# +3.0 V to +5.5V USB Power Control Switch 

Compliant to USB Specifications
■ +3.0 V to +5.5 V Input Voltage Range

- Two Independent Power Switches
- Two Error Flag Outputs, Open Drain
- 2.7V Undervoltage Lockout

■ 500mA Minimum Continuous Load Current Per Switch

- 1.25A Short Circuit Current Limit
- $140 \mathrm{~m} \Omega$ Maximum On-Resistance
- $80 \mu \mathrm{~A}$ On-State Supply Current
- $1 \mu \mathrm{~A}$ Shutdown Current
- Output Can Be Forced Higher Than Input (Off-State)
- Thermal Shutdown
- 1 ms Soft-Start Power Up
- Active-High Version: SP2526-1

■ Active-Low Version: SP2526-2


## DESCRIPTION

The SP2526 device is a dual +3.0 V to +5.5 V USB Supervisory Power Control Switch ideal for self-powered and bus-powered Universal Serial Bus (USB) applications. Each switch has low on-resistance ( $80 \mathrm{~m} \Omega$ typical) and can supply 500 mA minimum. The fault currents are limited to 2.0 A typical and the flag output pin for each switch is available to indicate fault conditions to the USB controller. The 1 ms soft start will eliminate any momentary voltage droop on the upstream port that may occur when the switch is enabled in bus-powered applications. The thermal shutdown feature will prevent damage to the device when subjected to excessive current loads. The undervoltage lockout feature will ensure that the device will remain off unless there is a valid input voltage present.


## ABSOLUTE MAXIMUM RATINGS

These are stress ratings only and functional operation of the device at these ratings or any other above those device. Permanent damage may occur on indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability. to handling this device. The protective foam should be discharged to the destination socket before devices are removed.
Supply Voltage. -0.3V, +6.0V
Operating Temperature.................... $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
Storage Temperature $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$

## Power Dissipation Per Package

8 -pin NSOIC
(derate $6.14 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ )................ 500 mW
8 -pin PDIP
(derate $11.8 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ). .1000 mW

## SPECIFICATIONS

Unless otherwise noted, the following specifications apply for $\mathrm{V}_{\text {IN }}=+5.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.

| PARAMETER | MIN. | TYP. | MAX. | UNITS | CONDITIONS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Operating Voltage Range, $\mathrm{V}_{\text {IN }}$ | 3.0 |  | 5.5 | V |  |
| Supply Current, $\mathrm{I}_{\mathbb{N}}$ |  | $\begin{gathered} 0.05 \\ 80 \end{gathered}$ | $\begin{aligned} & 5.0 \\ & 140 \end{aligned}$ | $\mu \mathrm{A}$ | enable off, no output load enable on, no output load |
| Enable Input Threshold Voltage | 0.8 | $\begin{aligned} & 2.1 \\ & 1.9 \end{aligned}$ | 2.4 | V | low to high transition high to low transition |
| Enable Input Current |  | $\begin{aligned} & 0.01 \\ & 0.01 \end{aligned}$ | $\begin{aligned} & 1 \\ & 1 \end{aligned}$ | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{EE}}=\mathrm{V}_{\text {OH(min) }}=2.4 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{EN}}=\mathrm{V}_{\mathrm{OL}(\text { max })}=0.8 \mathrm{~V} \end{aligned}$ |
| Enable Input Capacitance |  | 1 |  | pF |  |
| Switch Resistance |  | 80 | 140 | $\mathrm{m} \Omega$ | $3.3 \mathrm{~V} \leq \mathrm{V}_{1 \mathbb{N}} \leq 5.0 \mathrm{~V}, \mathrm{l}_{\text {L }} \leq 500 \mathrm{~mA}$ |
| Output Turn-On Delay |  | 0.5 |  | ms | $\mathrm{R}_{\mathrm{L}}=10 \Omega$ each output |
| Output Turn-On Rise Time |  | 1 |  | ms | $\mathrm{R}_{\mathrm{L}}=10 \Omega$ each output |
| Output Turn-Off Delay |  | 1 | 20 | $\mu \mathrm{s}$ | $\mathrm{R}_{\mathrm{L}}=10 \Omega$ each output |
| Output Turn-Off Fall Time |  | 1 | 20 | $\mu \mathrm{s}$ | $\mathrm{R}_{\mathrm{L}}=10 \Omega$ each output |
| Output Leakage Current |  | . 02 | 10 | $\mu \mathrm{A}$ | each output |
| Current Limit Threshold |  | 2.0 | 2.8 | A | each output, Ramped load applied to enable output |
| Short-Circuit Current Limit | 0.5 | 0.75 | 1.25 | A | each output, $\mathrm{V}_{\text {out }}=0 \mathrm{~V}$ |
| Over-Temperature Shutdown Threshold |  | $\begin{aligned} & 135 \\ & 125 \end{aligned}$ |  | ${ }^{\circ} \mathrm{C}$ | increasing temperature decreasing temperature |
| Error Flag Output Resistance |  | $\begin{aligned} & 10 \\ & 15 \end{aligned}$ | $\begin{aligned} & 25 \\ & 40 \end{aligned}$ | $\Omega$ | $\begin{aligned} & V_{I N}=5 \mathrm{~V}, \mathrm{I}_{\mathrm{L}}=10 \mathrm{~mA} \\ & V_{I N}=3.3 \mathrm{~V}, \mathrm{I}_{\mathrm{L}}=10 \mathrm{~mA} \end{aligned}$ |
| Error Flag Off Current |  | 0.01 | 1 | $\mu \mathrm{A}$ | $\mathrm{V}_{\text {FLAG }}=5 \mathrm{~V}$ |
| UVLO Threshold | 2.5 | $\begin{aligned} & 2.7 \\ & 2.6 \end{aligned}$ | 3.0 | V | $\mathrm{V}_{\text {IN }}$ increasing, $\mathrm{T}_{\text {MIN }}$ to $\mathrm{V}_{\text {IN }}$ decreasing, $\mathrm{T}_{\text {MAX }}$ |



## PIN ASSIGNMENTS

Pin 1 - ENA - Enable Input for MOSFET Switch A. This input pin is active high for the SP2526-1 and active low for the SP2526-2.

Pin 2 - FLGA - Fault Flag Output for Switch Channel A. This open drain output pin pulls low to indicate overcurrent, undervoltage lockout, and thermal shutdown conditions.

Pin 3 - FLGB — Fault Flag Output for Switch Channel B. This open drain output pin pulls low to indicate overcurrent, undervoltage lockout, and thermal shutdown conditions.

Pin 4 - ENB - Enable Input for MOSFET Switch B. This input pin is active high for the SP2526-1 and active low for the SP2526-2.

Pin 5 - OUTA - Output for MOSFET Switch Channel A. This is the output pin of the MOSFET source of channel A, typically connected to the switched side of the load.

Pin 6 - GND - Ground reference. Supply return.

Pin 7 - IN — Supply Input. This input pin is connected to the MOSFET drain of both switches. Connect +3.0 V to +5.5 V to this input pin. The voltage at this input pin also supplies the internal circuitry.

Pin 8 - OUTB — Output for MOSFET Switch Channel B. This is the output pin of the MOSFET source of channel B, typically connected to the switched side of the load.

## FEATURES

Features of the SP2526 device include current limiting, a +2.7 V undervoltage lockout, overtemperature shutdown, error flag output, soft start, a switch-on resistance of $100 \mathrm{~m} \Omega$ over the entire supply range, and a supply current of $100 \mu \mathrm{~A}$. The SP2526-1 device has an activehigh enable. The SP2526-2 device has an active-low enable.

## THEORY OF OPERATION

The SP2526 is a dual integrated high-side power switch optimized for self-powered and buspowered Universal Serial Bus (USB) applications. There are operational conditions that will enable or disable one or more of the output MOSFET switches, depending on the type of hazard. Certain conditions will activate the open-drain error flag transistors forcing the flag outputs to ground. The SP2526 provides the following functions:

1) Independent Solid State Switches: Independent MOSFET switches connect the IN pin to the OUTA and OUTB pins when enabled by logic signals at the ENA and ENB control pins.
2) Fault Flag Conditions: Both the FLGA and FLGB pins are N -channel, open-drain MOSFET outputs capable of sinking a 10 mA load to typically 100 mV above ground.
3) Undervoltage Lockout: The lockout mechanism monitors the input voltage and will enable the MOSFET switches only when the voltage at the $\mathrm{V}_{\text {IN }}$ pin is greater than +2.7 V .
4) Current Limiting Threshold: A current limiting threshold prevents damage to the device and external load.
5) Thermal Shutdown: This mechanism protects the SP2526 and signals a fault condition if the die temperature exceeds $135^{\circ} \mathrm{C}$ (typical). This function has $10^{\circ} \mathrm{C}$ of hysteresis that prevents the MOSFET switches from turning on until the die temperature drops to $125^{\circ} \mathrm{C}$ (typical).

## Input and Output

The independent solid state MOSFET switches connect the USB +5.0 V supply voltage at the IN pin to the OUTA and OUTB pins when enabled by logic signals at ENA and ENB. The IN pin is the power supply connection to the device and the drain of the output MOSFET switches.

Typically, the current will flow through the switches from IN to OUTA and OUTB towards the load. If $\mathrm{V}_{\text {out }}$ is greater than $\mathrm{V}_{\text {IN }}$ when a switch is enabled, the current will flow from OUTA or OUTB to the IN pin because the MOSFET channels are bidirectional when switched on.

Under normal operating condition, the MOSFET switches will present $100 \mathrm{~m} \Omega$ maximum resistance when closed. The output MOSFETs and driver circuitry are designed to allow the MOSFET source to be externally forced to a higher voltage than the drain when the switch is off.

## Fault Flag Conditions

Fault conditions disable one or both MOSFET switches, depending on the type of fault. FLGA and FLGB are N-channel, open-drain MOSFET outputs. The active low fault flag occurs when one of the following conditions exist:

1. Undervoltage
2. Overcurrent
3. Overtemperature

## Undervoltage Lockout

This voltage lockout mechanism prevents the MOSFET switches from turning on until $\mathrm{V}_{\text {IN }}$ is greater than +2.7 V .

After the switch turns on, if the voltage at $\mathrm{V}_{\text {IN }}$ drops below +2.6 V , the lockout circuitry shuts off both switches and signals the FLGA and FLGB fault flags. In the undervoltage lockout state, the FLGA and FLGB pins will be asserted. This detection functions only when at least one MOSFET switch is enabled.

## Output Current Limiting

The SP2526 has a preset current-limit threshold. These switch control circuits will implement current limiting which prevents damage to the SP2526 and the external load while allowing a


Figure 1. Internal Block Diagram for the SP2526


Figure 2. Test Circuit for the SP2526


Figure 3. Typical Application Circuit for the SP2526
minimum current of 0.5 A to be delivered.
The SP2526 MOSFET switches will exhibit very low resistance ( $\leq 100 \mathrm{~m} \Omega$ ) or voltage drop until the current limit is reached. The fold back current is the current that is delivered into a short circuit at the output.

If the SP2526 MOSFET switch is enabled into a heavy load or short-circuit, the switch will immediately go into a constant-current mode, reducing the output voltage. The respective fault flag will pull low until the condition is removed.

When a heavy load is connected to the SP2526 switch output, a large transient current may flow until the current limiting circuitry responds.

The SP2526 will provide a low resistance switch ( $100 \mathrm{~m} \Omega$ ) between the input and output pins. This low resistance will be maintained with increasing current until the 2.2 A limit is reached. If load current exceeds this limit, the switch will increase its resistance. The foldback current ( 500 mA minimum) is reached when there is a short applied to the output. The $100 \mathrm{~m} \Omega$ switch resistance is guaranteed for all load currents, increasing or decreasing, that are below 500 mA .

## Thermal Shutdown

Under nominal load conditions, the switch resistance is very low and internal power dissipation is low. Under short circuit conditions, current is limited and internal power dissipation is higher but not extreme. Under intermediate load conditions, both the voltage across the switch and the current through the switch are at intermediate values and internal power dissipation is highest. In this last condition, the die temperature will reach the thermal limit and the switches in both channels will be shut off. As the die subsequently cools, the switch will turn on again. If the load is not removed, the device will thermal cycle in this manner to protect itself from damage.

Thermal shutdown is asserted if the die temperature exceeds $135^{\circ} \mathrm{C}$ and will not release until the die termperature drops below $125^{\circ} \mathrm{C}$. Thermal shutdown will disable both output MOSFET switches and force both FLGA and FLGB fault flags low.

The delay between a current limit fault and thermal shutdown will vary with ambient temperature, board layout, and load impedance, but is typically several hundred milliseconds. A designer
can command a USB controller to recognize the fault and disable the appropriate channel within this time.

## TYPICAL APPLICATIONS

## Bypass Capacitors

A $0.1 \mu \mathrm{~F}$ to $1.0 \mu \mathrm{~F}$ bypass capacitor from the IN pin to the GND pin is recommended to control power supply transients. Refer to Figure 4. Without a bypass capacitor, an output short may cause sufficient ringing and damage the device. Without a bypass capacitor, excessive supply lead inductance is also a concern.

Input or output transients must not exceed the absolute maximum supply voltage of $\mathrm{V}_{\mathrm{IN}(\mathrm{MAX})}=$ +6.0 V even for a short duration to avoid risk of damage to the device.

## Transient Overcurrent Filter

When the SP2526 is enabled, large values of capacitance at the output of the device will cause inrush current to exceed the short circuit currentlimit threshold of the device and assert a flag fault condition for FLGA and/or FLGB. The duration of this time will depend on the size of the output capacitance. During the capacitance charging time, the device enters into foldback mode. As the capacitance is charged, the current decreases below the current-limit threshold and the fault flags that are present at the FLGA and/ or FLGB pins will then be deasserted.

In USB applications, it is required that output bulk capacitance is utilized to support hot-plug occurences. When the SP2526 is enabled, the flag may go active for about 1 ms due to inrush current exceeding the current-limit setpoint. Additionally, during hot-plug events, inrush currents may also cause the fault flags at the FLGA and/or FLGB pins to go active. Since these conditions are not valid overcurrent faults, the USB controller must ignore the fault flags during these events. To prevent this, a 1 ms RC filter can be implemented as shown in Figure 5. Alterna-
tively, a 1 ms debounce routine may be programmed into the USB logic controller, eliminating the need for the RC filter.

## Soft Start Condition

The soft start feature of the SP2526 is implemented by holding the output turn-on rise time to 1 ms . When off, the device has high impedance MOSFET channels that slowly become low impedance as the device powers on. This prevents an inrush current from causing voltage drops that result from charging a capacitive load and can pull the USB voltage bus below specified levels. This satisfies the USB voltage droop requirements for bus-powered applications. Refer to the circuit in Figure 6.

The SP2526 can provide inrush current limiting for applications with large load capacitances where $\mathrm{C}_{\text {bULK }}>10 \mu \mathrm{~F}$. Refer to the circuit in Figure 7 for a configuration that will meet USB transient regulation specifications with large load capacitances.

## Enable Input

The ENA and ENB control pins must be driven to a logic high or logic low for a clearly defined signal input. Floating these control lines may cause unpredictable operation.

## USB Compliance

The SP2526 is ideal for self-powered and buspowered Universal Serial Bus (USB) applications. A USB port provides a +5.0 V bus and ground return line in addition to a twisted pair for data.

The SP2526 will comply with the following USB requirements:

1) The fault current is well below the UL 25 VA safety requirements;
2) The Flag Outputs are available to indicate fault conditions to USB controllers;
3) The MOSFET switches' low on-resistance meets USB voltage drop requirements;
4) Each MOSFET switch channel can supply 500 mA as required by USB downstream devices;
5) Soft start eliminates any momentary voltage drops on the upstream port that may occur when the switches are enabled in bus-powered applications.

Refer to Table 1 for a USB compliance summary of the SP2526. Additional features include the following:
6) An Undervoltage Lockout ensures that the device remains off unless there is a valid input supply voltage present;
7) +3.3 V and +5.0 V logic compatible enable inputs;
8) Thermal Shutdown prevents the possiblity of catastrophic switch failure from high-current loads;
8) The device is available in both active-high and active-low versions.

Refer to Figures 8 to 26 for typical performance characteristics of the SP2526.



Figure 4. Bypass Capacitor at the Supply Pins


Figure 5. An RC Filter for Overcurrent Faults

| USB Requirement | SP2526 Compliant Feature |
| :--- | :--- |
| Inrush current limiting required | Soft start turns on in 1 ms |
| Suspend State of $<500 \mu \mathrm{~A}$ Required | Suspend Current of $1 \mu \mathrm{~A}$ maximum |
| Bus powered hubs must have 350 mV maximum <br> drop from cable plug to port | Switch on resistance of $140 \mathrm{~m} \Omega$ maximum <br> (translates to 70 mV at 500 mA$)$ |
| Voltage supplied to host or hub port is +4.75 V to +5.25 V | Operating range of +3.0 V to +5.5 V |
| A device that draws bus power must have a stable <br> supply within 100 ms of $\mathrm{V}_{\text {Bus }}$ reaching +4.4 V | Turns on in 1 ms |
| Over-current reporting capability required | Open drain fault flags |

Table 1. USB Protocol Compliance of the SP2526 device


Figure 6. Soft Start Circuit Configuration for a Single Channel USB-powered Application with the SP2526


Figure 7. Soft Start Circuit Configuration for SP2526 Applications with Large Load Capacitances

PERFORMANCE CHARACTERISTICS
$\mathrm{V}_{\text {IN }}=+5.0 \mathrm{~V}$, single MOSFET switch section, and $\mathrm{T}_{\text {AMB }}=+25^{\circ} \mathrm{C}$ unless otherwise noted.


Figure 8. Output On-Resistance vs. Supply Voltage


Figure 10. Undervoltage Threshold Voltage vs. Temperature


Figure 9. Output On-Resistance vs. Temperature


Figure 11. Control Threshold vs. Supply Voltage


Figure 12. On-state Supply Current vs. Supply Voltage


Figure 14. On-state Supply Current vs. Temperature


Figure 13. Off-state Supply Current vs. Supply Voltage

Figure 15. Off-state Supply Current vs. Temperature


Figure 16. Control Threshold vs. Temperature


Figure 18. Current-Limit Response With a Ramped Load


Figure 17. Input Voltage Response

Figure 19. Turn-on/Turn-off Characteristics Where $R_{L}=35 \Omega$ and $C_{L}=10 \mu F$


Figure 20. Turn-on/Turn-off Characteristics where $R_{L}=$ $35 \Omega$ and $C_{L}=150 \mu F$


Figure 22. Short Circuit Response (Short Applied To Output)


Figure 21. Short Circuit Response (Enable into Short Circuit)


Figure 23. Short Circuit Transient Response (Short Applied To Output)


## PERFORMANCE CHARACTERISTICS (continued)

$\mathrm{V}_{\mathrm{IN}}=+5.0 \mathrm{~V}$, single MOSFET switch section, and $\mathrm{T}_{\mathrm{AMB}}=+25^{\circ} \mathrm{C}$ unless otherwise noted.


Figure 24. Short Circuit Response (Enable into Short Circuit)


Figure 25. Short Circuit Response (Short Circuit applied at Output)

PACKAGE: PLASTIC SMALL OUTLINE (SOIC) (NARROW)


| DIMENSIONS (Inches) Minimum/Maximum (mm) | 8-PIN | 14-PIN | 16-PIN |
| :---: | :---: | :---: | :---: |
| A | $\begin{gathered} 0.053 / 0.069 \\ (1.346 / 1.748) \end{gathered}$ | $\begin{gathered} 0.053 / 0.069 \\ (1.346 / 1.748) \end{gathered}$ | $\begin{gathered} 0.053 / 0.069 \\ (1.346 / 1.748) \end{gathered}$ |
| A1 | $\begin{aligned} & 0.004 / 0.010 \\ & \mathbf{( 0 . 1 0 2 / 0 . 2 4 9} \end{aligned}$ | $\begin{gathered} 0.004 / 0.010 \\ (0.102 / 0.249) \end{gathered}$ | $\begin{gathered} 0.004 / 0.010 \\ (0.102 / 0.249) \end{gathered}$ |
| B | $\begin{aligned} & \text { 0.014/0.019 } \\ & (0.35 / 0.49) \end{aligned}$ | $\begin{gathered} 0.013 / 0.020 \\ (0.330 / 0.508) \end{gathered}$ | $\begin{gathered} 0.013 / 0.020 \\ (0.330 / 0.508) \end{gathered}$ |
| D | $\begin{aligned} & 0.189 / 0.197 \\ & (4.80 / 5.00) \end{aligned}$ | $\begin{gathered} 0.337 / 0.344 \\ (8.552 / 8.748) \end{gathered}$ | $\begin{array}{\|c\|} \hline 0.386 / 0.394 \\ (9.802 / 10.000) \end{array}$ |
| E | $\begin{gathered} 0.150 / 0.157 \\ (3.802 / 3.988) \end{gathered}$ | $\begin{gathered} 0.150 / 0.157 \\ (3.802 / 3.988) \end{gathered}$ | $\begin{gathered} 0.150 / 0.157 \\ (3.802 / 3.988) \end{gathered}$ |
| e | $\begin{aligned} & 0.050 \mathrm{BSC} \\ & \text { (1.270 BSC) } \end{aligned}$ | $\begin{gathered} 0.050 \mathrm{BSC} \\ \text { (1.270 BSC) } \end{gathered}$ | $\begin{aligned} & 0.050 \mathrm{BSC} \\ & \text { (1.270 BSC) } \end{aligned}$ |
| H | $\begin{gathered} 0.228 / 0.244 \\ (5.801 / 6.198) \end{gathered}$ | $\begin{gathered} 0.228 / 0.244 \\ (5.801 / 6.198) \end{gathered}$ | $\begin{gathered} 0.228 / 0.244 \\ (5.801 / 6.198) \end{gathered}$ |
| h | $\begin{gathered} 0.010 / 0.020 \\ (0.254 / 0.498) \end{gathered}$ | $\begin{gathered} 0.010 / 0.020 \\ (0.254 / 0.498) \end{gathered}$ | $\begin{gathered} 0.010 / 0.020 \\ (0.254 / 0.498) \end{gathered}$ |
| L | $\begin{gathered} 0.016 / 0.050 \\ (0.406 / 1.270) \end{gathered}$ | $\begin{gathered} 0.016 / 0.050 \\ (0.406 / 1.270) \end{gathered}$ | $\begin{gathered} 0.016 / 0.050 \\ (0.406 / 1.270) \end{gathered}$ |
| $\varnothing$ | $\begin{gathered} 0^{\circ} / 8^{\circ} \\ \left(0^{\circ} / 8^{\circ}\right) \end{gathered}$ | $\begin{gathered} 0^{\circ} / 8^{\circ} \\ \left(0^{\circ} / 8^{\circ}\right) \end{gathered}$ | $\begin{gathered} 0^{\circ} / 8^{\circ} \\ \left(0^{\circ} / 8^{\circ}\right) \end{gathered}$ |



PLASTIC
DUAL-IN-LINE
(NARROW)

| DIMENSIONS (Inches) Minimum/Maximum (mm) | 8-PIN | 14-PIN | 16-PIN | 18-PIN | 20-PIN | 22-PIN |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A2 | $\begin{gathered} 0.115 / 0.195 \\ (2.921 / 4.953) \end{gathered}$ | $\begin{gathered} 0.115 / 0.195 \\ (2.921 / 4.953) \end{gathered}$ | $\begin{gathered} 0.115 / 0.195 \\ (2.921 / 4.953) \end{gathered}$ | $\begin{gathered} 0.115 / 0.195 \\ (2.921 / 4.953) \end{gathered}$ | $\begin{gathered} 0.115 / 0.195 \\ (2.921 / 4.953) \end{gathered}$ | $\begin{gathered} 0.115 / 0.195 \\ (2.921 / 4.953) \end{gathered}$ |
| B | $\begin{gathered} 0.014 / 0.022 \\ (0.356 / 0.559) \end{gathered}$ | $\begin{gathered} 0.014 / 0.022 \\ (0.356 / 0.559) \end{gathered}$ | $\begin{gathered} 0.014 / 0.022 \\ (0.356 / 0.559) \end{gathered}$ | $\begin{gathered} 0.014 / 0.022 \\ (0.356 / 0.559) \end{gathered}$ | $\begin{gathered} 0.014 / 0.022 \\ (0.356 / 0.559) \end{gathered}$ | $\begin{gathered} 0.014 / 0.022 \\ (0.356 / 0.559) \end{gathered}$ |
| B1 | $\begin{gathered} 0.045 / 0.070 \\ (1.143 / 1.778) \end{gathered}$ | $\begin{gathered} 0.045 / 0.070 \\ (1.143 / 1.778) \end{gathered}$ | $\begin{gathered} 0.045 / 0.070 \\ (1.143 / 1.778) \end{gathered}$ | $\begin{gathered} 0.045 / 0.070 \\ (1.143 / 1.778) \end{gathered}$ | $\begin{gathered} 0.045 / 0.070 \\ (1.143 / 1.778) \end{gathered}$ | $\begin{gathered} 0.045 / 0.070 \\ (1.143 / 1.778) \end{gathered}$ |
| C | $\begin{gathered} 0.008 / 0.014 \\ (0.203 / 0.356) \end{gathered}$ | $\begin{gathered} 0.008 / 0.014 \\ (0.203 / 0.356) \end{gathered}$ | $\begin{gathered} 0.008 / 0.014 \\ (0.203 / 0.356) \end{gathered}$ | $\begin{gathered} \hline 0.008 / 0.014 \\ (0.203 / 0.356) \end{gathered}$ | $\begin{gathered} 0.008 / 0.014 \\ (0.203 / 0.356) \end{gathered}$ | $\begin{gathered} 0.008 / 0.014 \\ (0.203 / 0.356) \end{gathered}$ |
| D | $\begin{gathered} \hline 0.355 / 0.325 \\ (9.017 / 10.160) \end{gathered}$ | $0.735 / 0.775$ $(18.669 / 19.685)$ | $0.780 / 0.800$ $(19.812 / 20.320)$ | $0.880 / 0.920$ $(22.352 / 23.368)$ | $0.980 / 1.060$ $(24.892 / 26.924)$ | $\begin{gathered} 1.145 / 1.155 \\ (29.083 / 29.337) \end{gathered}$ |
| E | $\begin{gathered} 0.300 / 0.325 \\ (7.620 / 8.255) \end{gathered}$ | $\begin{gathered} 0.300 / 0.325 \\ (7.620 / 8.255) \end{gathered}$ | $\begin{gathered} 0.300 / 0.325 \\ (7.620 / 8.255) \end{gathered}$ | $\begin{gathered} 0.300 / 0.325 \\ (7.620 / 8.255) \end{gathered}$ | $\begin{gathered} 0.300 / 0.325 \\ (7.620 / 8.255) \end{gathered}$ | $\begin{gathered} 0.300 / 0.325 \\ (7.620 / 8.255) \end{gathered}$ |
| E1 | $\begin{gathered} 0.240 / 0.280 \\ (6.096 / 7.112) \end{gathered}$ | $\begin{gathered} 0.240 / 0.280 \\ (6.096 / 7.112) \end{gathered}$ | $\begin{gathered} 0.240 / 0.280 \\ (6.096 / 7.112) \end{gathered}$ | $\begin{gathered} 0.240 / 0.280 \\ (6.096 / 7.112) \end{gathered}$ | $\begin{gathered} 0.240 / 0.280 \\ (6.096 / 7.112) \end{gathered}$ | $\begin{gathered} 0.240 / 0.280 \\ (6.096 / 7.112) \end{gathered}$ |
| L | $\begin{gathered} 0.115 / 0.150 \\ (2.921 / 3.810) \end{gathered}$ | $\begin{gathered} 0.115 / 0.150 \\ (2.921 / 3.810) \end{gathered}$ | $\begin{gathered} 0.115 / 0.150 \\ (2.921 / 3.810) \end{gathered}$ | $\begin{gathered} 0.115 / 0.150 \\ (2.921 / 3.810) \end{gathered}$ | $\begin{gathered} 0.115 / 0.150 \\ (2.921 / 3.810) \end{gathered}$ | $\begin{gathered} 0.115 / 0.150 \\ (2.921 / 3.810) \end{gathered}$ |
| $\varnothing$ | $\begin{gathered} 0^{\circ} / 15^{\circ} \\ \left(0^{\circ} / 15^{\circ}\right) \end{gathered}$ | $\begin{gathered} 0 \% 15^{\circ} \\ \left(0^{\circ} / 15^{\circ}\right) \end{gathered}$ | $\begin{gathered} 0^{\circ} / 15^{\circ} \\ \left(0^{\circ} / 15^{\circ}\right) \end{gathered}$ | $\begin{gathered} 0^{\circ} / 15^{\circ} \\ \left(0^{\circ} / 15^{\circ}\right) \end{gathered}$ | $\begin{gathered} 0^{\circ} / 15^{\circ} \\ \left(0^{\circ} / 15^{\circ}\right) \end{gathered}$ | $\begin{gathered} 0^{\circ} / 15^{\circ} \\ \left(0^{\circ} / 15^{\circ}\right) \end{gathered}$ |



Please consult the factory for pricing and availability on a Tape-On-Reel option.

## Sipex

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