=OV	2,3	-	±21	± 35	-	-	-	mA
Thermal Resistance ③	Junction to Case	-	-	11	12	-	11	15	°C/W
INPUT									
Input Offset Voltage	$V_{IN} = 0V  A_V = -10V/V$	1	-	±1.0	±5.0	-	±1.0	±8.0	mV
Input Offset Voltage Drift	Bal. Pins = N/C	2,3	-	±6.0	±50	-	±10	-	µV/°C
Input Offset Adjust ③	RPOT = $10K\Omega$ to + Vcc	-	-	±8.0	-	-	±8.0	-	V
have the Connect	Vcm=0V	1	-	±10	±300	-	±10	± 300	pА
Input Bias Current	Either Input	2,3	-	±10	±100	-	-	-	nA
have the former the		-	-	±5.0	-	-	±5.0	-	pА
Input Offset Current ③	Vcm=0V	-	-	±5.0	-	-	-	-	nA
Input Impedance ③	F = DC	-	-	3x1012	- 2	-	3x1012	-	Ω
Common Mode Range ③		-	±22	±24	-	±22	±24	-	V
Common Mode Rejection Ratio	3 F=10KHz Vсм=±22V	4	90	100	-	90	100	-	dB
OUTPUT									
	$R_L = 50\Omega$ $A_V = -5V/V$	4	±27	±31	-	±27	±31	-	V
Output Voltage Swing	$R_L = 1 K\Omega$	4	±30	±33	-	± 30	±33	-	V
Output Current, Peak $R_L = 33\Omega A_V = -5V/V T_J < 175 ^{\circ}C$		4	±600	±800	-	±600	±800	-	mA
Settling Time ②③	0.1% 10V step	4	-	400	800	-	400	800	nS
TRANSFER CHARACTERISTICS									
Slew Rate	$V_{OUT} = \pm 10V R_L = 1K\Omega Av = -5V/V$	4	200	500	-	200	500	-	V/µS
Open Loop Voltage Gain ③	$R_L = 1K\Omega$ $F = 100Hz$	4	90	106	-	90	106	-	dB
Gain Bandwidth Product ③ F = 100KHz		4	800	1200	-	800	1200	-	MHz

### NOTES:

 $Rsc\!=\!0\Omega$  and  $\pm\,Vcc\!=\!36VDC$  unless otherwise specified.

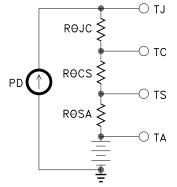
AV = -1, measured in false summing junction circuit.

- Devices shall be capable of meeting the parameter, but need not be tested. Typical parameters are for reference only.
- 1034667 Industrial grade devices shall be tested to subgroups 1 and 4 unless otherwise specified.
- Military grade devices ("B" suffix) shall be 100% tested to subgroups 1,2,3 and 4.
- Subgroup 5 and 6 testing available upon request. Subgroup 1.4  $Tc = +25^{\circ}C$ Subgroup 2.5  $TJ = +125^{\circ}C$ Subgroup 3.6  $TA = -55^{\circ}C$

### HEAT SINKING

To select the correct heat sink for your application, refer to the thermal model and governing equation below.

#### Thermal Model:



### Governing Equation:

 $T_J = P_D x (R_{\theta JC} + R_{\theta CS} + R_{\theta SA}) + T_A$ 

#### Where

T<sub>J</sub> = Junction Temperature

- PD = Total Power Dissipation
- $R_{\theta JC}$  = Junction to Case Thermal Resistance
- $R\theta CS = Case to Heat Sink Thermal Resistance$
- R<sub>0SA</sub> = Heat Sink to Ambient Thermal Resistance
- Tc = Case Temperature
- TA = Ambient Temperature

Ts = Sink Temperature

### Example:

In our example the amplifier application requires the output to drive a 20 volt peak sine wave across a  $400\Omega$  load for 50mA of peak output current. For a worst case analysis we will treat the 50mA peak output current as a D.C. output current. The power supplies shall be set to  $\pm 40$ VDC.

#### 1.) Find Driver Power Dissipation

- PD = [(quiescent current) x (+Vs (-Vs))] + [(+Vs-Vo) x IoUT]
  - $= [(50mA) \times (80V)] + [(20V) \times (0.05A)]$
  - = 4W + 1.0W
  - = 5Watts
- 2.) For conservative design, set  $T_J = +125 \,^{\circ}C$ .
- 3.) For this example, worst case  $TA = +50^{\circ}C$
- 4.)  $R_{\theta JC} = 12^{\circ}C/W$  from MSK 1461B Data Sheet
- 5.) R\_{0CS} = 0.15  $^{o}C/W$  for most thermal greases
- 6.) Rearrange governing equation to solve for R $\theta$ SA R $\theta$ SA = ((TJ - TA)/PD) - (R $\theta$ JC) - (R $\theta$ CS)
  - $= ((125^{\circ}C 50^{\circ}C) / 5W) (12^{\circ}C/W) (.15^{\circ}C/W)$  $= 2.85^{\circ}C/W$

The heat sink in this example must have a thermal resistance of no more than  $2.85^{\circ}$ C/W to maintain a junction temperature of no more than  $+ 125^{\circ}$ C.

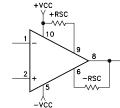
#### CURRENT LIMIT

The output current of the MSK 1461 is internally limited to approximately  $\pm750mA$  by two  $0.8\Omega$  internal current limit resistors. Additional current limit can be achieved through the use of two external current limit resistors. One resistor (+Rsc) limits the positive output current and the other (-Rsc) limits the negative output current. The value of the current limit resistors can be determined as follows:

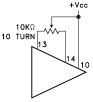
$$\pm\,\text{Rsc}~=~[(0.65\,\text{V}/\pm\,\text{ILIM})$$
 -  $0.8\Omega]$ 

Since the 0.65V term is obtained from the base to emitter voltage drop of a bipolar transistor, the equation only holds true for +25 °C operation. As case temperature increases, the 0.65V term will decrease making the actual current limit set point decrease slightly.

The following schematic illustrates how to connect each current limit resistor:



#### INPUT OFFSET ADJUST CONNECTION



### POWER SUPPLY BYPASSING

Both the negative and the positive power supplies must be effectively decoupled with a high and low frequency bypass circuit to avoid power supply induced oscillation. An effective decoupling scheme consists of a  $0.1\mu$ F ceramic capacitor in parallel with a  $4.7\mu$ F tantalum capacitor from each power supply pin to ground.

### SAFE OPERATING AREA

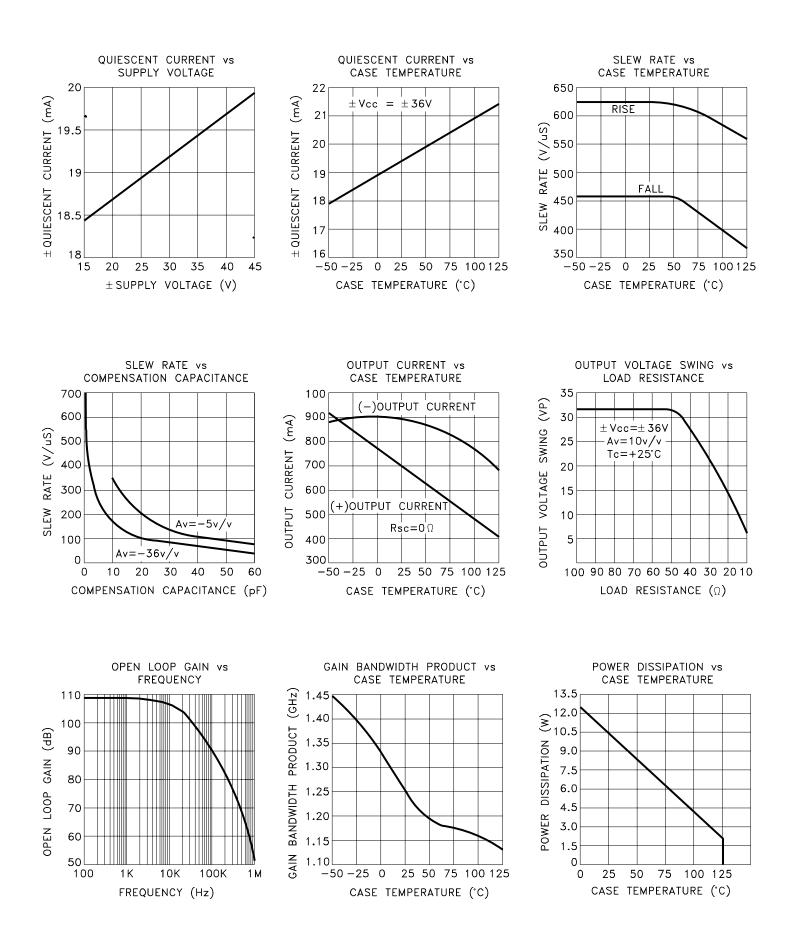
Any designer who has worked with power operational amplifiers is familiar with Safe Operating Area (S.O.A.) curves. S.O.A. curves are a graphical representation of the following three power limiting factors of any bipolar transistor output op-amp.

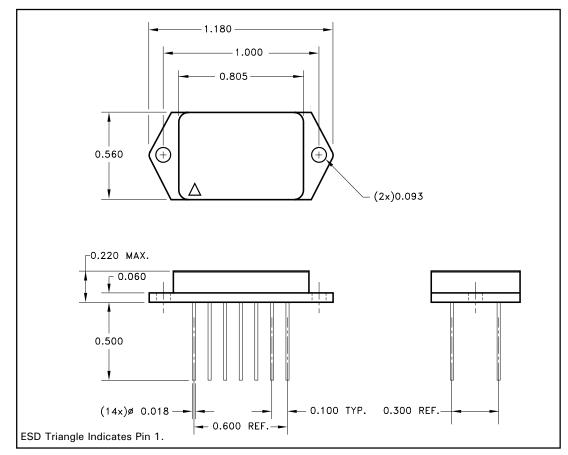
- 1. Wire Bond Current Carrying Capability
- 2. Transistor Junction Temperature
- 3. Secondary Breakdown Limitations

Since the MSK 1461 utilizes a MOSFET output, there are no secondary breakdown limitations and therefore no need for S.O.A. curves. The only limitation on output power is the junction temperature of the output drive transistors.

Whenever possible, junction temperature should be kept below 150°C to ensure high reliability. See "Heat Sinking" for more information involving junction temperature calculations.

### TYPICAL PERFORMANCE CURVES





NOTE: ALL DIMENSIONS ARE  $\pm 0.010$  UNLESS OTHERWISE LABELED.

# ORDERING INFORMATION

Part Number	Screening Level
MSK1461	Industrial
MSK1461B	Military-Mil-PRF-38534

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