

### 3V PROGRAMMABLE GAIN POWER AMPLIFIER

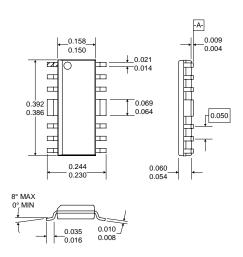
### Typical Applications

- Analog Communication Systems
- 900MHz Spread Spectrum Systems
- 400MHz Industrial Radios

- Driver Stage for Higher Power Applications
- 3V Applications

### **Product Description**

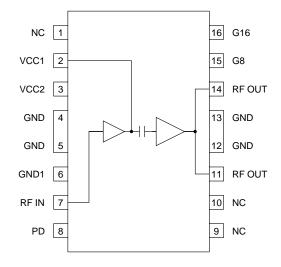
The RF2155 is a 3V medium power programmable gain amplifier IC. The device is manufactured on an advanced Gallium Arsenide Heterojunction Bipolar Transistor (HBT) process, and has been designed for use as the final RF amplifier in analog cellular phone transmitters or ISM applications operating at 915MHz. The device is self-contained with the exception of the output matching network and power supply feed line. A two-bit digital control provides 4 levels of power control, in 8dB steps.



Package Style: Standard Batwing

### Optimum Technology Matching® Applied

- ☐ Si BJT
  ☐ Si Bi-CMOS
- ✓ GaAs HBT☐ SiGe HBT
- ☐ GaAs MESFET☐ Si CMOS



Functional Block Diagram

#### **Features**

- Single 3V Supply
- 500mW CW Output Power
- 31 dB Small Signal Gain
- Up to 60% Efficiency
- Digitally Controlled Output Power
- 430MHz to 930MHz Frequency Range

#### Ordering Information

RF2155 3V Programmable Gain Power Amplifier RF2155 PCBA Fully Assembled Evaluation Board

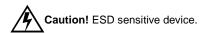
RF Micro Devices, Inc. 7628 Thorndike Road Greensboro, NC 27409, USA Tel (336) 664 1233 Fax (336) 664 0454 http://www.rfmd.com

Rev B3 010417 2-179

## RF2155

### **Absolute Maximum Ratings**

Parameter	Rating	Unit
Supply Voltage	-0.5 to +5.5	$V_{DC}$
Power Down Voltage (V <sub>PD</sub> )	-0.5 to +3.3	V
DC Supply Current	500	mA
Input RF Power	+10	dBm
Output Load VSWR	10:1	
Ambient Operating Temperature	-30 to +85	°C
Storage Temperature	-40 to +150	$^{\circ}$



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Davamatar	Specification		l lm!4	Condition		
Parameter	Min.	Тур.	Max.	Unit	Condition	
Overell					T=25 °C, V <sub>CC</sub> =3.6 V, V <sub>PD</sub> =3.0 V,	
Overall					$Z_{LOAD}$ =13 $\Omega$ , $P_{IN}$ =0dBm, Freq=915MHz	
Frequency Range		430 to 930		MHz		
Maximum CW Output Power		450		mW	V <sub>CC</sub> =3.6V	
		300		mW	$V_{CC}=3.0V$	
Small Signal Gain		31		dB		
Second Harmonic		-30		dBc	Without external second harmonic trap	
Third Harmonic		-40		dBc		
Fourth Harmonic		-36		dBc		
Input VSWR		2:1			All gain settings	
CW Efficiency	50	56		%	G16="high", G8="high", P <sub>IN</sub> =0dBm	
Output Load VSWR	6:1				Spurious<-60dBc	
Power Control						
Power Down "ON"	2.7	2.8	3.0	V	Voltage supplied to the input	
Power Down "OFF"	0	0.5	0.8	V	Voltage supplied to the input	
PD Input Current		3.7	5.0	mA	Only in "ON" state	
G16, G8 "ON"	2.2	2.5	3.0	V	Voltage supplied to the input	
G16, G8 "OFF"	0	0.3	0.5	V	Voltage supplied to the input	
G16, G8 Input Current	0.8	1.0	1.6	mA	Only in "ON" state	
Output Power	+25.5	+26.5	+28.0	dBm	G16="high", G8="high", P <sub>IN</sub> =0dBm	
	+16.0	+18.5	+21.0	dBm	G16="high", G8="low", P <sub>IN</sub> =0dBm	
	+8.0	+10.5	+13.0	dBm	G16="low", G8="high", P <sub>IN</sub> =0dBm	
	-1.0	+1.5	+4.0	dBm	G16="low", G8="low", P <sub>IN</sub> =0dBm	
Turn On/Off Time			100	ns		
Power Supply						
Power Supply Voltage		3.6		V	Specifications	
	3.0		5.0	V	Operating limits	
Power Supply Current		225	300	mA	G16="high", G8="high", P <sub>IN</sub> =0dBm	
		90	115	mA	G16="high", G8="low", P <sub>IN</sub> =0dBm	
		37	55	mA	G16="low", G8="high", P <sub>IN</sub> =0dBm	
		25	35	mA	G16="low", G8="low", P <sub>IN</sub> =0dBm	
	20	50	110	mA	G16="high", G8="high", No RF In	
		1	10	μΑ	G16="low", G8="low", PD="low"	

2-180 Rev B3 010417

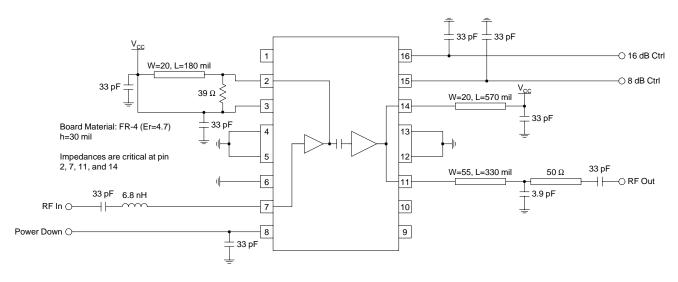
3 4 5 6 7	VCC2 GND GND1	Not internally connected. Positive supply for the first stage (driver) amplifier. This is an unmatched transistor collector output. This pin should see an inductive path to AC ground ( $V_{CC}$ with a UHF bypassing capacitor). This inductance can be achieved with a short, thin microstrip line (approximately equivalent to 0.4 nH). At lower frequencies, the inductance value should be larger (longer microstrip line) and $V_{CC}$ should be bypassed with a larger bypass capacitor. This inductance forms a matching network with the amplifier stages, setting the amplifier's frequency of maximum gain. An additional $1\mu F$ bypass capacitor in parallel with the UHF bypass capacitor is also recommended, but placement of this component is not as critical. A resistor of $39\Omega$ from this pin to pin 3 is necessary to ensure stability under extreme output VSWR conditions. Positive supply for the bias circuits. This pin should be bypassed with a single UHF capacitor, placed as close as possible to the package. Ground connection. Keep traces physically short and connect immediately to the ground plane for best performance.	Interface Schematic  VCC1 From Bias Stages
3 4 5 6 7	VCC2 GND GND GND1	Positive supply for the first stage (driver) amplifier. This is an unmatched transistor collector output. This pin should see an inductive path to AC ground ( $V_{CC}$ with a UHF bypassing capacitor). This inductance can be achieved with a short, thin microstrip line (approximately equivalent to 0.4nH). At lower frequencies, the inductance value should be larger (longer microstrip line) and $V_{CC}$ should be bypassed with a larger bypass capacitor. This inductance forms a matching network with the amplifier stages, setting the amplifier's frequency of maximum gain. An additional $1\mu F$ bypass capacitor in parallel with the UHF bypass capacitor is also recommended, but placement of this component is not as critical. A resistor of $39\Omega$ from this pin to pin 3 is necessary to ensure stability under extreme output VSWR conditions. Positive supply for the bias circuits. This pin should be bypassed with a single UHF capacitor, placed as close as possible to the package. Ground connection. Keep traces physically short and connect immediately to the ground plane for best performance.	RF IN From Bias
4 5 6 7	GND GND GND1	tance can be achieved with a short, thin microstrip line (approximately equivalent to 0.4nH). At lower frequencies, the inductance value should be larger (longer microstrip line) and $V_{CC}$ should be bypassed with a larger bypass capacitor. This inductance forms a matching network with the amplifier stages, setting the amplifier's frequency of maximum gain. An additional $1\mu F$ bypass capacitor in parallel with the UHF bypass capacitor is also recommended, but placement of this component is not as critical. A resistor of $39\Omega$ from this pin to pin 3 is necessary to ensure stability under extreme output VSWR conditions. Positive supply for the bias circuits. This pin should be bypassed with a single UHF capacitor, placed as close as possible to the package. Ground connection. Keep traces physically short and connect immediately to the ground plane for best performance.	From Bias =
4 5 6 7	GND GND GND1	larger bypass capacitor. This inductance forms a matching network with the amplifier stages, setting the amplifier's frequency of maximum gain. An additional $1\mu F$ bypass capacitor in parallel with the UHF bypass capacitor is also recommended, but placement of this component is not as critical. A resistor of $39\Omega$ from this pin to pin 3 is necessary to ensure stability under extreme output VSWR conditions. Positive supply for the bias circuits. This pin should be bypassed with a single UHF capacitor, placed as close as possible to the package. Ground connection. Keep traces physically short and connect immediately to the ground plane for best performance. Same as pin 4.	
4 5 6 7	GND GND GND1	single UHF capacitor, placed as close as possible to the package.  Ground connection. Keep traces physically short and connect immediately to the ground plane for best performance.  Same as pin 4.	
5 6 7	GND GND1	ately to the ground plane for best performance.  Same as pin 4.	
7	GND1	· · · · · · · · · · · · · · · · · · ·	
7			
		Ground return for the first stage; this should be connected to a via very close to the device.	
8	RF IN	Amplifier RF input. This is a $50\Omega$ RF input port to the amplifier. To improve the input match over all four gain control settings, an input inductor of 6.8 nH should be added. The amplifier does not contain internal DC blocking and, therefore, should be externally DC blocked before connecting to any device which has DC present or which contains a DC path to ground. A series UHF capacitor is recommended for the DC blocking.	See pin 2.
	PD	Power down control voltage. When this pin is at 0V, the device will be in power down mode, dissipating minimum DC power. When this pin is at 3V the device will be in full power mode delivering maximum available gain and output power capability. This pin should not, in any circumstance, be higher than 3.3V. This pin should also have an external UHF and HF bypassing capacitor.	To RF Stages
9	NC	Not internally connected.	
10	NC	Not internally connected.	
11	RF OUT	Amplifier RF output. This is an unmatched collector output of the final amplifier transistor. It is internally connected to pins 11 and 14 to provide low series inductance and flexibility in output matching. Bias for the final power amplifier output transistor must also be provided through one of these pins. Typically, pin 14 is used to supply bias. A transmission line of approximately 500 mils length, followed by a bypass capacitor, is adequate. This pin can also be used to create a second harmonic trap. A UHF and large tantalum (1 $\mu$ F) capacitor should be placed on the power supply side of the bias inductor. Pin 11 should be used for the RF output with a matching network that presents the optimum load impedance to the PA for maximum power and efficiency, as well as providing DC blocking at the output.	RF OUT  From Blas Stages
12	GND	Same as pin 4.	
13	GND	Same as pin 4.	
14	RF OUT	Same as pin 11.	
15	G8	RF output power gain control 8dB bit (see specification table for logic). The control voltage at this pin should never exceed 3.3V and a logic high should be at least 2.7V. This pin should also have an external UHF bypassing capacitor.	VCC2  To RF Stages

Rev B3 010417 2-181

## RF2155

Pin	Function	Description	Interface Schematic
16	G16	RF output power gain control 16dB bit (see specification table for logic). The control voltage at this pin should never exceed 3.3V and a logic high should be at least 2.7V. This pin should also have an external UHF bypassing capacitor.	Same as pin 15.

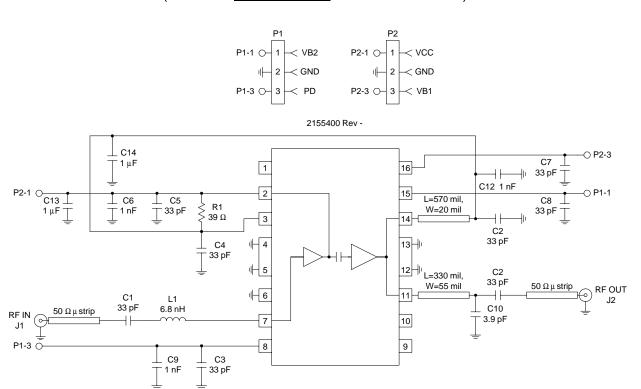
## Application Schematic 915 MHz



2-182 Rev B3 010417

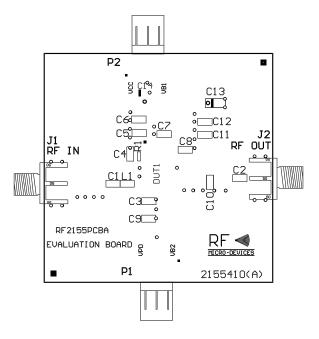
### **Evaluation Board Schematic**

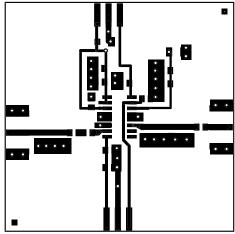
(Download Bill of Materials from www.rfmd.com.)



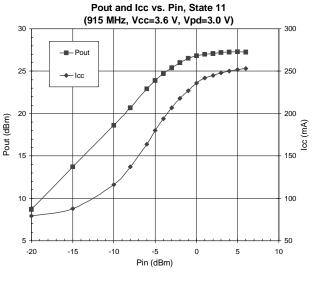
Rev B3 010417 2-183

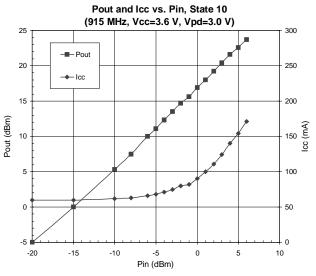
# Evaluation Board Layout Board Size 2.0" x 2.0"

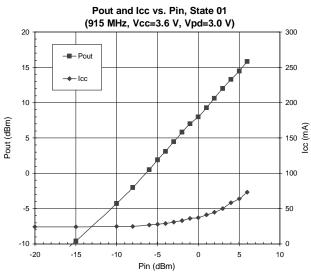


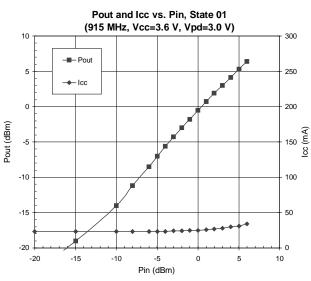


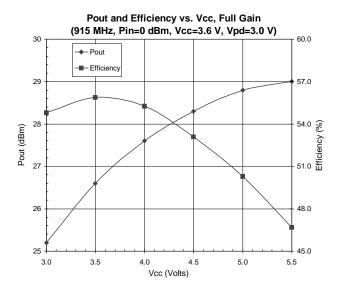
Rev B3 010417 2-184

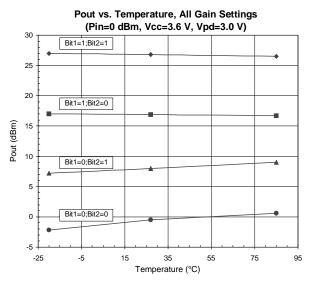




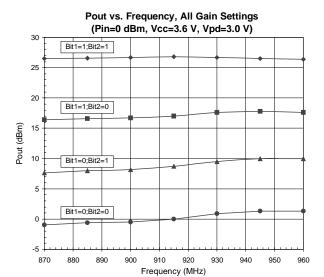


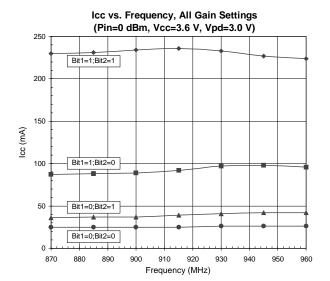






Rev B3 010417 2-185





2-186 Rev B3 010417