Isolated Wide Input Range 15-Watt DC/DC Converters

Featuring a full 15 Watt or greater output in one square inch of board area, the UEI series isolated DC/DC converter family offers efficient regulated DC power for printed

muRata **Murata Power Solutions**



FEATURES

- Small footprint DC/DC converter, ideal for high current applications
- Industry standard 0.96" x 1.1" X 0.32" open frame package and pinout
- Wide range input voltages 9-36 and 18-75 Vdc
- Assembly and attachment for RoHS standards
- Isolation up to 2250 VDC (basic)
- Up to 15 Watts or greater total output power with overtemperature shutdown
- High efficiency synchronous rectifier forward topology
- Stable operation with no required external components
- Usable -40 to 85°C temperature range (with derating)
- Certified to UL 60950-1, CAN/CSA-C22.2 No. 60950-1, IEC60950-1, EN60950-1 safety approvals. 2nd edition
- Extensive self-protection shut down features

circuit board mounting.

PRODUCT OVERVIEW

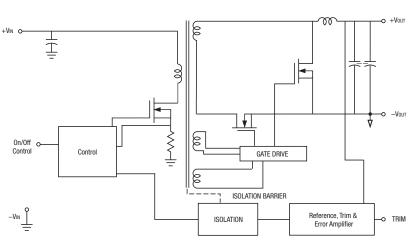
Wide range 4:1 inputs on the 0.96" x 1.1" x 0.32" converter are either 9 to 36 Volts DC (Q12 models) or 18 to 75 Volts DC (Q48 models), ideal for batterypowered and telecom equipment. The industrystandard pinout fits larger 1" x 2" converters. Fixed output voltages from 3.3 VDC to 15 VDC are regulated to within ±0.2% or less and may be trimmed within ±10% of nominal output. Applications include small instruments, area-limited microcontrollers, computer-based systems, data communications equipment, remote sensor systems, vehicle and portable electronics.

The UEI 15W series includes full magnetic and optical isolation up to 2250 Volts DC (basic insulation). For connection to digital systems, the outputs offer fast settling to current step loads and tolerance of higher capacitive loads. Excellent ripple

and noise specifications assure compatibility to circuits using CPU's, ASIC's, programmable logic and FPGA's. For systems requiring controlled startup/shutdown, an external switch, transistor or digital logic may be used to activate the remote On/Off control.

A wealth of self-protection features avoid both converter and external circuit problems. These include input undervoltage lockout and overtemperature shutdown. The outputs current limit using the "hiccup" autorestart technique and the outputs may be short-circuited indefinitely. Additional features include output overvoltage and reverse conduction elimination.

The high efficiency offers minimal heat buildup and "no fan" operation.



Typical topology is shown.

Figure 1. Simplified block diagram-3.3V and 5Vout models only.

For full details go to w.murata-ps.com/roh



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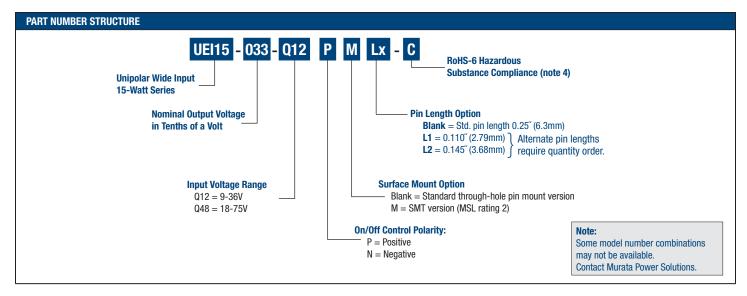
UEI15 Series

PERFORMANCE SPECIFICATIONS AND ORDERING GUIDE ${ m I}$																
	Output								Input					Open Frame		
				R/N (m)	/p-p) ②	Regulatio	on (Max.)			lin =	I іN =	Effici	Efficiency Package – C75			
Root Models ①	Vоит (V)	Іоит (А) З	Power (W)	Тур.	Max.	Line	Load	VIN Nom. (V)	Range (V)	no Ioad (mA)	full Ioad (A)	Min.	Тур.	(inches)	(mm)	Case Pinout
UEI15-033-Q12	3.3	4.5	14.85	60	90	±0.05%	±0.1%	24	9-36	43	0.71	86.5%	89%			
UEI15-033-Q48	3.3	5	16.5	60	90	±0.2%	±0.2%	48	18-75	24	0.4	85.5%	88%			DOF
UEI15-050-Q12	5	3	15	70	125	±0.05%	±0.075%	24	9-36	41	0.72	86.3%	87.3%			
UEI15-050-Q48	5	3	15	60	95	±0.05%	±0.06%	48	18-75	28	0.36	84.3%	86.0%	1 1 0 00 0 00		
UEI15-120-Q12	12	1.3	15.6	110	150	±0.05%	±0.05%	24	9-36	15	0.77	82.3%	84.5%	1.1×0.96×0.32	27.9×24.4×8.1	P85
UEI15-120-Q48	12	1.3	15.6	85	120	±0.0 75%	±0.05%	48	18-75	15	0.38	83.3%	85%			
UEI15-150-Q12	15	1.1	16.5	130	175	±0.05%	±0.05%	24	9-36	18	0.81	83.5%	85%			
UEI15-150-Q48	15	1.1	16.5	80	120	±0.05%	±0.05%	48	18-75	14	0.4	83.3%	85.3%			

 $\odot\,$ Please refer to the part number structure for additional options and complete ordering part numbers.

② All specifications are typical at nominal line voltage and full load, +25 deg.C. unless otherwise noted. See detailed specifications. ③ Minimum output load for all models is 10% of maximum current.

④ RoHS-6 compliance does not claim EU RoHS exemption 7b (lead in solder).



SPECIAL MODEL NUMBERS:

UEI-31204-C, Restricted AVL and surface mount



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INPUT CHARACT	TERISTICS													
	Start-up	Under- voltage	Reflected (back)	Input Current					Recom- mended	Internal	Reverse	Remote On/Off Control		
	threshold	Shut- down	Ripple Current	Inrush Transient	Load	Output Short Circuit	Low Line	Standby Mode	Fast-blow Fuse	Input Filter	Polarity Protec-	On/Off Current	Positive Logic blank model	Negative Logic "N" model
Model Family	V	V	mA pk-pk	A ² sec	mA	mA	A	mA	A	Туре	tion	mA	suffix	suffix
UEI15-033-Q12	9.1 [®]	8.6			100		1.86		4					
UEI15-033-Q48	16.7	15.6			65		1.03		2	С				OFF=open pin
UEI15-050-Q12	9.5	8.5			105		1.90		4		None. Install external		OFF=Gnd pin or -0.7 to +0.7V max. ON=open	or +10 to +
UEI15-050-Q48	16.7	15.6	30	0.05	60	- 50	0.96	1	2	L		1		
UEI15-120-Q12	9.5	8.4		0.00	110	50	2.04	4		fuse. See	'	pin or +10 to	ON=Gnd pin	
UEI15-120-Q48	16.7	15.6			56		1.02		1.5	C	note 15.		+15V max.	or -0.7 to
UEI15-150-Q12	9.5	8.4			130]	2.13		4	U				+0.8V max.
UEI15-150-Q48	16.7	16.2			60		1.06		2					

OUTPUT CHARACTER	RISTICS										
Model Family	lout [®] Max. mA	Vout Accuracy 50% Load % of Vnom	Adjustment Range % of Vnom	Temperature Coefficient % of Vouт /°C	Capacitive Loading Max. Low ESR <0.02Ω Max, μF	Overvoltage protection V	0V protection method	Voltage Out- put Range	Ripple/Noise (20 MHz bandwidth) ⁸	Line/Load Regulation	Efficiency
UEI15-033-Q12						4.5					
UEI15-033-Q48	See				1,000	3.9	1				
UEI15-050-Q12	Ordering			0.00		5.9	1				
UEI15-050-Q48	Guide	±1	±10 ¹⁷⁷			5.9	Magnetic feedback				
UEI15-120-Q12	(Minimum	±1	±10	±0.02		14.1		See ordering guide			
UEI15-120-Q48	load is 10%				470	15					
UEI15-150-Q12	of Imax)					19.5]				
UEI15-150-Q48]					20					

ISOLATION CHARACT		DYNAMIC CHARACTERISTICS								
	lunut to	Isolation	le el ettere	la a la tian		Dynamic Load Response	Peak	Start-	up Time	Quitabian
Model Family	Input to Output Min. Vdc	Resistance Min. MΩ	Isolation Capacitance pF	lsolation Safety Rating	Model Family	(50-75-50% load step) μSec to 1% Vour ①	Deviation mV		Remote On/Off to Vour regulated max.) mSec	
UEI15-033-Q12	2000	10	1000		UEI15-033-Q12	150	±125			350
UEI15-033-Q48	2250	10	1500		UEI15-033-Q48		±35			350
UEI15-050-Q12	2000	10	1500		UEI15-050-Q12	100	±60			350
UEI15-050-Q48	2250	10	1000	Basic	UEI15-050-Q48		±30	50	50	375
UEI15-120-Q12	2000	10	1000	insulation	UEI15-120-Q12	150	±150	50	50	340
UEI15-120-Q48	2250	10	1000		UEI15-120-Q48		±200			350
UEI15-150-Q12	2000	10	1500		UEI15-150-Q12	100	±125	1		340
UEI15-150-Q48	2250	10	1000		UEI15-150-Q48		±175			380

MISCELLANEOUS CHARACTERISTICS

Model Family	Current Limit Inception 98% of Vout, after warmup A	Short Circuit Protection Method	Short Circuit Current A	Short Circuit Duration (output shorted to ground)①	Pre- biased setup	Calculated MTBF Hours ④	Operating Temperature Range See Derating Curves	Storage Temperature Range ℃	Thermal protection/ shutdown °C
UEI15-033-Q12	6.0				2 x 10 ⁶				
UEI15-033-Q48	7.2					3.49 x 10 ⁶			
UEI15-050-Q12	4.6	Current							115
UEI15-050-Q48	4.5	limiting,	0.0	Cantinuaua	Monotonio	2 x 10 ⁶	40 to . 0500	-55 to +125 ℃	
UEI15-120-Q12	2.0	hiccup	0.3	Continuous	Monotonic		-40 to +85°C	-55 10 +125 %	
UEI15-120-Q48	1.8	autorestart				4.1 x 10 ⁶	-		135
UEI15-150-Q12	1.6					2.1 x 10 ⁶			445
UEI15-150-Q48	1.7	1				2 x 10 ⁶			115

 ① Remove overload for recovery.

 ② DATEL

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ABSOLUTE M	ABSOLUTE MAXIMUM RATINGS							
	Q12 models	Volts Max. continuous	36 VDC					
Input		Volts, transient 100mS	50 VDC					
Voltage	Q48 models	Volts Max. continuous	75 VDC					
	Q40 IIIUUUUIS	Volts, transient 100mS	100 VDC					
0n/0ff	control,	Volts, Min.	-0.3					
referre	d to –Vin	Volts, Max.	15					
Input Reve	rse Polarity Prot	ection	See fuse section					
Output Ove	rvoltage, Volts N	Vout nom. +20%						
Output Cur	rent, sustained s	Current-limited, see specs						
Storago To	moratura	Range, Min. °C	-55					
Storage Te	IIIperatule	Max. °C	+125					

Absolute Maximum Ratings

Absolute maximums are stress ratings. Exposure of devices to greater than any of these conditions may adversely affect long-term reliability. Proper operation under conditions other than those listed in the Performance/Functional Specifications is neither implied nor recommended.

SPECIFICATION NOTES

(1) All models are tested and specified with external capacitors listed in the table below. The external capacitors listed below are ONLY for establishing test specifications. They are required for our test fixtures and equipment. Your application may not need them. The converter is stable with no external capacitors but Murata Power Solutions strongly recommends external caps. All caps are low-ESR types. Where two or more capacitors are listed, these are connected in parallel. All caps should mount close to the DC/DC using short leads.

All specifications are typical unless noted. General conditions for Specifications are +25 deg.C, V_{IN} =nominal, V_{OUT} =nominal, full load. Adequate airflow must be supplied for extended testing under power.

(2) Input Ripple Current is tested and specified over a 5 Hz to 20 MHz bandwidth. Input filtering is CIN=33 μ F, 100V tantalum, CBUS=220 μ F, 100V electrolytic, LBUS=12 μ H.

(3) Note that Maximum Power Derating curves indicate an <u>average</u> current at nominal input voltage. At higher temperatures and/or lower airflow, the DC/DC converter will tolerate brief full current outputs if the total RMS current over time does not exceed the Derating curve. All Derating curves are presented at sea level altitude. Be aware of reduced power dissipation with increasing density altitude.

INPUT/OUTPUT EXTERNAL TEST CAPACITORS							
Model	Input Capacitor	Output Capacitor(s)					
UEI15-033-Q12	100 µF	1 μF & 10 μF					
UEI15-033-Q48	4.7 µF ceramic	1 μF & 10 μF					
UEI15-050-Q12	100 µF	1 μF & 10 μF					
UEI15-050-Q48	4.7 µF ceramic	1 μF & 10 μF					
UEI15-120-Q12	100 µF	1 μF & 10 μF					
UEI15-120-Q48	4.7 µF ceramic	1 μF & 10 μF					
UEI15-150-Q12	100 µF	1 μF & 10 μF					
UEI15-150-Q48	4.7 µF ceramic	1 μF & 10 μF					

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(4) Mean Time Before Failure is calculated using the Telcordia (Belcore) SR-332 Method 1, Case 3, ground fixed conditions, Tpcboard=+25 deg.C, full load, natural air convection.

(5) The On/Off Control is normally controlled by a switch. But it may also be driven with external logic or by applying appropriate external voltages which are referenced to Input Common. The On/Off Control Input should use either an open collector or open drain transistor.

(6) Output current limiting begins when the output voltage degrades approximately 2% from the selected setting.

(7) The outputs are not intended to sink appreciable reverse current. This may damage the outputs.

(8) Output noise may be further reduced by adding an external filter. See I/O Filtering and Noise Reduction.

(9) All models are fully operational and meet published specifications, including "cold start" at -40 °C.

(10) Regulation specifications describe the deviation as the line input voltage or output load current is varied from a nominal midpoint value to either extreme.

(11) The output overvoltage protection is automatic recovery after fault removal. The overvoltage may occur either from internal failure or from an external forcing voltage as in a shared power system.

(12) Output current limit and short circuit protection is non-latching. When the overcurrent fault is removed, the converter will immediately recover.

(13) Do not exceed maximum power specifications when adjusting the output trim.

(14) At zero output current, the output may contain low frequency components which exceed the ripple specification. The output may be operated indefinitely with no load.

(15) If reverse polarity is accidentally applied to the input, to ensure reverse input protection with full output load, always connect an external input **fuse** in series with the +V_N input. Use approximately twice the full input current rating with nominal input voltage.

(16) "Hiccup" operation repeatedly attempts to restart the converter with a brief, full-current output. If the overcurrent condition still exists, the restart current will be removed and then tried again. This short current pulse prevents overheating and damaging the converter. Once the fault is removed, the converter immediately recovers normal operation.

(18) Shown at Vin = 10V.

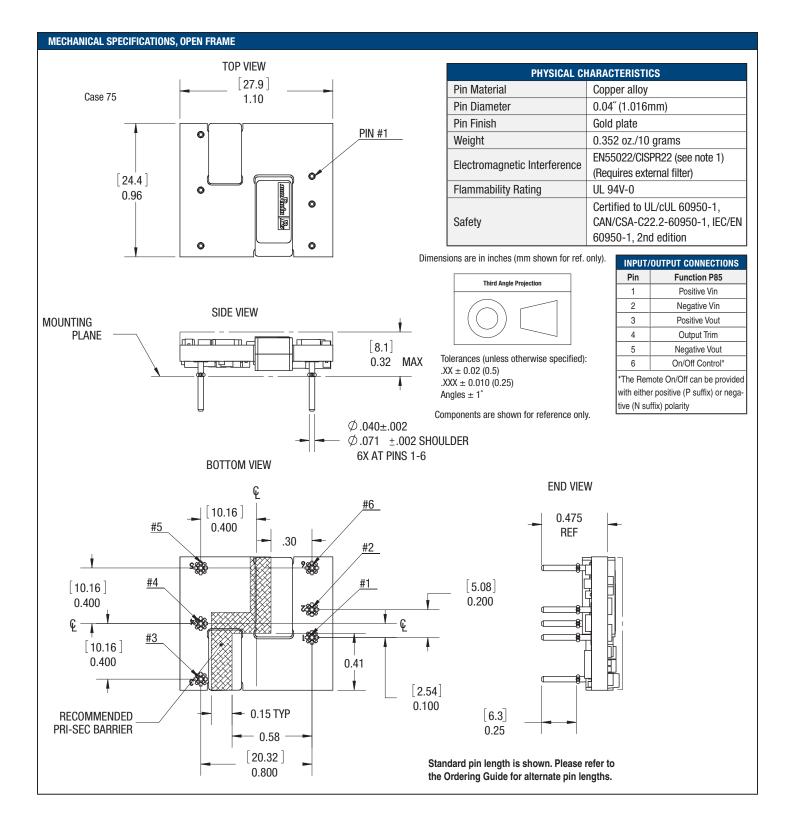
CAUTION: This product is not internally fused. To comply with safety agency certifications and to avoid injury to personnel or equipment, the user **must** connect an external fast-blow fuse to the input terminals. See fuse information.

(19) All models require 10% of Imax minimum output load to meet specifications. However, they will not be damaged at zero output load.



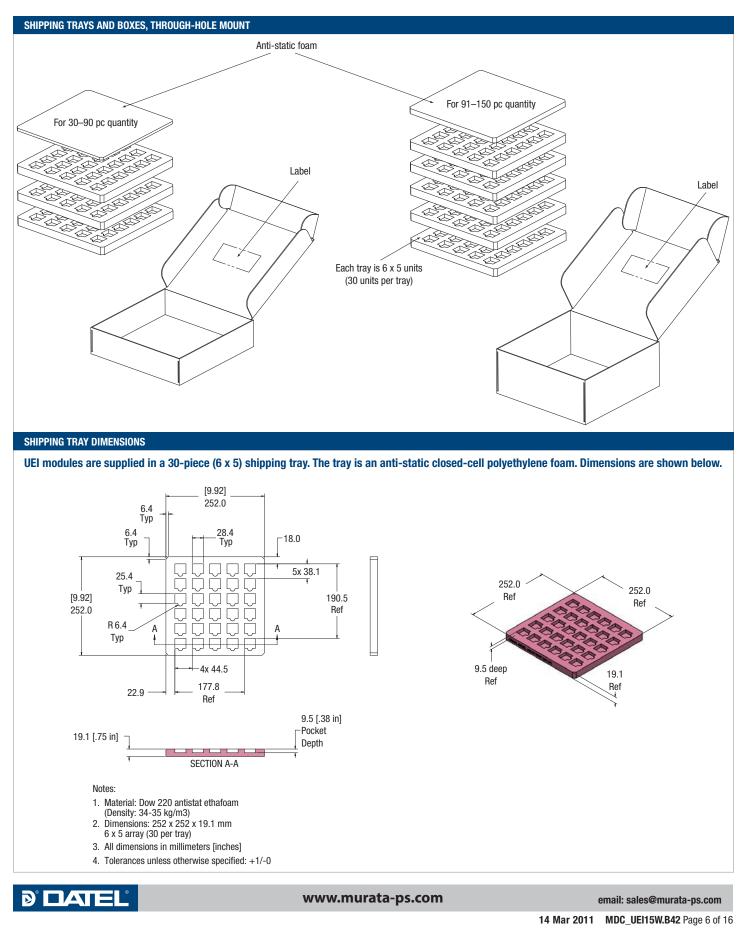
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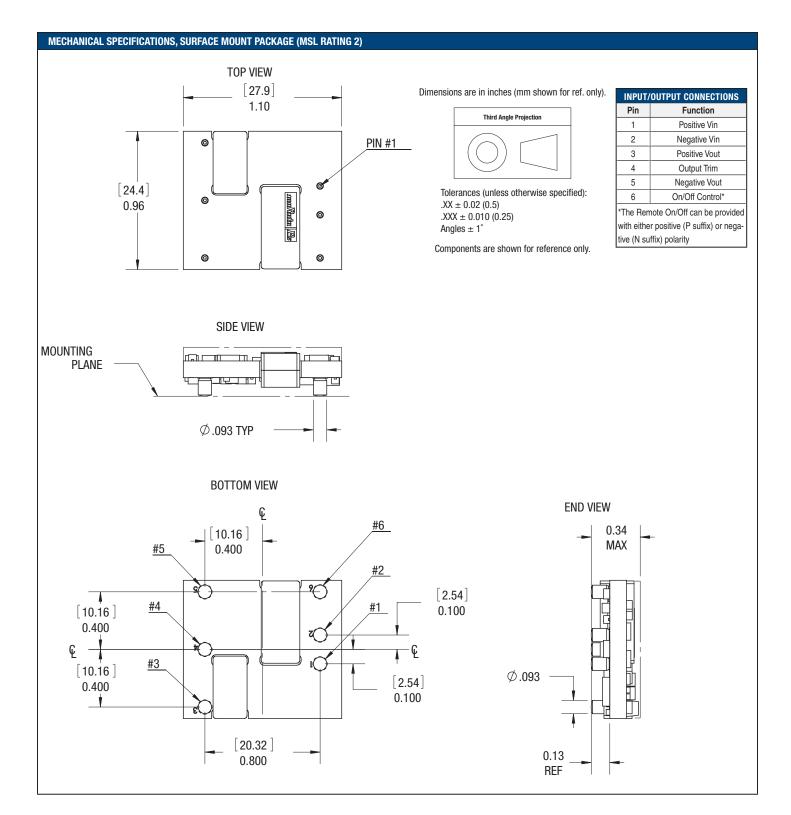


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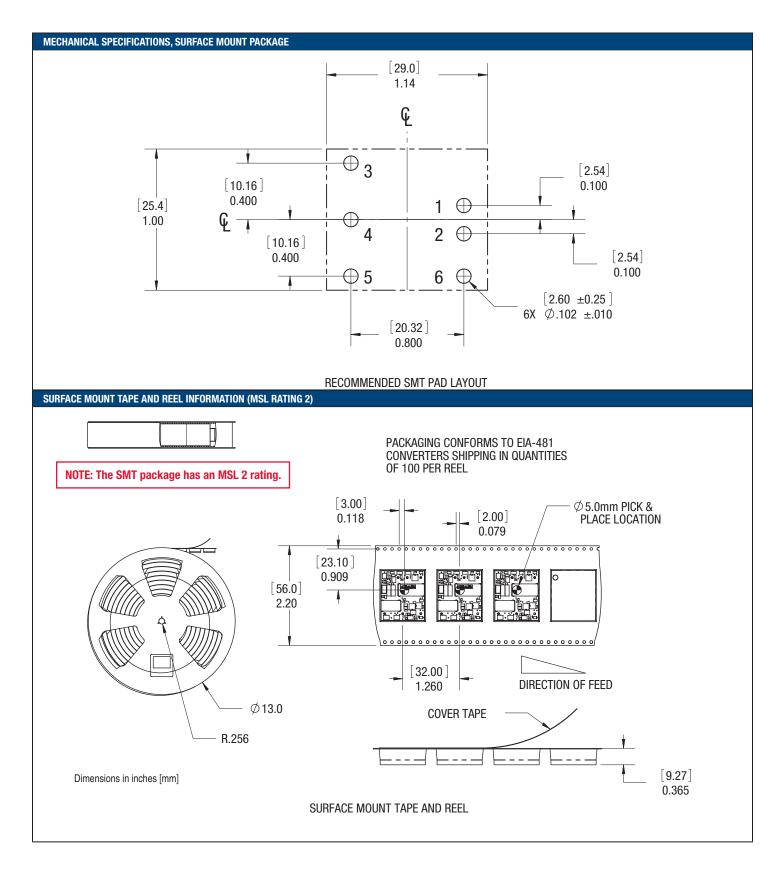




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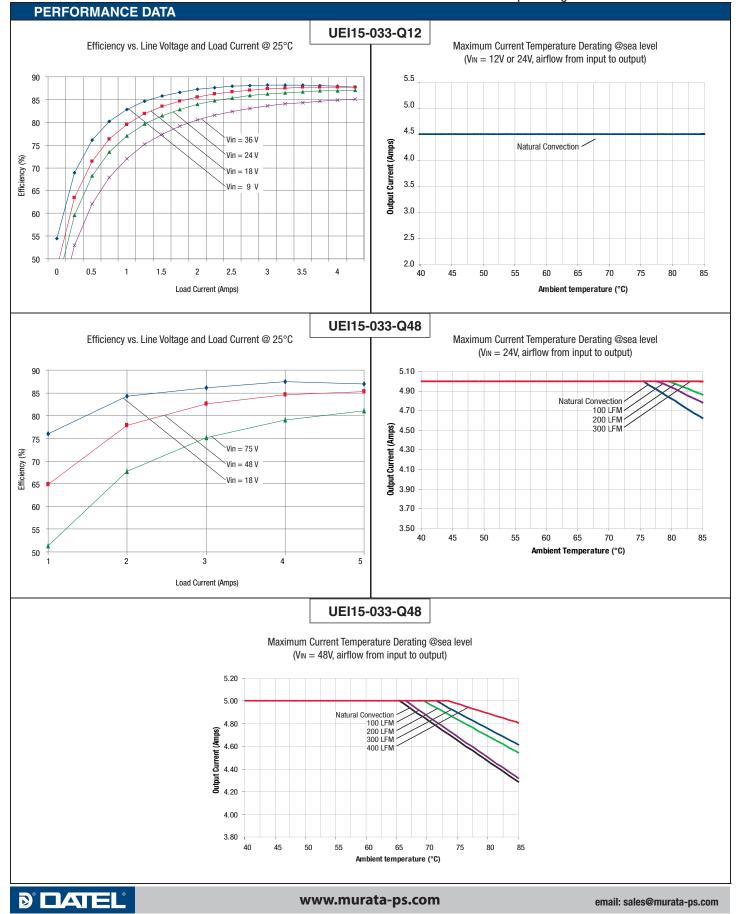


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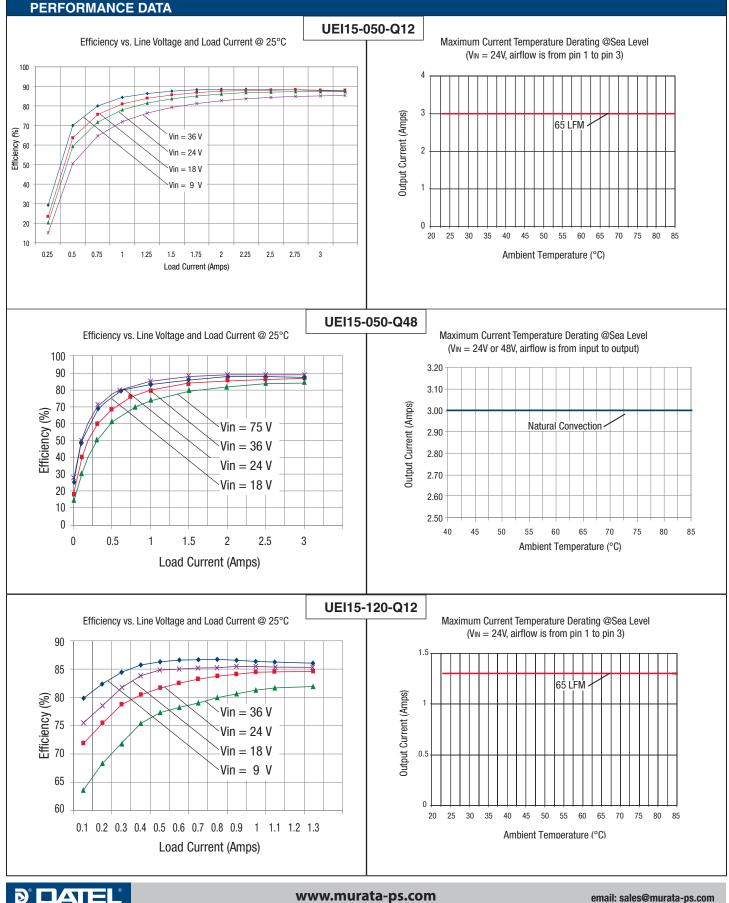
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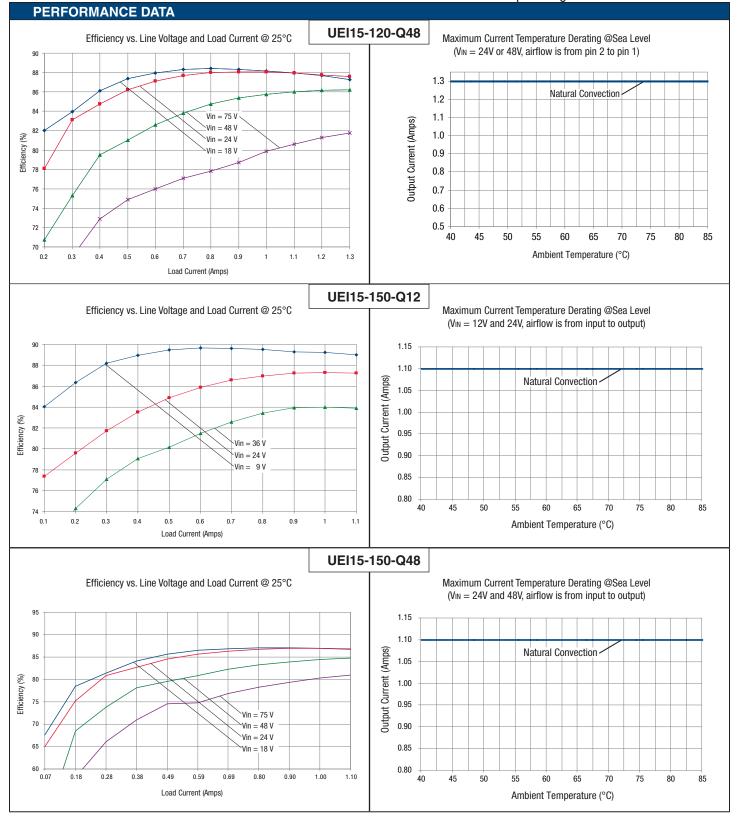
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TECHNICAL NOTES

Input Fusing

Certain applications and/or safety agencies may require fuses at the inputs of power conversion components. Fuses should also be used when there is the possibility of sustained input voltage reversal which is not current-limited. For greatest safety, we recommend a fast blow fuse installed in the ungrounded input supply line.

The installer must observe all relevant safety standards and regulations. For safety agency approvals, install the converter in compliance with the end-user safety standard, i.e. IEC/EN/UL 60950-1.

Input Reverse-Polarity Protection

If the input voltage polarity is reversed, an internal diode will become forward biased and likely draw excessive current from the power source. If this source is not current-limited or the circuit appropriately fused, it could cause permanent damage to the converter.

Input Under-Voltage Shutdown and Start-Up Threshold

Under normal start-up conditions, converters will not begin to regulate properly until the ramping-up input voltage exceeds and remains at the Start-Up Threshold Voltage (see Specifications). Once operating, converters will not turn off until the input voltage drops below the Under-Voltage Shutdown Limit. Subsequent restart will not occur until the input voltage rises again above the Start-Up Threshold. This built-in hysteresis prevents any unstable on/off operation at a single input voltage.

Users should be aware however of input sources near the Under-Voltage Shutdown whose voltage decays as input current is consumed (such as capacitor inputs), the converter shuts off and then restarts as the external capacitor recharges. Such situations could oscillate. To prevent this, make sure the operating input voltage is well above the UV Shutdown voltage AT ALL TIMES.

Start-Up Time

Assuming that the output current is set at the rated maximum, the VIN to VOUT Start-Up Time (see Specifications) is the time interval between the point when the ramping input voltage crosses the Start-Up Threshold and the fully loaded regulated output voltage enters and remains within its specified accuracy band. Actual measured times will vary with input source impedance, external input capacitance, input voltage slew rate and final value of the input voltage as it appears at the converter.

These converters include a soft start circuit to moderate the duty cycle of its PWM controller at power up, thereby limiting the input inrush current.

The On/Off Remote Control interval from On command to Vout regulated assumes that the converter already has its input voltage stabilized above the Start-Up Threshold before the On command. The interval is measured from the On command until the output enters and remains within its specified accuracy band. The specification assumes that the output is fully loaded at maximum rated current. Similar conditions apply to the On to Vout regulated specification such as external load capacitance and soft start circuitry.

Input Source Impedance

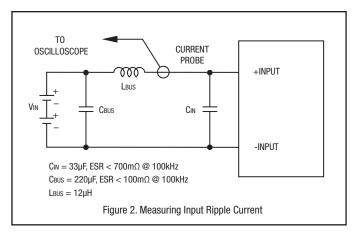
These converters will operate to specifications without external components, assuming that the source voltage has very low impedance and reasonable input voltage regulation. Since real-world voltage sources have finite impedance, performance is improved by adding external filter components. Sometimes only a small ceramic capacitor is sufficient. Since it is difficult to totally characterize all applications, some experimentation may be needed. Note that external input capacitors must accept high speed switching currents.

Because of the switching nature of DC/DC converters, the input of these converters must be driven from a source with both low AC impedance and adequate DC input regulation. Performance will degrade with increasing input inductance. Excessive input inductance may inhibit operation. The DC input regulation specifies that the input voltage, once operating, must never degrade below the Shut-Down Threshold under all load conditions. Be sure to use adequate trace sizes and mount components close to the converter.

I/O Filtering, Input Ripple Current and Output Noise

All models in this converter series are tested and specified for input reflected ripple current and output noise using designated external input/output components, circuits and layout as shown in the figures below. External input capacitors (Cin in the figure) serve primarily as energy storage elements, minimizing line voltage variations caused by transient IR drops in the input conductors. Users should select input capacitors for bulk capacitance (at appropriate frequencies), low ESR and high RMS ripple current ratings. In the figure below, the Cbus and Lbus components simulate a typical DC voltage bus. Your specific system configuration may require additional considerations. Please note that the values of Cin, Lbus and Cbus will vary according to the specific converter model.

In critical applications, output ripple and noise (also referred to as periodic and random deviations or PARD) may be reduced by adding filter elements such as multiple external capacitors. Be sure to calculate component temperature rise from reflected AC current dissipated inside capacitor ESR.



In figure 3, the two copper strips simulate real-world printed circuit impedances between the power supply and its load. In order to minimize circuit errors and standardize tests between units, scope measurements should be made using BNC connectors or the probe ground should not exceed one half inch and soldered directly to the fixture.

Floating Outputs

Since these are isolated DC/DC converters, their outputs are "floating" with respect to their input. The essential feature of such isolation is ideal ZERO CURRENT FLOW between input and output. Real-world converters however do exhibit tiny leakage currents between input and output (see Specifications). These leakages consist of both an AC stray capacitance coupling component



COPPER STRIP COPPER STRIP C1 C1 C2 SCOPE C1 C1 C2 SCOPE C0PPER STRIP C1 = 0.1 μ F CERAMIC C2 = 10 μ F TANTALUM LOAD 2-3 INCHES (51-76mm) FROM MODULE Figure 3. Measuring Output Ripple and Noise (PARD)

and a DC leakage resistance. When using the isolation feature, do not allow the isolation voltage to exceed specifications. Otherwise the converter may be damaged. Designers will normally use the negative output (-Output) as the ground return of the load circuit. You can however use the positive output (+Output) as the ground return to effectively reverse the output polarity.

Minimum Output Loading Requirements

These converters employ a synchronous rectifier design topology. Models UEI15-033-Q12, UEI15-120-Q12, and UEI15-050-Q12 require 10% minimum load to meet specifications. Operation under less than 10% load may slightly increase regulation, ripple, and noise.

Thermal Shutdown

To prevent many over temperature problems and damage, these converters include thermal shutdown circuitry. If environmental conditions cause the temperature of the DC/DC's to rise above the Operating Temperature Range up to the shutdown temperature, an on-board electronic temperature sensor will power down the unit. When the temperature decreases below the turn-on threshold, the converter will automatically restart. There is a small amount of hysteresis to prevent rapid on/off cycling.

<u>CAUTION</u>: If you operate too close to the thermal limits, the converter may shut down suddenly without warning. Be sure to thoroughly test your application to avoid unplanned thermal shutdown.

Temperature Derating Curves

The graphs in the next section illustrate typical operation under a variety of conditions. The Derating curves show the maximum continuous ambient air temperature and decreasing maximum output current which is acceptable under increasing forced airflow measured in Linear Feet per Minute ("LFM"). Note that these are AVERAGE measurements. The converter will accept brief increases in temperature and/or current or reduced airflow as long as the average is not exceeded.

Note that the temperatures are of the ambient airflow, not the converter itself which is obviously running at higher temperature than the outside air. Also note that "natural convection" is defined as very flow rates which are not using fan-forced airflow. Depending on the application, "natural convection" is usually about 30-65 LFM but is not equal to still air (0 LFM).

MPS makes Characterization measurements in a closed cycle wind tunnel with calibrated airflow. We use both thermocouples and an infrared camera system to observe thermal performance. As a practical matter, it is quite difficult to insert an anemometer to precisely measure airflow in most applications. Sometimes it is possible to estimate the effective airflow if you thoroughly understand the enclosure geometry, entry/exit orifice areas and the fan flowrate specifications. If in doubt, contact MPS to discuss placement and measurement techniques of suggested temperature sensors.

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<u>CAUTION</u>: If you routinely or accidentally exceed these Derating guidelines, the converter may have an unplanned Over Temperature shut down. Also, these graphs are all collected at slightly above Sea Level altitude. Be sure to reduce the derating for higher density altitude.

Output Overvoltage Protection

This converter monitors its output voltage for an over-voltage condition using an on-board electronic comparator. The signal is optically coupled to the primary side PWM controller. If the output exceeds OVP limits, the sensing circuit will power down the unit, and the output voltage will decrease. After a time-out period, the PWM will automatically attempt to restart, causing the output voltage to ramp up to its rated value. It is not necessary to power down and reset the converter for the this automatic OVP-recovery restart.

If the fault condition persists and the output voltage climbs to excessive levels, the OVP circuitry will initiate another shutdown cycle. This on/off cycling is referred to as "hiccup" mode. It safely tests full current rated output voltage without damaging the converter.

Output Fusing

The converter is extensively protected against current, voltage and temperature extremes. However your output application circuit may need additional protection. In the extremely unlikely event of output circuit failure, excessive voltage could be applied to your circuit. Consider using an appropriate fuse in series with the output.

Output Current Limiting

As soon as the output current increases to approximately 125% to 150% of its maximum rated value, the DC/DC converter will enter a current-limiting mode. The output voltage will decrease proportionally with increases in output current, thereby maintaining a somewhat constant power output. This is commonly referred to as power limiting.

Current limiting inception is defined as the point at which full power falls below the rated tolerance. See the Performance/Functional Specifications. Note particularly that the output current may briefly rise above its rated value. This enhances reliability and continued operation of your application. If the output current is too high, the converter will enter the short circuit condition.

Output Short Circuit Condition

When a converter is in current-limit mode, the output voltage will drop as the output current demand increases. If the output voltage drops too low, the magnetically coupled voltage used to develop primary side voltages will also drop, thereby shutting down the PWM controller. Following a time-out period, the PWM will restart, causing the output voltage to begin ramping up to its appropriate value. If the short-circuit condition persists, another shutdown cycle will initiate. This on/off cycling is called "hiccup mode". The hiccup cycling



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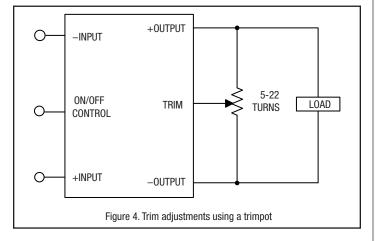
UEI15 Series

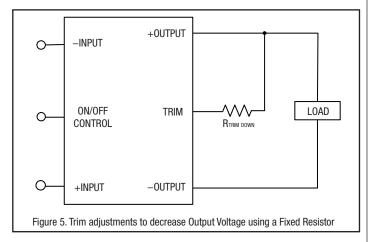
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reduces the average output current, thereby preventing excessive internal temperatures. A short circuit can be tolerated indefinitely.

Trimming the Output Voltage

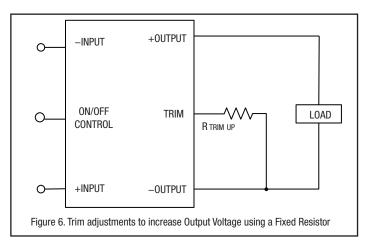
The Trim input to the converter allows the user to adjust the output voltage over the rated trim range (please refer to the Specifications). In the trim equations and circuit diagrams that follow, trim adjustments use either a trimpot or a single fixed resistor connected between the Trim input and either the positive or negative output terminals. (On some converters, an external user-supplied precision DC voltage may also be used for trimming). Trimming resistors should have a low temperature coefficient (± 100 ppm/deg.C or less) and be mounted close to the converter. Keep leads short. If the trim function is not used, leave the trim unconnected. With no trim, the converter will exhibit its specified output voltage accuracy.





There are two CAUTIONs to be aware for the Trim input:

<u>CAUTION</u>: To avoid unplanned power down cycles, do not exceed EITHER the maximum output voltage OR the maximum output power when setting the trim. Be particularly careful with a trimpot. If the output voltage is excessive, the OVP circuit may inadvertantly shut down the converter. If the maximum power is exceeded, the converter may enter current limiting. If the power is exceeded for an extended period, the converter may overheat and encounter overtem-perature shut down.



<u>CAUTION</u>: Be careful of external electrical noise. The Trim input is a senstive input to the converter's feedback control loop. Excessive electrical noise may cause instability or oscillation. Keep external connections short to the Trim input. Use shielding if needed.

Trim Equations

Trim Up <connect trim<br="">resistor between Trim and –Vout></connect>	Trim Down <connect trim<br="">resistor between Trim and +Vour></connect>
UEI15-033	3-Q12, Q48
$R_{T_{\rm UP}}(\Omega) = \frac{12775}{V_{\rm O} - 3.3} - 2050$	$R_{T_{\text{DOWN}}}(\Omega) = \frac{5110 (\text{Vo} - 2.5)}{3.3 - \text{Vo}} - 2050$
UEI15-050	0-Q12, Q48
$R_{T_{\rm UP}}(\Omega) = \frac{12775}{V_{\rm O}-5} - 2050$	$R_{T_{\text{DOWN}}}(\Omega) = \frac{5110 (\text{Vo} - 2.5)}{5 - \text{Vo}} - 2050$
UEI15-120)-Q12, Q48
$R_{T_{\rm UP}}(\Omega) = \frac{25000}{V_0 - 12} - 5110$	$R_{T_{\text{DOWN}}}(\Omega) = \frac{10000 \text{ (Vo-2.5)}}{12 - V_{\text{O}}} - 5110$
UEI15-150	0-Q12, Q48
$R_{T_{\rm UP}}(\Omega) = \frac{25000}{V_0 - 15} - 5110$	$Rr_{DOWN}(\Omega) = \frac{10000 \text{ (Vo-2.5)}}{15 - Vo} - 5110$

Where Vo = Desired output voltage. Adjustment accuracy is subject to resistor tolerances and factory-adjusted output accuracy. Mount trim resistor close to converter. Use short leads.

Remote On/Off Control

On the input side, a remote On/Off Control can be ordered with either polarity.

Positive-polarity models are enabled when the On/Off pin is left open or is pulled high to +15V max. with respect to $-V_{IN}$. Some models will also turn on at lower intermediate voltages (see Specifications). Positive-polarity devices are



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Isolated Wide Input Range 15-Watt DC/DC Converters

disabled when the On/Off is grounded or brought to within a low voltage (see Specifications) with respect to $-V_{I\!N}$.

Negative-polarity devices are on (enabled) when the On/Off is grounded or brought to within a low voltage (see Specifications) with respect to $-V_{IN}$. The device is off (disabled) when the On/Off is left open or is pulled high to $+15V_{\text{DC}}$ max. with respect to $-V_{IN}$.

Dynamic control of the On/Off function should be able to sink appropriate signal current when brought low and withstand appropriate voltage when brought high. Be aware too that there is a finite time in milliseconds (see Specifications) between the time of On/Off Control activation and stable, regulated output. This time will vary slightly with output load type and current and input conditions.

There are three CAUTIONs for the On/Off Control:

<u>CAUTION</u>: To retain full output circuit isolation, control the On/Off from the input side ONLY. If you must control it from circuits in the output, use some form of optoisolation to the On/Off Control. This latter condition is unlikely because the device controlling the On/Off would have to remain powered on and not be powered from the converter.

<u>CAUTION</u>: While it is possible to control the On/Off with external logic if you carefully observe the voltage levels, the preferred circuit is either an open drain/open collector transistor, a switch or a relay (which can thereupon be controlled by logic).

<u>CAUTION</u>: Do not apply voltages to the On/Off pin when there is no input power voltage. Otherwise the converter may be permanently damaged.

On/Off Enable Control Ground Bounce Protection

To improve reliability, if you use a small signal transistor or other external circuit to select the Remote On/Off control, make sure to return the LO side directly to the –Vin power input on the DC/DC converter. To avoid ground bounce errors, do not connect the On/Off return to a distant ground plane or current-carrying bus. If necessary, run a separate small return wire directly to the –Vin terminal. There is very little current (typically 1-5 mA) on the On/Off control however, large current changes on a return ground plane or ground bus can accidentally trigger the converter on or off. If possible, mount the On/Off transistor or other control circuit adjacent to the converter.

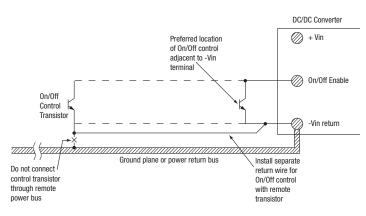
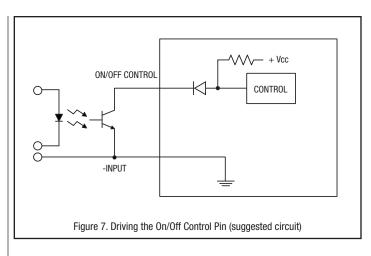


Figure 8. On/Off Enable Control Ground Bounce Protection



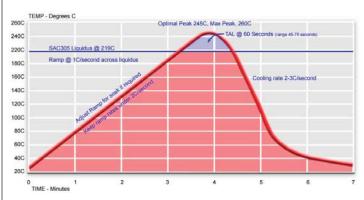
Through-hole Soldering Guidelines

Murata Power Solutions recommends the TH soldering specifications below when installing these converters. These specifications vary depending on the solder type. Exceeding these specifications may cause damage to the product. Your production environment may differ; therefore please thoroughly review these guidelines with your process engineers.

Wave Solder Operations for through-hole	mounted products (THMT)
For Sn/Ag/Cu based solders:	
Maximum Preheat Temperature	115° C.
Maximum Pot Temperature	270° C.
Maximum Solder Dwell Time	7 seconds
For Sn/Pb based solders:	
Maximum Preheat Temperature	105° C.
Maximum Pot Temperature	250° C.
Maximum Solder Dwell Time	6 seconds

SMT Reflow Soldering Guidelines

The surface-mount reflow solder profile shown below is suitable for SAC305 type leadfree solders. This graph should be used only as a *guideline*. Many other factors influence the success of SMT reflow soldering. Since your production environment may differ, please thoroughly review these guidelines with your process engineers.





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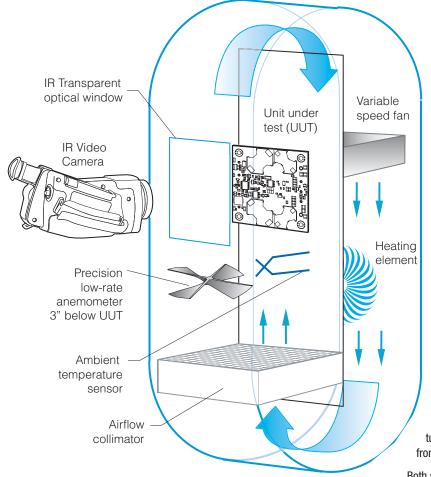


Figure 9. Vertical Wind Tunnel

Vertical Wind Tunnel

Murata Power Solutions employs a custom-designed enclosed vertical wind tunnel, infrared video camera system and test instrumentation for accurate airflow and heat dissipation analysis of power products. The system includes a precision low flow-rate anemometer, variable speed fan, power supply input and load controls, temperature gauges and adjustable heating element.

The IR camera can watch thermal characteristics of the Unit Under Test (UUT) with both dynamic loads and static steadystate conditions. A special optical port is used which is transparent to infrared wavelengths. The computer files from the IR camera can be studied for later analysis.

Both through-hole and surface mount converters are soldered down to a host carrier board for realistic heat absorption and spreading. Both longitudinal and transverse airflow studies are possible by rotation of this carrier board since there are often significant differences in the heat dissipation in the two airflow directions. The combination of both adjustable airflow, adjustable ambient heat and adjustable Input/Output currents and voltages mean that a very wide range of measurement conditions can be studied.

The airflow collimator mixes the heat from the heating element to make uniform temperature distribution. The collimator also reduces the amount of turbulence adjacent to the UUT by restoring laminar airflow. Such turbulence can change the effective heat transfer characteristics and give false readings. Excess turbulence removes more heat from some surfaces and less heat from others, possibly causing uneven overheating.

Both sides of the UUT are studied since there are different thermal gradients on each side. The adjustable heating element and fan, built-in temperature gauges and no-contact IR camera mean that power supplies are tested in real-world conditions.

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