

VOLTAGE REGULATOR WITH ON/OFF SWITCH

FEATURES

- High Voltage Precision at $\pm 2.0\%$
- Active Low On/Off Control
- Very Low Dropout Voltage 80 mV at 30 mA
- Very Low Noise
- Very Small SOT23L or SOT89 Surface Mount Packages
- Internal Thermal Shutdown
- Short Circuit Protection

DESCRIPTION

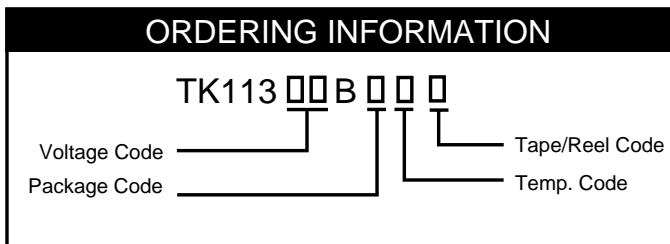
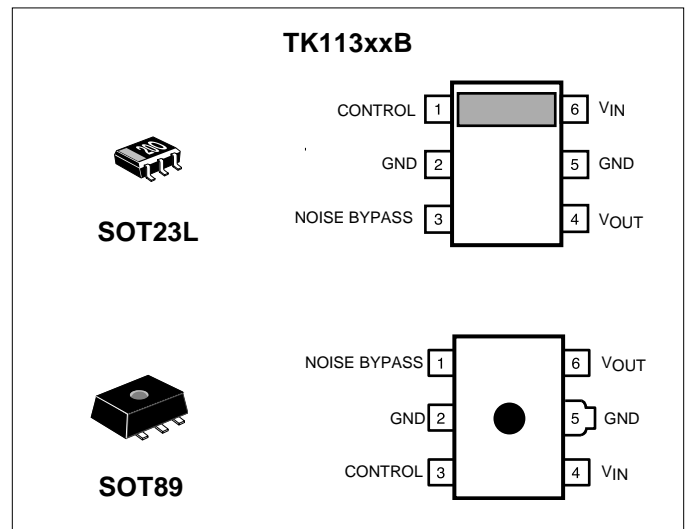
The TK113xxB is a low dropout linear regulator with a built-in electronic switch. The device is in the "on" state when the control pin is pulled to a low level. An external capacitor can be connected to the noise bypass pin to lower the output noise level to 30 μ Vrms.

An internal PNP pass transistor is used to achieve a low dropout voltage of 80 mV (typ.) at 30 mA load current. The TK113xxB has a very low quiescent current of 170 μ A at no load and 1 mA with a 30 mA load. The standby current is typically 100 nA. The internal thermal shut down circuitry limits the junction temperature to below 150 °C. The load current is internally monitored and the device will shut down in the presence of a short circuit or overcurrent condition at the output.

APPLICATIONS

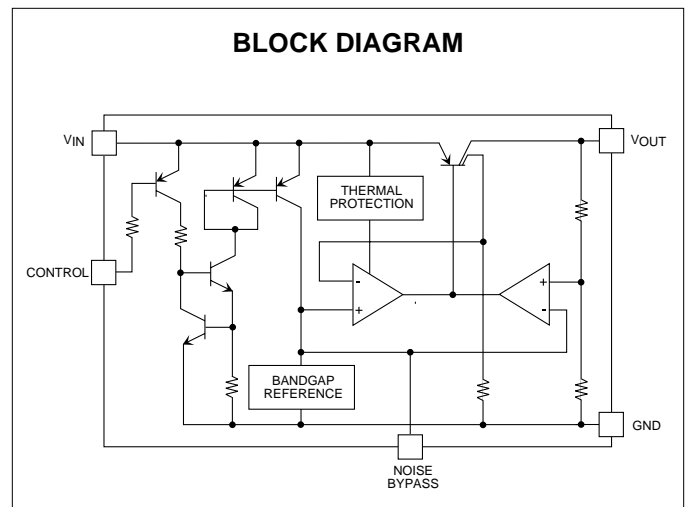
- Battery Powered Systems
- Cellular Telephones
- Pagers
- Personal Communications Equipment
- Portable Instrumentation
- Portable Consumer Equipment
- Radio Control Systems
- Toys
- Low Voltage Systems

The TK113xxB is available in either 6-pin SOT23L or 5-pin SOT89 surface mount packages.



VOLTAGE CODE	PACKAGE CODE:	TAPE/REEL CODE
20 = 2.0 V	M: SOT23L	L: Tape Left (SOT23L)
21 = 2.1 V	U: SOT89	B: Tape Left (SOT89)
22 = 2.2 V		
23 = 2.3 V		
24 = 2.4 V		
25 = 2.5 V		
26 = 2.6 V		
27 = 2.7 V		
28 = 2.8 V		
29 = 2.9 V		
30 = 3.0 V		
31 = 3.1 V		
32 = 3.2 V		
33 = 3.3 V		
34 = 3.4 V		
35 = 3.5 V		
36 = 3.6 V		
37 = 3.7 V		
38 = 3.8 V		
39 = 3.9 V		
40 = 4.0 V		
41 = 4.1 V		
42 = 4.2 V		
43 = 4.3 V		
44 = 4.4 V		
45 = 4.5 V		
46 = 4.6 V		
47 = 4.7 V		
48 = 4.8 V		
49 = 4.9 V		
50 = 5.0 V		
55 = 5.5 V		
60 = 6.0 V		
80 = 8.0 V		

TEMP. CODE:
 C: -30 to +80 °C
 I: -40 to +85 °C



TK113xxBM/U

ABSOLUTE MAXIMUM RATINGS ($V_{OUT} \geq 2.0$ V)

Supply Voltage	16 V	Storage Temperature Range	-55 to +150 °C
Output Current	260 mA	Operating Temperature Range	-30 to +80 °C
Power Dissipation SOT-23L (Note 1)	600 mW	Voltage Range	1.8 to 14.5 V
Power Dissipation SOT-23L (Note 1)	900 mW	Operating Junction Temperature	150 °C
Reverse Bias	10 V		

TK113xxBM/UC ELECTRICAL CHARACTERISTICS ($V_{OUT} \geq 2.0$ V)

Test conditions: $T_A = 25$ °C, unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
I_Q	Quiescent Current	$I_{OUT} = 0$ mA, Excluding I_{CONT}		170	250	μ A
I_{STBY}	Standby Current	$V_{IN} = 8$ V, Output OFF			0.1	μ A
V_{OUT}	Output Voltage	$I_{OUT} = 30$ mA	See Table 1			V
Line Reg	Line Regulation	$V_{OUT} \leq 5.5$ V, (Note 2)		3.0	20	mV
		$V_{OUT} \geq 5.6$ V, (Note 2)		15	40	mV
Load Reg	Load Regulation	$I_{OUT} = 1$ to 60 mA, (Note 2)		6	30	mV
		$I_{OUT} = 1$ to 100 mA, (Note 2)		18	60	mV
		$I_{OUT} = 1$ to 150 mA, (Note 2)		23	90	mV
V_{DROP}	Dropout Voltage	$I_{OUT} = 60$ mA, (Note 2)		0.12	0.20	V
		$I_{OUT} = 150$ mA, (Note 2)		0.26	0.39	V
I_{OUT}	Continuous Output Current	(Note 2)			150	mA
$I_{OUT(PULSE)}$	Pulse Output Current	5 ms pulse, 12.5 % duty cycle			180	mA
RR	Ripple Rejection	$f = 400$ Hz, $C_L = 10$ μ F, $C_N = 0.1$ μ F, $V_{IN} = V_{OUT} + 1.5$ V, $I_{OUT} = 30$ mA, $V_{RIPPLE} = 100$ mVrms, (Note 3)		60		dB
V_{NO}	Output Noise Voltage	10 Hz $\leq f \leq 80$ kHz, $C_L = 10$ μ F, $C_N = 0.1$ μ F, $V_{CN} = V_{OUT} + 1.5$ V, $I_{OUT} = 60$ mA, (Notes 3,4)		30		μ Vrms
V_{ref}	Noise Bypass Terminal Voltage			1.25		V
$\Delta V_{OUT} / \Delta T$	Temperature Coefficient	$I_{OUT} = 10$ mA		40		ppm/° C

CONTROL TERMINAL SPECIFICATIONS

I_{CONT}	Control Current	$V_{CONT} = 1.8$ V, Output ON		12	35	μ A
$V_{CONT(ON)}$	Control Voltage ON	Output ON			$V_{CC} - 1.8$	V
$V_{CONT(OFF)}$	Control Voltage OFF	Output OFF	$V_{CC} - 0.6$			V

Note 1: When mounted as recommended. Derate at 4.8 mW/°C for SOT-23L and 6.4 mW/°C for SOT-89 packages for operation above 25°C.

Note 2: Refer to "Definition of Terms."

Note 3: Ripple rejection and noise voltage are affected by the value and characteristics of the capacitor used.

Note 4: Output noise voltage can be reduced by connecting a capacitor to a noise pass terminal.

Gen. Note: Parameters with min. or max. values are 100% tested at $T_A = 25$ °C.

ABSOLUTE MAXIMUM RATINGS TK113xxBI ($V_{OUT} \geq 2.5 V$)

Supply Voltage	16 V	Storage Temperature Range	-55 to +150 °C
Power Dissipation SOT-23L (Note1)	600 mW	Operating Temperature Range	-40 to +85 °C
Power Dissipation SOT-89 (Note1)	900 mW	Operating Voltage Range	1.8 to 14.5 V
Reverse Bias	10 V	Junction Temperature	150 °C

TK113xxBM/UI ELECTRICAL CHARACTERISTICS ($V_{OUT} \geq 2.5 V$)

Test conditions: $T_A = -40$ to 85 °C, unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
I_Q	Quiescent Current	$I_{OUT} = 0$ mA, Excluding I_{CONT}		170	300	μ A
I_{STBY}	Standby Current	$V_{IN} = 8$ V, Output OFF			0.2	μ A
V_{OUT}	Output Voltage	$I_{OUT} = 30$ mA	See Table 2			V
Line Reg	Line Regulation	$V_{OUT} \leq 5.5$ V, (Note 2)		3.0	25	mV
		$V_{OUT} = 5.6$ V, (Note 2)		15	40	mV
Load Reg	Load Regulation	$I_{OUT} = 1$ to 60 mA, (Note 2)		6	40	mV
		$I_{OUT} = 1$ to 100 mA, (Note 2)		18	80	mV
		$I_{OUT} = 1$ to 150 mA, (Note 2)		23	110	mV
V_{DROP}	Dropout Voltage	$I_{OUT} = 60$ mA, (Note 2)		0.12	0.23	V
		$I_{OUT} = 150$ mA, (Note 2)		0.26	0.40	V
I_{OUT}	Continuous Output Current	(Note 2)			150	mA
$I_{OUT(PULSE)}$	Pulse Output Current	5 ms pulse, 12.5 % duty cycle			180	mA
RR	Ripple Rejection	$f = 400$ Hz, $C_L = 10$ μ F, $C_N = 0.1$ μ F, $V_{IN} = V_{OUT} + 1.5$ V, $I_{OUT} = 30$ mA, $V_{RIPPLE} = 100$ mVrms, (Note 3)		60		dB
V_{NO}	Output Noise Voltage	10 Hz $\leq f \leq 80$ kHz, $C_L = 10$ μ F, $C_N = 0.1$ μ F, $V_{CN} = V_{OUT} + 1.5$ V, $I_{OUT} = 60$ mA, (Notes 3,4)		30		μ Vrms
V_{ref}	Noise Bypass Terminal Voltage			1.25		V
$\Delta V_{OUT} / \Delta T$	Temperature Coefficient	$I_{OUT} = 10$ mA		40		ppm/° C

CONTROL TERMINAL SPECIFICATIONS

I_{CONT}	Control Current	$V_{CONT} = 1.8$ V, Output ON		12	40	μ A
$V_{CONT(ON)}$	Control Voltage ON	Output ON			$V_{CC} - 2.0$	V
$V_{CONT(OFF)}$	Control Voltage OFF	Output OFF	$V_{CC} - 0.5$			V

Note 1: When mounted as recommended. Derate at 4.8 mW/°C for SOT-23L and 6.4 mW/°C for SOT-89 packages for operation above 25 °C.

Note 2: Refer to "Definition of Terms."

Note 3: Ripple rejection and noise voltage are affected by the value and characteristics of the capacitor used.

Note 4: Output noise voltage can be reduced by connecting a capacitor to a noise pass terminal.

Gen Note: Parameters with min. or max. values are 100% tested at $T_A = 25$ °C.

Gen Note: For Line Regulation, typ. and max. is changed to $V_{OUT} > 5.6$ V.

TK113xxBM/U

TK113xxBM/UC ELECTRICAL CHARACTERISTICS TABLE 1

Test conditions: $T_A = 25\text{ }^\circ\text{C}$, $I_{OUT} = 30\text{ mA}$, unless otherwise specified.

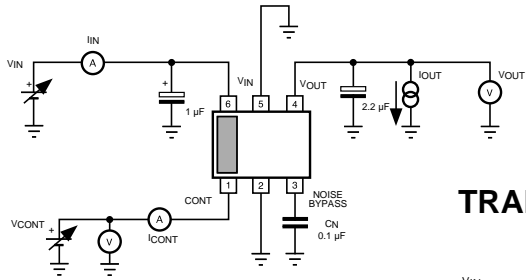
Output Voltage	Voltage Code	$V_{OUT(MIN)}$	$V_{OUT(MAX)}$	Test Voltage	Output Voltage	Voltage Code	$V_{OUT(MIN)}$	$V_{OUT(MAX)}$	Test Voltage
2.0 V	20	1.940 V	2.060 V	3.0 V	3.7 V	37	3.630 V	3.770 V	4.7 V
2.1 V	21	2.040 V	2.160 V	3.1 V	3.8 V	38	3.725 V	3.875 V	4.8 V
2.2 V	22	2.140 V	2.260 V	3.2 V	3.9 V	39	3.825 V	3.975 V	4.9 V
2.3 V	23	2.240 V	2.360 V	3.3 V	4.0 V	40	3.920 V	4.080 V	5.0 V
2.4 V	24	2.340 V	2.460 V	3.4 V	4.1 V	41	4.020 V	4.180 V	5.1 V
2.5 V	25	2.440 V	2.560 V	3.5 V	4.2 V	42	4.120 V	4.280 V	5.2 V
2.6 V	26	2.540 V	2.660 V	3.6 V	4.3 V	43	4.215 V	4.385 V	5.3 V
2.7 V	27	2.640 V	2.760 V	3.7 V	4.4 V	44	4.315 V	4.485 V	5.4 V
2.8 V	28	2.740 V	2.860 V	3.8 V	4.5 V	45	4.410 V	4.590 V	5.5 V
2.9 V	29	2.840 V	2.960 V	3.9 V	4.6 V	46	4.510 V	4.690 V	5.6 V
3.0 V	30	2.940 V	3.060 V	4.0 V	4.7 V	47	4.605 V	4.795 V	5.7 V
3.1 V	31	3.040 V	3.160 V	4.1 V	4.8 V	48	4.705 V	4.895 V	5.8 V
3.2 V	32	3.140 V	3.260 V	4.2 V	4.9 V	49	4.800 V	5.000 V	5.9 V
3.3 V	33	3.240 V	3.360 V	4.3 V	5.0 V	50	4.900 V	5.100 V	6.0 V
3.4 V	34	3.335 V	3.465 V	4.4 V	5.5 V	55	5.390 V	5.610 V	6.5 V
3.5 V	35	3.435 V	3.565 V	4.5 V	6.0 V	60	5.880 V	6.120 V	7.0 V
3.6 V	36	3.535 V	3.665 V	4.6 V	8.0 V	80	7.840 V	8.160 V	9.0 V

TK113xxBM/UI ELECTRICAL CHARACTERISTICS TABLE 2Test Conditions: $V_{IN} = V_{OUT(TYP)} + 1\text{ V}$, $I_{OUT} = 30\text{ mA}$, unless otherwise specified.

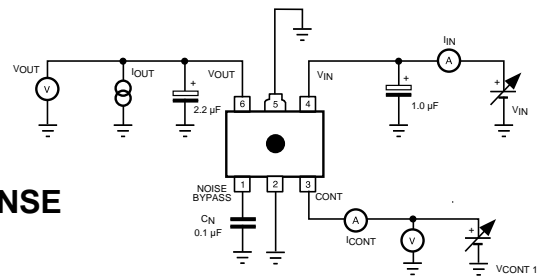
Output Voltage	Voltage Code	Room Temp. Range ($T_A = 25\text{ }^\circ\text{C}$)		Full Temp. Range ($T_A = -40\text{ to }+85\text{ }^\circ\text{C}$)	
		$V_{OUT(MIN)}$	$V_{OUT(MAX)}$	$V_{OUT(MIN)}$	$V_{OUT(MAX)}$
2.5 V	25	2.440 V	2.560 V	2.400 V	2.600 V
2.6 V	26	2.540 V	2.660 V	2.500 V	2.700 V
2.7 V	27	2.640 V	2.760 V	2.600 V	2.800 V
2.8 V	28	2.750 V	2.860 V	2.700 V	2.900 V
2.9 V	29	2.840 V	2.960 V	2.800 V	3.000 V
3.0 V	30	2.940 V	3.060 V	2.900 V	3.100 V
3.1 V	31	3.040 V	3.160 V	3.000 V	3.200 V
3.2 V	32	3.140 V	3.260 V	3.095 V	3.305 V
3.3 V	33	3.240 V	3.360 V	3.190 V	3.410 V
3.4 V	34	3.335 V	3.465 V	3.290 V	3.510 V
3.5 V	35	3.435 V	3.565 V	3.385 V	3.615 V
3.6 V	36	3.535 V	3.665 V	3.485 V	3.720 V
3.7 V	37	3.630 V	3.770 V	3.580 V	3.820 V
3.8 V	38	3.725 V	3.875 V	3.675 V	3.925 V
3.9 V	39	3.825 V	3.975 V	3.770 V	4.030 V
4.0 V	40	3.920 V	4.080 V	3.870 V	4.130 V
4.1 V	41	4.020 V	4.180 V	3.965 V	4.235 V
4.2 V	42	4.120 V	4.280 V	4.060 V	4.335 V
4.3 V	43	4.215 V	4.385 V	4.160 V	4.440 V
4.4 V	44	4.315 V	4.485 V	4.255 V	4.545 V
4.5 V	45	4.410 V	4.590 V	4.350 V	4.645 V
4.6 V	46	4.510 V	4.690 V	4.450 V	4.750 V
4.7 V	47	4.605 V	4.795 V	4.545 V	4.850 V
4.8 V	48	4.705 V	4.895 V	4.640 V	4.955 V
4.9 V	49	4.800 V	5.000 V	4.740 V	5.060 V
5.0 V	50	4.900 V	5.100 V	4.835 V	5.165 V
5.5 V	55	5.390 V	5.610 V	5.320 V	5.680 V
6.0 V	60	5.880 V	6.120 V	5.805 V	6.195 V
8.0 V	80	7.840 V	8.160 V	7.745 V	8.265 V

TEST CIRCUITS

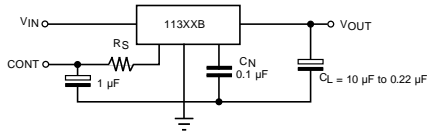
SOT23L



SOT89



TRANSIENT RESPONSE

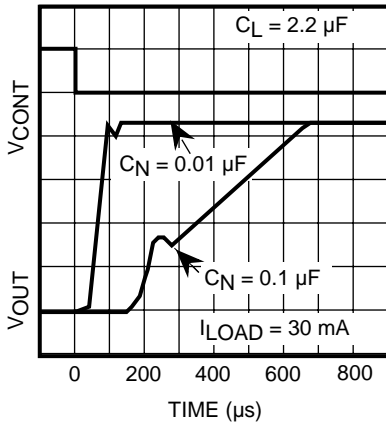


Note: Connect pin 5 to ground for heat sink

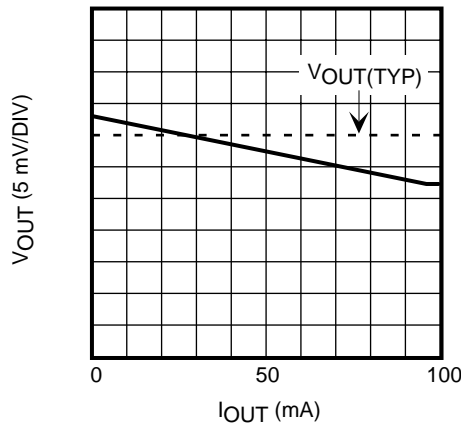
TYPICAL PERFORMANCE CHARACTERISTICS

$T_A = 25^\circ\text{C}$, unless otherwise specified.

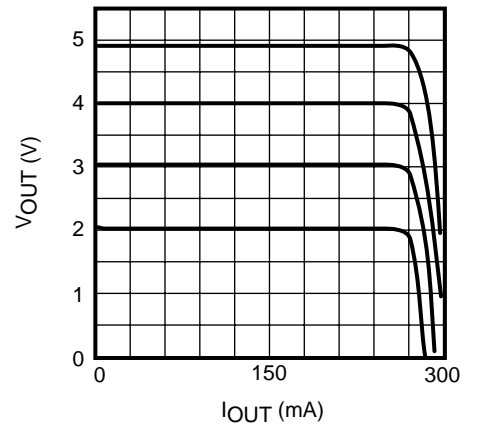
OUTPUT VOLTAGE RESPONSE (OFF→ON)



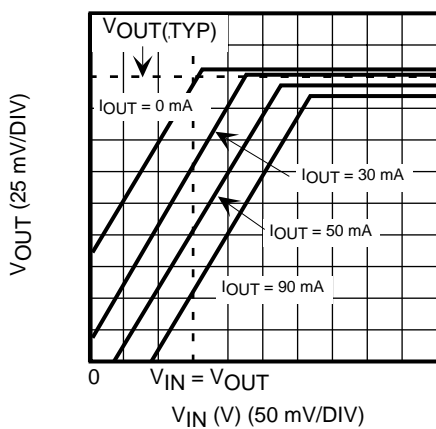
LOAD REGULATION



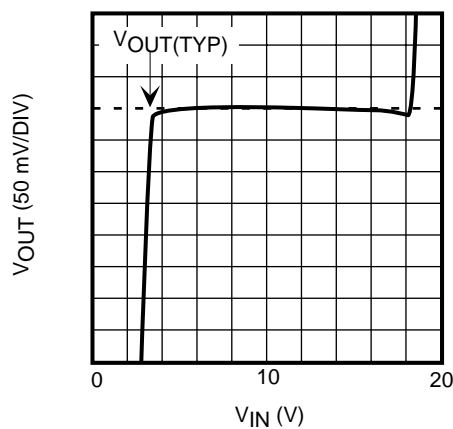
SHORT CIRCUIT CURRENT



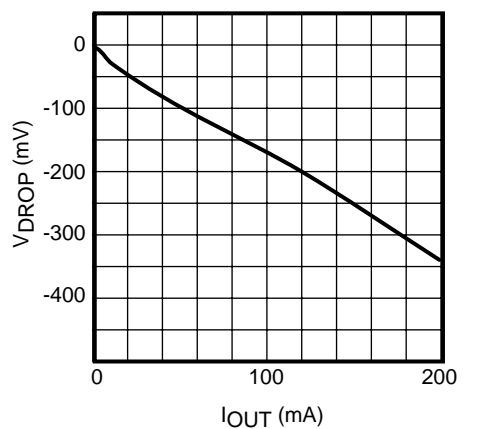
OUTPUT VOLTAGE vs. INPUT VOLTAGE



LINE REGULATION



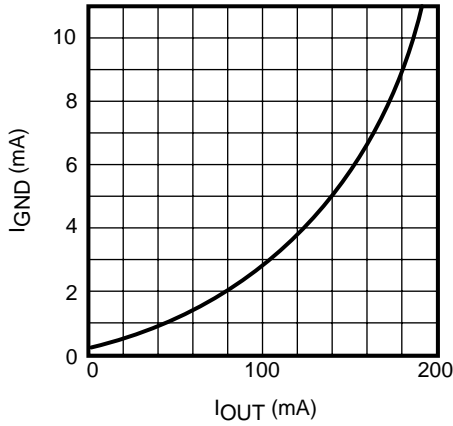
DROPOUT VOLTAGE vs. OUTPUT CURRENT



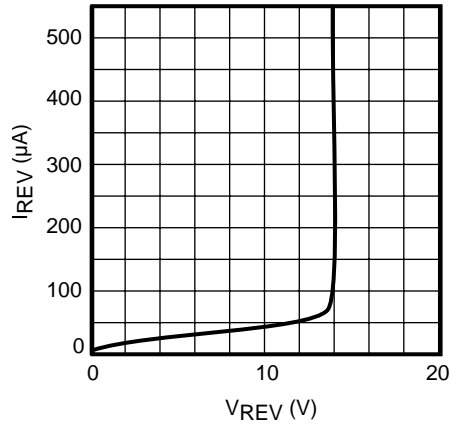
TYPICAL PERFORMANCE CHARACTERISTICS (CONT.)

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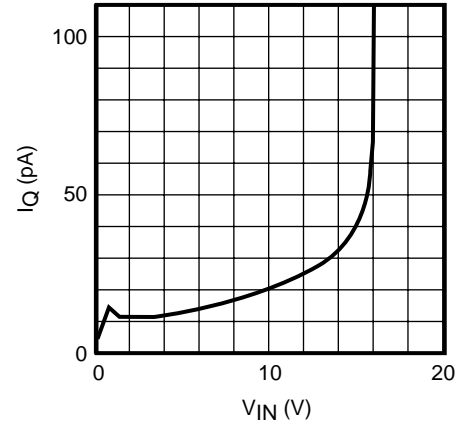
GROUND CURRENT vs. OUTPUT CURRENT



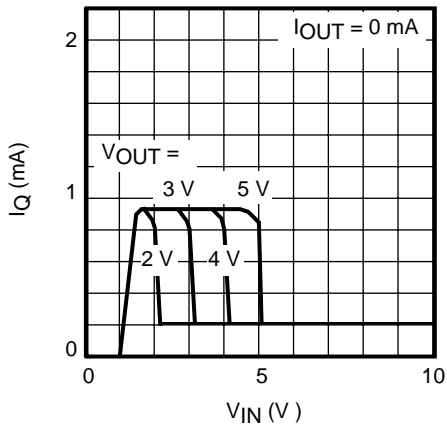
REVERSE BIAS CURRENT ($V_{IN} = 0\text{ V}$)



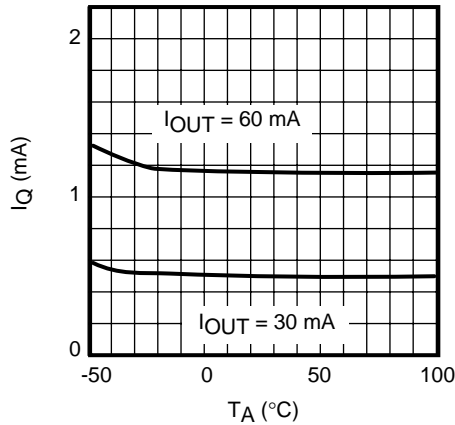
QUIESCENT CURRENT (OFF MODE) vs. INPUT VOLTAGE



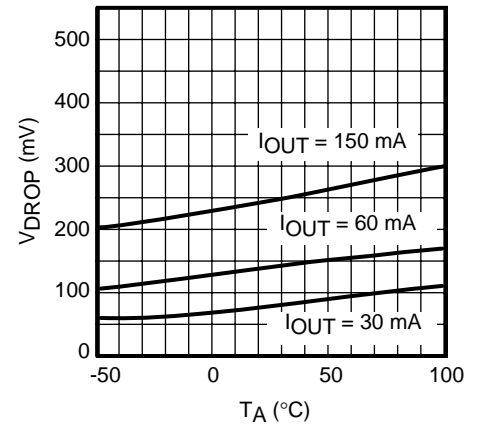
QUIESCENT CURRENT (ON MODE) vs. INPUT VOLTAGE



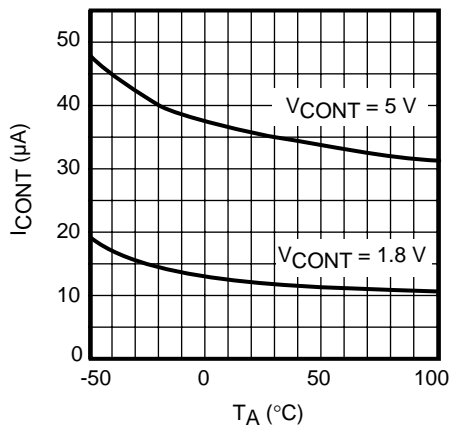
QUIESCENT CURRENT



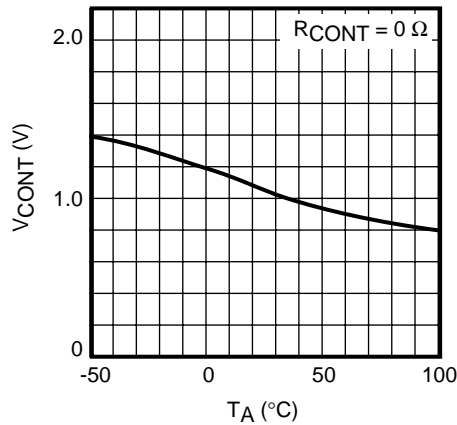
DROPOUT VOLTAGE



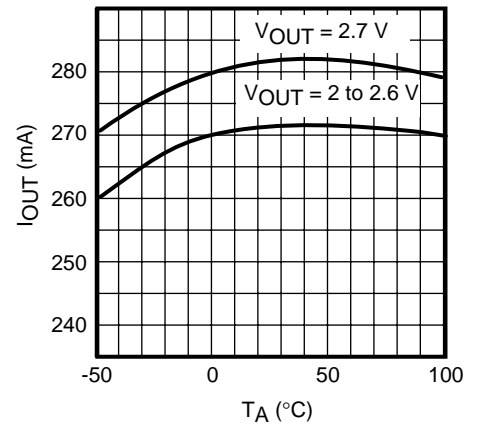
CONTROL CURRENT



V_{CONT} (V_{OUT} ON POINT)

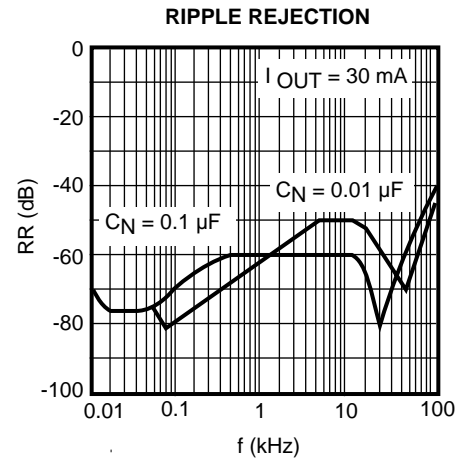
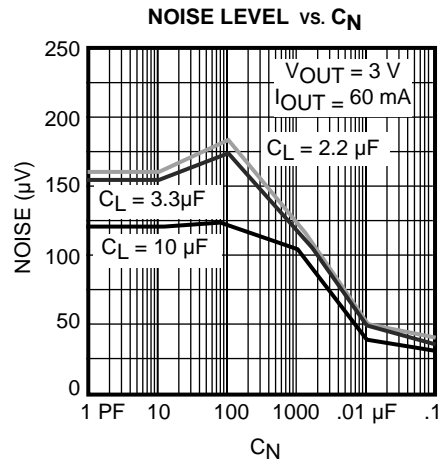
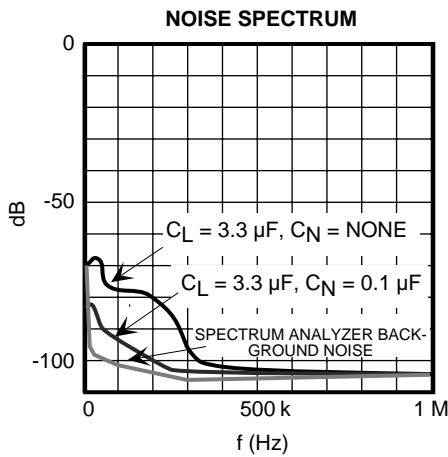
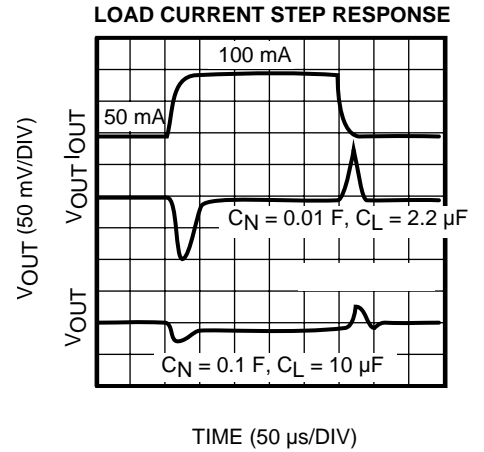
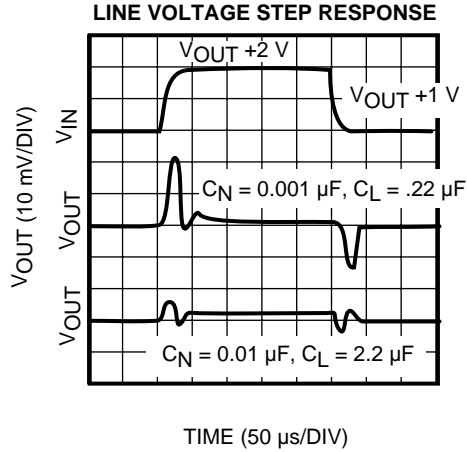
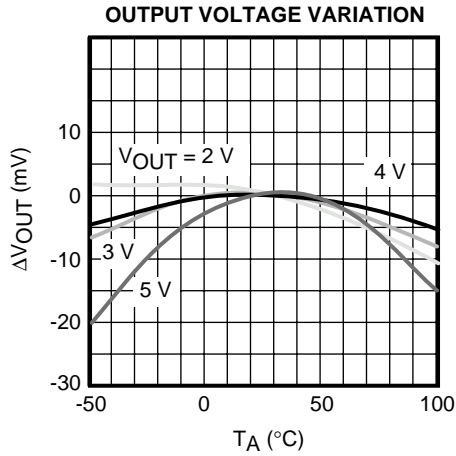


MAXIMUM OUTPUT CURRENT



TYPICAL PERFORMANCE CHARACTERISTICS (CONT.)

$T_A = 25\text{ }^\circ\text{C}$, unless otherwise specified.



DEFINITION AND EXPLANATION OF TECHNICAL TERMS

OUTPUT VOLTAGE (V_{OUT})

The output voltage is specified with $V_{IN} = (V_{OUT(TYP)} + 1 \text{ V})$ and $I_{OUT} = 30 \text{ mA}$.

DROPOUT VOLTAGE (V_{DROP})

The dropout voltage is the difference between the input voltage and the output voltage at which point the regulator starts to fall out of regulation. Below this value, the output voltage will fall as the input voltage is reduced. It is dependent upon the load current and the junction temperature.

OUTPUT CURRENT ($I_{OUT(MAX)}$)

This is the maximum continuous output current as specified under the condition where the output voltage drops 0.3 V below the value specified with $I_{OUT} = 30 \text{ mA}$. The input voltage is set to $V_{OUT} + 1 \text{ V}$, and the current is pulsed to minimize temperature effect.

CONTINUOUS OUTPUT CURRENT (I_{OUT})

Normal operating output current. This is limited by package power dissipation.

PULSE OUTPUT CURRENT ($I_{OUT(PULSE)}$)

Max pulse width 5 ms, Duty cycle 12.5%: pulse load only.

LINE REGULATION (Line Reg)

Line regulation is the ability of the regulator to maintain a constant output voltage as the input voltage changes. The line regulation is specified as the input voltage is changed from $V_{IN} = V_{OUT(TYP)} + 1 \text{ V}$ to $V_{IN} = V_{OUT(TYP)} + 6 \text{ V}$.

LOAD REGULATION (Load Reg)

Load regulation is the ability of the regulator to maintain a constant output voltage as the load current changes. It is a pulsed measurement to minimize temperature effects with the input voltage set to $V_{IN} = V_{OUT} + 1 \text{ V}$. The load regulation is specified under two output current step conditions of 1 mA to 60 mA and 1 mA to 100 mA.

QUIESCENT CURRENT (I_Q)

The quiescent current is the current which flows through the ground terminal under no load conditions ($I_{OUT} = 0 \text{ mA}$).

GROUND CURRENT (I_{GND})

Ground current is the current which flows through the ground pin(s). It is defined as $I_{IN} - I_{OUT}$, excluding control current.

RIPPLE REJECTION RATIO (RR)

Ripple rejection is the ability of the regulator to attenuate the ripple content of the input voltage at the output. It is specified with 100 mVrms, 400 Hz superimposed on the input voltage, where $V_{IN} = V_{OUT} + 1.5 \text{ V}$. The output decoupling capacitor is set to 10 μF , the noise bypass capacitor is set to 0.1 μF , and the load current is set to 30 mA. Ripple rejection is the ratio of the ripple content of the output vs. the input and is expressed in dB.

STANDBY CURRENT (I_{STBY})

Standby current is the current which flows into the regulator when the output is turned off by the control function ($V_{CONT} = V_{IN}$). It is measured with $V_{IN} = 8 \text{ V}$ (9 V for the 8 V output device).

SENSOR CIRCUITS

Overcurrent Sensor

The overcurrent sensor protects the device in the event that the output is shorted to ground.

Thermal Sensor

The thermal sensor protects the device in the event that the junction temperature exceeds the safe value ($T_J = 150 \text{ }^\circ\text{C}$). This temperature rise can be caused by external heat, excessive power dissipation caused by large input to output voltage drops, or excessive output current. The regulator will shut off when the temperature exceeds the safe value. As the junction temperatures decrease, the regulator will begin to operate again. Under sustained fault conditions, the regulator output will oscillate as the device turns off then resets. Damage may occur to the device under extreme fault conditions.

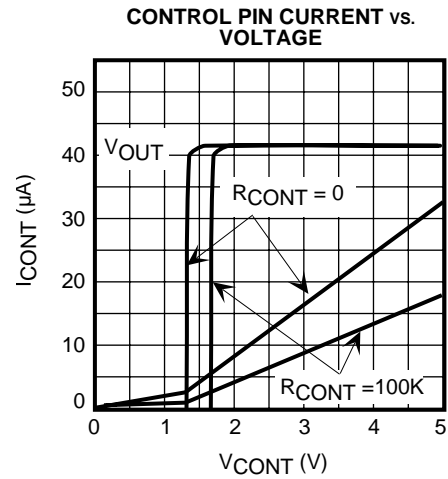
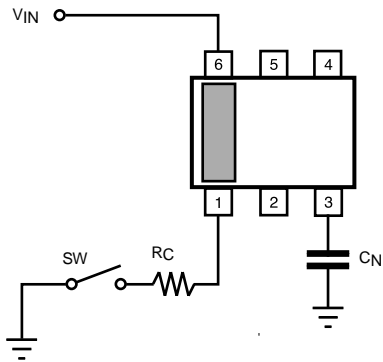
DEFINITION AND EXPLANATION OF TECHNICAL TERMS (CONT.)

Reverse Voltage Protection

Reverse voltage protection prevents damage due to the output voltage being higher than the input voltage. This fault condition can occur when the output capacitor remains charged and the input is reduced to zero, or when an external voltage higher than the input voltage is applied to the output side.

If the control function is not used, connect the control terminal to V_{IN} . When the control function is used, the control current can be reduced by inserting a series resistor (R_{CONT}) between the control terminal and V_{IN} . The value of this resistor should be determined from the graph below.

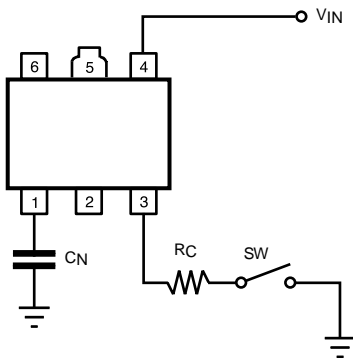
CONTROL CURRENT



Note: V_{CONT} = differential voltage from V_{IN} pin to V_{CONT} pin.

SOT23L

The requirement for the pullup resistor (R_{PULLUP}) is determined by the external control circuitry. For example, open collector/open drain logic may require R_{PULLUP} over temperature; CMOS logic will not require R_{PULLUP} .

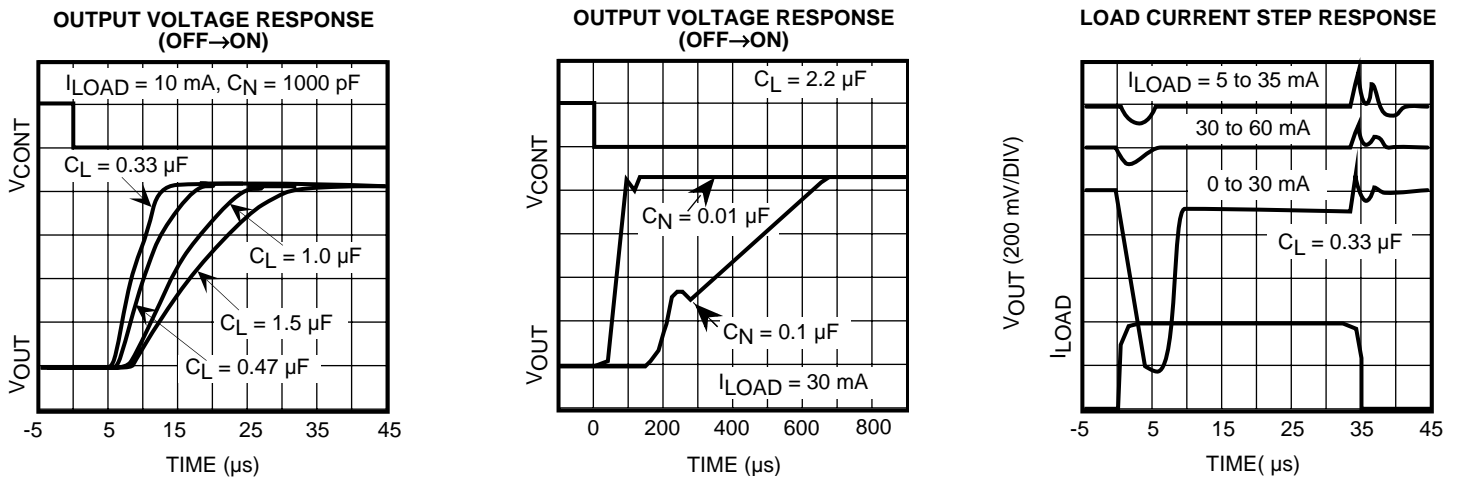


SOT89

DEFINITION AND EXPLANATION OF TECHNICAL TERMS (CONT.)

ON/OFF RESPONSE WITH CONTROL AND LOAD TRANSIENT RESPONSE

The turn-on time depends upon the value of the output capacitor and the noise bypass capacitor. The turn-on time will increase with the value of either capacitor. The graphs below show the relationship between turn-on time and load capacitance. If the value of these capacitors is reduced, the load and line regulation will suffer and the noise voltage will increase. If the value of these capacitors is increased, the turn-on time will increase.



REDUCTION OF OUTPUT NOISE

Although the architecture of the Toko regulators is designed to minimize semiconductor noise, further reduction can be achieved by the selection of external components. The obvious solution is to increase the size of the output capacitor. A more effective solution would be to add a capacitor to the noise bypass terminal. The value of this capacitor should be $0.1\text{ }\mu\text{F}$ or higher (higher values provide greater noise reduction). Although stable operation is possible without the noise bypass capacitor, this terminal has a high impedance and care should be taken to avoid a large circuit area on the printed circuit board when the capacitor is not used. Please note that several parameters are affected by the value of the capacitors and bench testing is recommended when deviating from standard values.

DEFINITION AND EXPLANATION OF TECHNICAL TERMS (CONT.)

PACKAGE POWER DISSIPATION (P_D)

This is the power dissipation level at which the thermal sensor is activated. The IC contains an internal thermal sensor which monitors the junction temperature. When the junction temperature exceeds the monitor threshold of 150 °C, the IC is shut down. The junction temperature rises as the difference between the input power ($V_{IN} \times I_{IN}$) and the output power ($V_{OUT} \times I_{OUT}$) increases. The rate of temperature rise is greatly affected by the mounting pad configuration on the PCB, the board material, and the ambient temperature. When the IC mounting has good thermal conductivity, the junction temperature will be low even if the power dissipation is great. When mounted on the recommended mounting pad, the power dissipation of the SOT-23L is increased to 600 mW. For operation at ambient temperatures over 25 °C, the power dissipation of the SOT-23L device should be derated at 4.8 mW/°C. The power dissipation of the SOT-89 package is 900 mW when mounted as recommended. Derate the power dissipation at 7.2 mW/°C for operation above 25 °C. To determine the power dissipation for shutdown when mounted, attach the device on the actual PCB and deliberately increase the output current (or raise the input voltage) until the thermal protection circuit is activated. Calculate the power dissipation of the device by subtracting the output power from the input power. These measurements should allow for the ambient temperature of the PCB. The value obtained from $P_D / (150\text{ °C} - T_A)$ is the derating factor. The PCB mounting pad should provide maximum thermal conductivity in order to maintain low device temperatures. As a general rule, the lower the temperature, the better the reliability of the device. The thermal resistance when mounted is expressed as follows:

$$T_j = \theta_{jA} \times P_D + T_A$$

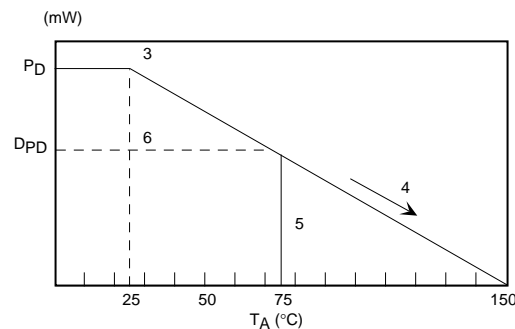
For Toko ICs, the internal limit for junction temperature is 150 °C. If the ambient temperature (T_A) is 25 °C, then:

$$\begin{aligned} 150\text{ °C} &= \theta_{jA} \times P_D + 25\text{ °C} \\ \theta_{jA} \times P_D &= 125\text{ °C} \\ \theta_{jA} &= 125\text{ °C} / P_D \end{aligned}$$

P_D is the value when the thermal sensor is activated. A simple way to determine PD is to calculate $V_{IN} \times I_{IN}$ when

the output side is shorted. Input current gradually falls as temperature rises. You should use the value when thermal equilibrium is reached.

The range of usable currents can also be found from the graph below.



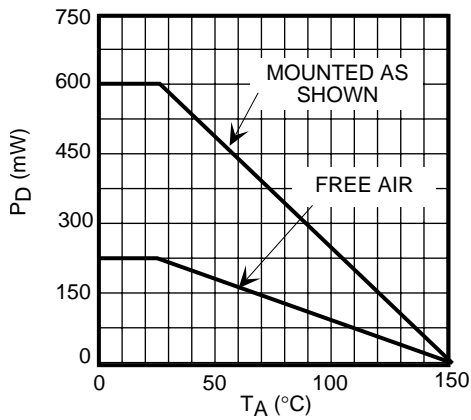
Procedure:

- 1) Find P_D
- 2) P_{D1} is taken to be $P_D \times (\sim 0.8 - 0.9)$
- 3) Plot P_{D1} against 25 °C
- 4) Connect P_{D1} to the point corresponding to the 150 °C with a straight line.
- 5) In design, take a vertical line from the maximum operating temperature (e.g., 75 °C) to the derating curve.
- 6) Read off the value of P_D against the point at which the vertical line intersects the derating curve. This is taken as the maximum power dissipation, D_{PD} .

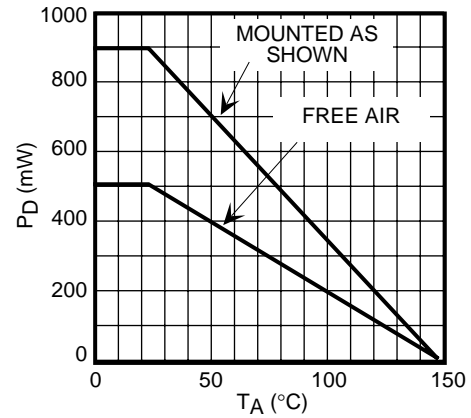
The maximum operating current is:

$$I_{OUT} = (D_{PD} / (V_{IN(MAX)} - V_{OUT}))$$

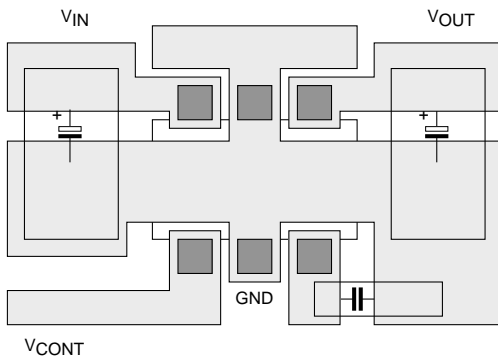
DEFINITION AND EXPLANATION OF TECHNICAL TERMS (CONT.)



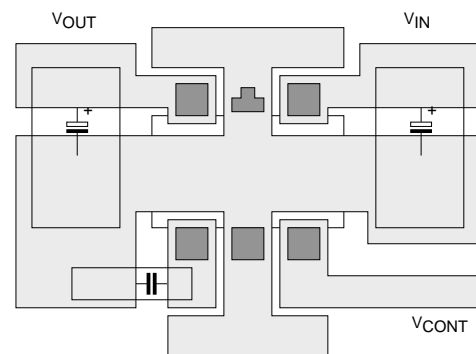
SOT23L POWER DISSIPATION CURVE



SOT89 POWER DISSIPATION CURVE



SOT23L BOARD LAYOUT



SOT89 BOARD LAYOUT

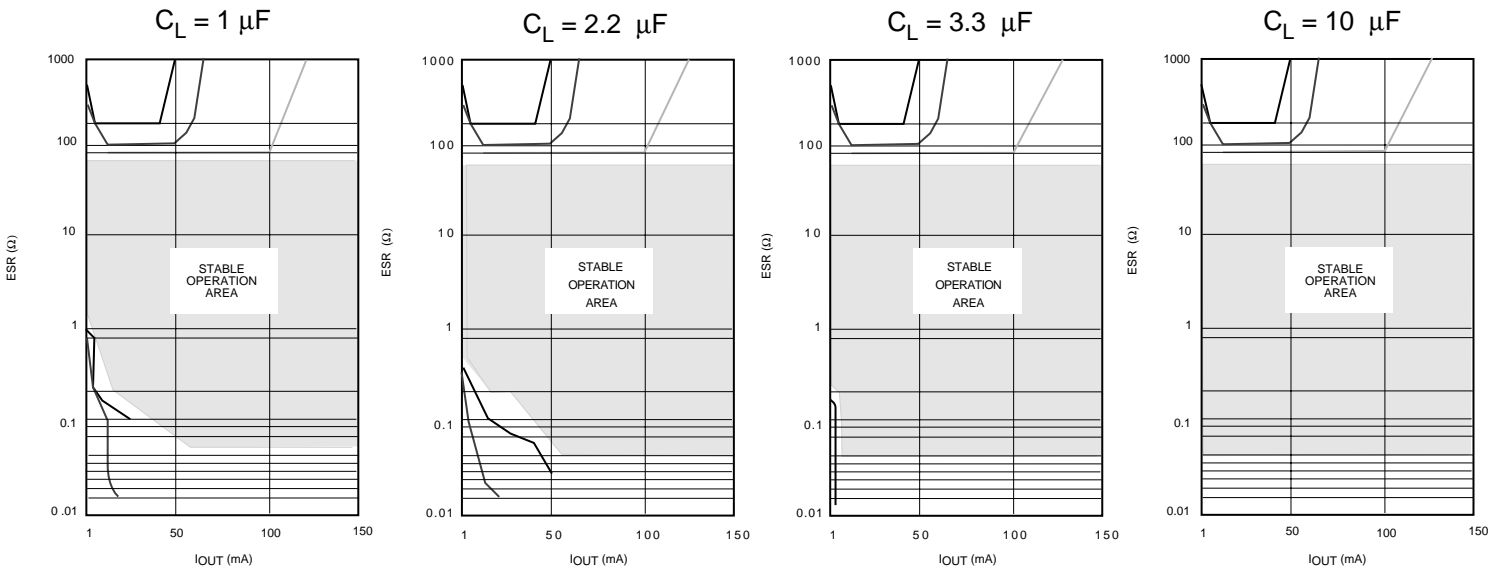
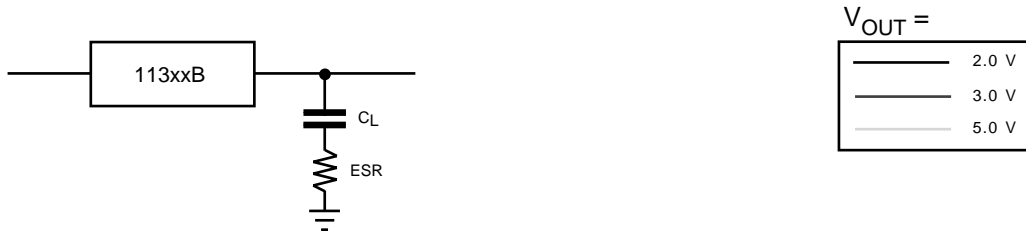
APPLICATION NOTE

Copper pattern should be as large as possible. Power dissipation is 600 mW for SOT23L and 900 mW for SOT89. A low Equivalent Series Resistance (ESR) capacitor is recommended. For low temperature operation, select a capacitor with a low ESR at the lowest operating temperature to prevent oscillation, degradation of ripple rejection and increase in noise. The minimum recommended capacitance is 2.2 μF .

APPLICATION INFORMATION

INPUT-OUTPUT CAPACITORS

Linear regulators require an output capacitor in order to maintain regulator loop stability. This capacitor should be selected to ensure stable operation over the desired temperature and load range. The graphs below show the effects of capacitance value and ESR on the stable operation area.



In general, the capacitor should be at least $1 \mu\text{F}$ (aluminum electrolytic) and be rated for the actual ambient operating temperature range. The table below shows typical characteristics for several types and values of capacitance. Please note that the ESR varies widely depending upon manufacturer, type, size, and material.

ESR Capacitance	Aluminum Capacitor	Tantalum Capacitor	Ceramic Capacitor
1.0 μF	2.4 Ω	2.3 Ω	0.140 Ω
2.2 μF	2.0 Ω	1.9 Ω	0.059 Ω
3.3 μF	4.6 Ω	1.0 Ω	0.049 Ω
10 μF	1.4 Ω	0.5 Ω	0.025 Ω

Note: ESR is measured at 10 kHz.

NOTES

PACKAGE OUTLINE

SOT23L-6

Note: Pin 2 and Pin 5 should be grounded for heat dissipation

Dimensions are shown in millimeters
Tolerance: x.x = ± 0.2 mm (unless otherwise specified)

Marking Information

Product Code	Q	Voltage Code
TK11320B		20
TK11321B		21
TK11322B		22
TK11323B		23
TK11324B		24
TK11325B		25
TK11326B		26
TK11327B		27
TK11328B		28
TK11329B		29
TK11330B		30
TK11331B		31
TK11332B		32
TK11333B		33
TK11334B		34
TK11335B		35
TK11336B		36
TK11337B		37
TK11338B		38
TK11339B		39
TK11340B		40
TK11341B		41
TK11342B		42
TK11343B		43
TK11344B		44
TK11345B		45
TK11346B		46
TK11347B		47
TK11348B		48
TK11349B		49
TK11350B		50
TK11355B		55
TK11360B		60
TK11380B		80

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Dimensions are shown in millimeters
Tolerance: x.x = ± 0.2 mm (unless otherwise specified)



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