## 402/384-OUTPUT TFT-LCD SOURCE DRIVER (64 GRAY SCALE)

## DESCRIPTION

The $\mu$ PD16647 is a source driver for TFT-LCD 64 gray scale displays. Its logic circuit operates at 3.3 V and the driver circuit operates at 5.0 V . The input data is digital data at 6 bits $\times 3$ dots, and 260,000 colors can be displayed in 64 -value outputs $\gamma$-corrected by the internal D/A converter and 10 external power supplies. The clock frequency is 50 MHz MIN. $\mu$ PD16647 can be used in TFT-LCD panels conforming to the SVGA standards.

## FEATURES

- CMOS level input
- 402/384 outputs
- 6 bits (gray scale data) x 3 dots input
-64-value output by 10 external power supplies and internal D/A converter
- Output dynamic range : Vss2 +0.1 V to V do2 -0.1 V
- High-speed data transfer: fmax $=50 \mathrm{MHz}$ MIN.(internal data transfer rate at supply voltage Vod1 of logic circuit =3.0 V )
- Level of $\gamma$-corrected power supply can be inverted
- Input data inversion function (INV)
- Precharge-less output buffer
- Logic supply voltage (VDD1) : $3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$
- Driver supply voltage (VDD2) : $5.0 \mathrm{~V} \pm 0.5 \mathrm{~V}$
- Slim TCP


## ORDERING INFORMATION

| Part Number | Package |
| :---: | :---: |
| $\mu$ PD16647N-xxx | TCP (TAB package $)$ |

Remark The TCP package is a custom-ordered item. Users are requested to consult with an NEC sales representative.

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Not all devices/types available in every country. Please check with local NEC representative for availability and additional information.

## 1. BLOCK DIAGRAM



Remark /xxx indicates active low signal.
2. PIN CONFIGURATION ( $\mu$ PD16647N-xxx)


Remark This figure does not specify the TCP package.

## 3. PIN DESCRIPTION

| Pin Symbol | Pin Name | Description |
| :---: | :---: | :---: |
| $\mathrm{S}_{1}$ to $\mathrm{S}_{402 / 384}$ | Driver output | Output 64 gray-scale analog voltages converted from digital signals. <br> Osel $=\mathrm{H}$ or open: 402 outputs ( $\mathrm{S}_{1}$ to $\mathrm{S}_{402 / 384}$ ) <br> Osel $=\mathrm{L}: 384$ outputs ( $\mathrm{S}_{1}$ to $\mathrm{S}_{192}, \mathrm{~S}_{211 / 193}$ to $\mathrm{S}_{402 / 384}$ ) <br> $\mathrm{S}_{193}$ to $\mathrm{S}_{210}$ outputs are invalid in 384 outputs. |
| Doo to D05 | Display data input | Inputs 18-bit-wide display gray scale data ( 6 bits) $\times 3$ dots (RGB). <br> Dxo : LSB, Dx5 : MSB |
| $\mathrm{D}_{10}$ to $\mathrm{D}_{15}$ |  |  |
| $\mathrm{D}_{20}$ to $\mathrm{D}_{25}$ |  |  |
| R,/L | Shift direction select input | This pin inputs/outputs start pulses in cascade mode. Shift direction of shift register is as follows: R,/L $=\mathrm{H}:$ STHR input, $\mathrm{S}_{1} \rightarrow$ S402, STHL output $R, / L=L: S T H L$ input, $S_{402} \rightarrow S_{1}$, STHR output |
| STHR | Right shift start pulse I/O | $R, / L=H$ : Inputs start pulse <br> $R, / L=L$ : Outputs start pulse |
| STHL | Left shift start pulse I/O | $R / L=H$ : Outputs start pulse <br> $R / L=L$ : Inputs start pulse |
| Bcont | Bias control | This pin can be used to finely control the bias current inside the output amplifier. In cases when fine-control is necessary, connect this pin to Vod2 using a resistor of 10 to $100 \mathrm{k} \Omega$ (per IC). When this fine-control function is not required, short-circuit this pin to VDD2. Refer to 7. Bias Current Control Function/Bcont. |
| CLK | Shift clock input | Inputs shift clock to shift register. Display data is loaded to data register at rising edge of this pin. Start pulse output goes high at rising edge of 134th clock after start pulse has been input, and serves as start pulse to driver in next stage. 134th clock of driver in first stage serves as start pulse of driver in next stage. |
| STB | Latch input | Contents of data register are latched at rising edge, transferred to D/A converter, and output as analog voltage corresponding to display data. Contents of internal shift register are cleared after STB has been input. One pulse of this signal is input when $\mu$ PD16647 is started, and then device operates normally. <br> For STB input timing, refer to 9. Switching Characteristics Waveform. |
| Osel | Selection of number of outputs | Selects number of outputs. This pin is internally pulled up to Vod1. <br> Osel $=\mathrm{H}$ or open : 402 outputs ( $\mathrm{S}_{1}$ to $\mathrm{S}_{402 / 384}$ ) <br> Osel $=\mathrm{L}: 384$ outputs ( $\mathrm{S}_{1}$ to $\mathrm{S}_{192 \text {, }} \mathrm{S}_{211 / 193}$ to $\mathrm{S}_{402 / 384}$ ) |
| V 0 to $\mathrm{V}_{9}$ | $\gamma$-corrected power supply | Inputs $\gamma$-corrected power from external source. $\begin{aligned} & V_{\mathrm{SS} 2} \leq \mathrm{V}_{9} \leq \mathrm{V}_{8} \leq \mathrm{V}_{7} \leq \mathrm{V}_{6} \leq \mathrm{V}_{5} \leq \mathrm{V}_{4} \leq \mathrm{V}_{3} \leq \mathrm{V}_{2} \leq \mathrm{V}_{1} \leq V_{0} \leq V_{\mathrm{DD} 2} \text { or } \\ & V_{\mathrm{SS} 2} \leq \mathrm{V}_{0} \leq \mathrm{V}_{1} \leq \mathrm{V}_{2} \leq \mathrm{V}_{3} \leq \mathrm{V}_{4} \leq \mathrm{V}_{5} \leq \mathrm{V}_{6} \leq \mathrm{V}_{7} \leq \mathrm{V}_{8} \leq \mathrm{V}_{9} \leq \mathrm{V}_{\mathrm{DD} 2} \end{aligned}$ <br> Maintain gray scale power supply during gray scale voltage output. |
| INV | Data inversion input | Input data can be inverted when display data is loaded. <br> INV = H : Inverts and loads input data. <br> INV = L: Does not invert input data. |
| VDD1 | Logic circuit power supply | $3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ |
| VDD2 | Driver circuit power supply | $5.0 \mathrm{~V} \pm 0.5 \mathrm{~V}$ |
| Vss1 | Logic ground | Ground |
| Vss2 | Driver ground | Ground |

Caution Be sure to turn on power in the order $V_{D D 1}$, logic input, $V_{D D 2}$, and gray scale power ( $V_{0}$ to $V_{9}$ ), and turn off power in the reverse order, to prevent the $\mu$ PD16647 from being damaged by latchup. Be sure to observe this power sequence even during a transition period.

## 4. RELATION BETWEEN INPUT DATA AND OUTPUT VOLTAGE VALUE

The 10 major points on the $\gamma$-characteristic curve of the LCD panel are arbitrarily set by external power supplies $V_{0}$ through $\mathrm{V}_{9}$. If the display data is 00 H or 3 FH , gray scale voltage $\mathrm{V}_{0}$ or $\mathrm{V}_{9}$ is output. If the display data is in the range 01 H to 3 EH , the high-order 3 bits select an external power pair $\mathrm{V}_{\mathrm{n}+1}, \mathrm{~V}_{\mathrm{n}}$. The low-order 3 bits evenly divide the range of $\mathrm{V}_{\mathrm{n}+1}$ to $\mathrm{V}_{\mathrm{n}}$ into eight segments by means of $\mathrm{D} / \mathrm{A}$ conversion (however, the ranges from $\mathrm{V}_{8}$ to $\mathrm{V}_{7}$ and from $\mathrm{V}_{1}$ to $\mathrm{V}_{0}$ are divided into seven segments) to output a 64 gray scale voltage.


Figure4-1. Relationship between Input Data and $\gamma$-corrected Voltage


Table 4-1. Relationship between Input Data and Output Voltage

| Input Data | D×5 | Dx4 | Dx3 | Dx2 | Dx1 | Dxo | Output Voltage |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 00H | 0 | 0 | 0 | 0 | 0 | 0 | $V_{0}$ |
| 01H | 0 | 0 | 0 | 0 | 0 | 1 | $\mathrm{V}_{1}+\left(\mathrm{V}_{0}-\mathrm{V}_{1}\right) \times 6 / 7$ |
| 02H | 0 | 0 | 0 | 0 | 1 | 0 | $\mathrm{V}_{1}+\left(\mathrm{V}_{0}-\mathrm{V}_{1}\right) \times 5 / 7$ |
| 03H | 0 | 0 | 0 | 0 | 1 | 1 | $\mathrm{V}_{1}+\left(\mathrm{V}_{0}-\mathrm{V}_{1}\right) \times 4 / 7$ |
| 04H | 0 | 0 | 0 | 1 | 0 | 0 | $\mathrm{V}_{1}+\left(\mathrm{V}_{0}-\mathrm{V}_{1}\right) \times 3 / 7$ |
| 05H | 0 | 0 | 0 | 1 | 0 | 1 | $V_{1}+\left(V_{0}-V_{1}\right) \times 2 / 7$ |
| 06H | 0 | 0 | 0 | 1 | 1 | 0 | $\mathrm{V}_{1}+\left(\mathrm{V}_{0}-\mathrm{V}_{1}\right) \times 1 / 7$ |
| 07H | 0 | 0 | 0 | 1 | 1 | 1 | $\mathrm{V}_{1}$ |
| 08H | 0 | 0 | 1 | 0 | 0 | 0 | $\mathrm{V}_{2}+\left(\mathrm{V}_{1}-\mathrm{V}_{2}\right) \times 7 / 8$ |
| 09H | 0 | 0 | 1 | 0 | 0 | 1 | $\mathrm{V}_{2}+\left(\mathrm{V}_{1}-\mathrm{V}_{2}\right) \times 6 / 8$ |
| OAH | 0 | 0 | 1 | 0 | 1 | 0 | $\mathrm{V}_{2}+\left(\mathrm{V}_{1}-\mathrm{V}_{2}\right) \times 5 / 8$ |
| OBH | 0 | 0 | 1 | 0 | 1 | 1 | $\mathrm{V}_{2}+\left(\mathrm{V}_{1}-\mathrm{V}_{2}\right) \times 4 / 8$ |
| OCH | 0 | 0 | 1 | 1 | 0 | 0 | $\mathrm{V}_{2}+\left(\mathrm{V}_{1}-\mathrm{V}_{2}\right) \times 3 / 8$ |
| ODH | 0 | 0 | 1 | 1 | 0 | 1 | $\mathrm{V}_{2}+\left(\mathrm{V}_{1}-\mathrm{V}_{2}\right) \times 2 / 8$ |
| 0EH | 0 | 0 | 1 | 1 | 1 | 0 | $\mathrm{V}_{2}+\left(\mathrm{V}_{1}-\mathrm{V}_{2}\right) \times 1 / 8$ |
| 0FH | 0 | 0 | 1 | 1 | 1 | 1 | $\mathrm{V}_{2}$ |
| 10 H | 0 | 1 | 0 | 0 | 0 | 0 | $\mathrm{V}_{3}+\left(\mathrm{V}_{2}-\mathrm{V}_{3}\right) \times 7 / 8$ |
| 11H | 0 | 1 | 0 | 0 | 0 | 1 | $V_{3}+\left(V_{2}-V_{3}\right) \times 6 / 8$ |
| 12 H | 0 | 1 | 0 | 0 | 1 | 0 | $\mathrm{V}_{3}+\left(\mathrm{V}_{2}-\mathrm{V}_{3}\right) \times 5 / 8$ |
| 13H | 0 | 1 | 0 | 0 | 1 | 1 | $\mathrm{V}_{3}+\left(\mathrm{V}_{2}-\mathrm{V}_{3}\right) \times 4 / 8$ |
| 14H | 0 | 1 | 0 | 1 | 0 | 0 | $\mathrm{V}_{3}+\left(\mathrm{V}_{2}-\mathrm{V}_{3}\right) \times 3 / 8$ |
| 15H | 0 | 1 | 0 | 1 | 0 | 1 | $V_{3}+\left(V_{2}-V_{3}\right) \times 2 / 8$ |
| 16H | 0 | 1 | 0 | 1 | 1 | 0 | $\mathrm{V}_{3}+\left(\mathrm{V}_{2}-\mathrm{V}_{3}\right) \times 1 / 8$ |
| 17H | 0 | 1 | 0 | 1 | 1 | 1 | $\mathrm{V}_{3}$ |
| 18 H | 0 | 1 | 1 | 0 | 0 | 0 | $\mathrm{V}_{4}+\left(\mathrm{V}_{3}-\mathrm{V}_{4}\right) \times 7 / 8$ |
| 19H | 0 | 1 | 1 | 0 | 0 | 1 | $\mathrm{V}_{4}+\left(\mathrm{V}_{3}-\mathrm{V}_{4}\right) \times 6 / 8$ |
| 1AH | 0 | 1 | 1 | 0 | 1 | 0 | $\mathrm{V}_{4}+\left(\mathrm{V}_{3}-\mathrm{V}_{4}\right) \times 5 / 8$ |
| 1BH | 0 | 1 | 1 | 0 | 1 | 1 | $\mathrm{V}_{4}+\left(\mathrm{V}_{3}-\mathrm{V}_{4}\right) \times 4 / 8$ |
| 1 CH | 0 | 1 | 1 | 1 | 0 | 0 | $\mathrm{V}_{4}+\left(\mathrm{V}_{3}-\mathrm{V}_{4}\right) \times 3 / 8$ |
| 1DH | 0 | 1 | 1 | 1 | 0 | 1 | $\mathrm{V}_{4}+\left(\mathrm{V}_{3}-\mathrm{V}_{4}\right) \times 2 / 8$ |
| 1EH | 0 | 1 | 1 | 1 | 1 | 0 | $\mathrm{V}_{4}+\left(\mathrm{V}_{3}-\mathrm{V}_{4}\right) \times 1 / 8$ |
| 1FH | 0 | 1 | 1 | 1 | 1 | 1 | $\mathrm{V}_{4}$ |
| 20 H | 1 | 0 | 0 | 0 | 0 | 0 | $\mathrm{V}_{5}+\left(\mathrm{V}_{4}-\mathrm{V}_{5}\right) \times 7 / 8$ |
| 21H | 1 | 0 | 0 | 0 | 0 | 1 | $V_{5}+\left(V_{4}-V_{5}\right) \times 6 / 8$ |
| 22 H | 1 | 0 | 0 | 0 | 1 | 0 | $\mathrm{V}_{5}+\left(\mathrm{V}_{4}-\mathrm{V}_{5}\right) \times 5 / 8$ |
| 23H | 1 | 0 | 0 | 0 | 1 | 1 | $\mathrm{V}_{5}+\left(\mathrm{V}_{4}-\mathrm{V}_{5}\right) \times 4 / 8$ |
| 24H | 1 | 0 | 0 | 1 | 0 | 0 | $\mathrm{V}_{5}+\left(\mathrm{V}_{4}-\mathrm{V}_{5}\right) \times 3 / 8$ |
| 25H | 1 | 0 | 0 | 1 | 0 | 1 | $V_{5}+\left(V_{4}-V_{5}\right) \times 2 / 8$ |
| 26H | 1 | 0 | 0 | 1 | 1 | 0 | $\mathrm{V}_{5}+\left(\mathrm{V}_{4}-\mathrm{V}_{5}\right) \times 1 / 8$ |
| 27H | 1 | 0 | 0 | 1 | 1 | 1 | $\mathrm{V}_{5}$ |
| 28H | 1 | 0 | 1 | 0 | 0 | 0 | $\mathrm{V}_{6}+\left(\mathrm{V}_{5}-\mathrm{V}_{6}\right) \times 7 / 8$ |
| 29H | 1 | 0 | 1 | 0 | 0 | 1 | $\mathrm{V}_{6}+\left(\mathrm{V}_{5}-\mathrm{V}_{6}\right) \times 6 / 8$ |
| 2AH | 1 | 0 | 1 | 0 | 1 | 0 | $\mathrm{V}_{6}+\left(\mathrm{V}_{5}-\mathrm{V}_{6}\right) \times 5 / 8$ |
| 2BH | 1 | 0 | 1 | 0 | 1 | 1 | $\mathrm{V}_{6}+\left(\mathrm{V}_{5}-\mathrm{V}_{6}\right) \times 4 / 8$ |
| 2CH | 1 | 0 | 1 | 1 | 0 | 0 | $\mathrm{V}_{6}+\left(\mathrm{V}_{5}-\mathrm{V}_{6}\right) \times 3 / 8$ |
| 2DH | 1 | 0 | 1 | 1 | 0 | 1 | $\mathrm{V}_{6}+\left(\mathrm{V}_{5}-\mathrm{V}_{6}\right) \times 2 / 8$ |
| 2EH | 1 | 0 | 1 | 1 | 1 | 0 | $\mathrm{V}_{6}+\left(\mathrm{V}_{5}-\mathrm{V}_{6}\right) \times 1 / 8$ |
| 2FH | 1 | 0 | 1 | 1 | 1 | 1 | $\mathrm{V}_{6}$ |
| 30 H | 1 | 1 | 0 | 0 | 0 | 0 | $\mathrm{V}_{7}+\left(\mathrm{V}_{6}-\mathrm{V}_{7}\right) \times 7 / 8$ |
| 31 H | 1 | 1 | 0 | 0 | 0 | 1 | $\mathrm{V}_{7}+\left(\mathrm{V}_{6}-\mathrm{V}_{7}\right) \times 6 / 8$ |
| 32 H | 1 | 1 | 0 | 0 | 1 | 0 | $\mathrm{V}_{7}+\left(\mathrm{V}_{6}-\mathrm{V}_{7}\right) \times 5 / 8$ |
| 33H | 1 | 1 | 0 | 0 | 1 | 1 | $\mathrm{V}_{7}+\left(\mathrm{V}_{6}-\mathrm{V}_{7}\right) \times 4 / 8$ |
| 34H | 1 | 1 | 0 | 1 | 0 | 0 | $\mathrm{V}_{7}+\left(\mathrm{V}_{6}-\mathrm{V}_{7}\right) \times 3 / 8$ |
| 35H | 1 | 1 | 0 | 1 | 0 | 1 | $\mathrm{V}_{7}+\left(\mathrm{V}_{6}-\mathrm{V}_{7}\right) \times 2 / 8$ |
| 36H | 1 | 1 | 0 | 1 | 1 | 0 | $\mathrm{V}_{7}+\left(\mathrm{V}_{6}-\mathrm{V}_{7}\right) \times 1 / 8$ |
| 37 H | 1 | 1 | 0 | 1 | 1 | 1 | $\mathrm{V}_{7}$ |
| 38 H | 1 | 1 | 1 | 0 | 0 | 0 | $\mathrm{V}_{8}+\left(\mathrm{V}_{7}-\mathrm{V}_{8}\right) \times 6 / 7$ |
| 39H | 1 | 1 | 1 | 0 | 0 | 1 | $\mathrm{V}_{8}+\left(\mathrm{V}_{7}-\mathrm{V}_{8}\right) \times 5 / 7$ |
| 3AH | 1 | 1 | 1 | 0 | 1 | 0 | $\mathrm{V}_{8}+\left(\mathrm{V}_{7}-\mathrm{V}_{8}\right) \times 4 / 7$ |
| 3BH | 1 | 1 | 1 | 0 | 1 | 1 | $\mathrm{V}_{8}+\left(\mathrm{V}_{7}-\mathrm{V}_{8}\right) \times 3 / 7$ |
| 3CH | 1 | 1 | 1 | 1 | 0 | 0 | $\mathrm{V}_{8}+\left(\mathrm{V}_{7}-\mathrm{V}_{8}\right) \times 2 / 7$ |
| 3DH | 1 | 1 | 1 | 1 | 0 | 1 | $\mathrm{V}_{8}+\left(\mathrm{V}_{7}-\mathrm{V}_{8}\right) \times 1 / 7$ |
| 3EH | 1 | 1 | 1 | 1 | 1 | 0 | $\mathrm{V}_{8}$ |
| 3FH | 1 | 1 | 1 | 1 | 1 | 1 | V9 |

## $4.1 \gamma$-Corrected Power Circuit

The reference power supply of the D/A converter consists of a ladder circuit with a total of 64 resistors, and resistance Eri between $\gamma$-corrected power pins differs depending on each pair of $\gamma$-corrected power pins. One pair of $\gamma$-corrected power pins consists of seven or eight series resistors, and resistance $\Sigma r i$ in the figure below is indicated as the sum of the seven or eight resistors. The resistance ratio between the $\gamma$-corrected power pins ( $\Sigma$ ri ratio) is designed to be a value relatively close to the ratio of the $\gamma$-corrected voltages $\mathrm{V}_{1}$ through $\mathrm{V}_{8}$ (gray scale voltages in 7 steps) used in an actual LCD panel. Under ideal conditions where there is no difference between the two, therefore, there is no voltage difference between the voltage of the $\gamma$-corrected power supplies and the gray scale voltages in 7 steps of the resistor ladder circuits of the $\mu$ PD16647, and no current flows into the $\gamma$-corrected power pins $\mathrm{V}_{1}$ through $\mathrm{V}_{8}$. As a result, a voltage follower circuit is not necessary.

Figure4-2. $\boldsymbol{\gamma}$-Corrected Power Circuit


## 5. RELATION BETWEEN INPUT DATA AND OUTPUT VOLTAGE

Data format: 6 bits $\times$ RGB ( 3 dots)
Input width : 18 bits (1 pixel data)
(1) $R, / L=H$ (right shift)

| Output | $S_{1}$ | $S_{2}$ | $S_{3}$ | $S_{4}$ | $\ldots$ | $S_{401 / 383}$ | $S_{402 / 384}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Data | $D_{00}$ to $D_{05}$ | $D_{10}$ to $D_{15}$ | $D_{20}$ to $D_{25}$ | $D_{00}$ to $D_{05}$ | $\ldots$ | $D_{10}$ to $D_{15}$ | $D_{20}$ to $D_{25}$ |

(2) $R, / L=L$ (left shift)

| Output | $\mathrm{S}_{1}$ | $\mathrm{~S}_{2}$ | $\mathrm{~S}_{3}$ | $\mathrm{~S}_{4}$ | $\ldots$ | $\mathrm{~S}_{401 / 383}$ | $\mathrm{~S}_{402 / 384}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Data | $\mathrm{D}_{00}$ to $\mathrm{D}_{05}$ | $\mathrm{D}_{10}$ to $\mathrm{D}_{15}$ | $\mathrm{D}_{20}$ to $\mathrm{D}_{25}$ | $\mathrm{D}_{00}$ to $\mathrm{D}_{05}$ | $\ldots$ | $\mathrm{D}_{10}$ to $\mathrm{D}_{15}$ | $\mathrm{D}_{20}$ to $\mathrm{D}_{25}$ |

## 6. OPERATION OF OUTPUT BUFFER

The output buffer consists of an operational amplifier circuit that does not perform precharge operation. Therefore, driver output current Ivoн $1 / 2$ is the charging current to the LCD, and IvoL $1 / 2$ is the discharging current.

Figure6-1. LCD panel driving waveform of $\mu$ PD16647


## 7. BIAS CURRENT CONTROL FUNCTION/Bcont

It is possible to fine-control the current consumption by using the bias current control function (Bcont pin). When using this function, connect this pin to the stabilized Vod2 potential using an external resistor (RExt). When not using this function, however, short-circuit this pin to Vdd2.

Figure7-1. Bias Current Control Function/Bcont


Refer to the table below for the percentage of current regulation when using the bias current control function.

Table7-1. Current Consumption Regulation Percentage Compared to Normal Mode

| RExT | Current Consumption Regulation Percentage |
| :---: | :---: |
| SHORT | $100 \%$ |
| $10 \mathrm{k} \Omega$ | $95 \%$ |
| $20 \mathrm{k} \Omega$ | $91 \%$ |
| $40 \mathrm{k} \Omega$ | $85 \%$ |
| $80 \mathrm{k} \Omega$ | $79 \%$ |

Remark Be aware that the above current consumption regulation percentages are not product-
characteristic guaranteed as they are based on the results of simulation.

Caution Because the low-power and bias-current control functions control the bias current in the output amplifier and regulate the over-all current consumption of the driver IC, when this occurs, the characteristics of the output amplifier will simultaneously change. Therefore, when using these functions, be sure to sufficiently evaluate the picture quality.

## 8. ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings ( $\mathrm{Vss}_{\mathrm{ss}}=\mathrm{V}_{\mathrm{ss} 2}=0 \mathrm{~V}$ )

| Parameter | Symbol | Ratings | Unit |
| :---: | :---: | :---: | :---: |
| Logic Supply Voltage | VDD1 | -0.3 to +4.5 | V |
| Driver Supply Voltage | V DD2 | -0.3 to +6.0 | V |
| Input Voltage | V | -0.3 to $\mathrm{V}_{\mathrm{DDL}, 2}+0.3$ | V |
| Output Voltage | Vo | -0.3 to $\mathrm{V}_{\mathrm{DD} 1,2}+0.3$ | V |
| Operating Ambient Temperature | $\mathrm{T}_{\mathrm{A}}$ | -10 to +75 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | $\mathrm{T}_{\text {stg }}$ | -55 to +125 | ${ }^{\circ} \mathrm{C}$ |

Caution If the absolute maximum rating of even one of the above parameters is exceeded even momentarily, the quality of the product may be degraded. Absolute maximum ratings, therefore, specify the values exceeding which the product may be physically damaged. Be sure to use the product within the range of the absolute maximum ratings.

Recommended Operating Range ( $\mathrm{T}_{\mathrm{A}}=\mathbf{- 1 0}$ to $+75^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{ss} 1}=\mathrm{V}_{\mathrm{ss} 2}=\mathbf{0} \mathrm{V}$ )

| Parameter | Symbol | MIN. | TYP. | MAX. | Unit |
| :--- | :--- | :---: | :---: | :---: | :---: |
| Logic Supply Voltage | $\mathrm{V}_{\mathrm{DD} 1}$ | 3.0 | 3.3 | 3.6 | V |
| Driver Supply Voltage | $\mathrm{V}_{\mathrm{DD} 2}$ | 4.5 | 5.0 | 5.5 | V |
| High-level Input Voltage | $\mathrm{V}_{\mathrm{IH}}$ | $0.7 \mathrm{~V}_{\mathrm{DD} 1}$ |  | $\mathrm{~V}_{\mathrm{DD} 1}$ | V |
| Low-level Input Voltage | $\mathrm{V}_{\mathrm{IL}}$ | 0 |  | 0.3 VDD | V |
| $\gamma$-corrected Supply Voltage | $\mathrm{V}_{0}$ to $\mathrm{V}_{9}$ | $\mathrm{~V}_{\mathrm{SS} 2}+0.1$ |  | $\mathrm{~V}_{\mathrm{DD} 2}-0.1$ | V |
| Maximum Clock Frequency | fmax. | 50 |  |  | MHz |

Electrical Characteristics ( $\mathrm{T}_{\mathrm{A}}=-10$ to $+75^{\circ} \mathrm{C}, \mathrm{VDD} 1^{\mathrm{V}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD} 2}=5.0 \mathrm{~V} \pm 0.5 \mathrm{~V}, \mathrm{Vss}^{2}=\mathrm{V}_{\mathrm{ss} 2}=0 \mathrm{~V}$ )

| Parameter | Symbol | Condition |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input Leakage Current | IIL | $\begin{aligned} & \mathrm{D}_{00}-\mathrm{D}_{05}, \mathrm{D}_{10}-\mathrm{D}_{15}, \mathrm{D}_{20}-\mathrm{D}_{25} \\ & \mathrm{R}, / \mathrm{L}, \mathrm{STB} \end{aligned}$ |  |  |  | $\pm 1.0$ | $\mu \mathrm{A}$ |
| Pull-up Resistor | Rpu | $\mathrm{V}_{\mathrm{DD} 1}=3.3 \mathrm{~V}$ |  | 40 | 100 | 250 | $k \Omega$ |
| High-level Output Voltage | Vон | STHR(STHL), $\mathrm{lo}=-1.0 \mathrm{~mA}$ |  | VDD1-0.5 |  |  | V |
| Low-level Output Voltage | Vol | STHR(STHL), $\mathrm{lo}=+1.0 \mathrm{~mA}$ |  |  |  | 0.5 | V |
| Static Current Consumption of $\gamma$-corrected Power | Ivn1 | $\begin{aligned} & V_{D D 1}=3.3 \mathrm{~V}, \\ & V_{n}-V_{n+1}=0.5 \mathrm{~V}, \\ & V_{D D 2}=5.0 \mathrm{~V} \end{aligned}$ | $\mathrm{V}_{0}-\mathrm{V}_{1}$ | 126 | 253 | 506 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V}_{1}-\mathrm{V}_{2}$ | 145 | 291 | 582 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V}_{2}-\mathrm{V}_{3}$ | 289 | 579 | 1158 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V}_{3}-\mathrm{V}_{4}$ | 252 | 504 | 1008 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V}_{4}-\mathrm{V}_{5}$ | 343 | 686 | 1372 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V}_{5}-\mathrm{V}_{6}$ | 315 | 631 | 1262 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V}_{6}-\mathrm{V}_{7}$ | 237 | 474 | 948 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V}_{7}-\mathrm{V}_{8}$ | 158 | 316 | 632 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V}_{8}-\mathrm{V}_{9}$ | 40 | 80 | 160 | $\mu \mathrm{A}$ |
| Driver Output Current | Ivoh2 | $\begin{aligned} & \text { Vout }=4.4 \mathrm{~V}, \mathrm{Vx}=4.9 \mathrm{~V} \text { Note } 1 \\ & \mathrm{~V}_{\mathrm{DD} 1}=3.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD} 2}=5.0 \mathrm{~V} \end{aligned}$ |  | (-0.12) |  | -0.03 | mA |
|  | Ivol2 | $\begin{aligned} & \text { Vout }=0.6 \mathrm{~V}, \mathrm{~V}_{\mathrm{X}}=0.1 \mathrm{~V} \text { Note1 } \\ & \mathrm{V}_{\mathrm{DD} 1}=3.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD} 2}=5.0 \mathrm{~V} \end{aligned}$ |  | 0.04 |  | (0.16) | mA |
| Output Voltage Deviation | $\Delta \mathrm{V}$ 。 | $\begin{aligned} & V_{\mathrm{DD} 1}=3.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD} 2}=5.0 \mathrm{~V}, \\ & \text { Vout }^{2} .5 \mathrm{~V} \text { Note1 } \end{aligned}$ |  |  | $\pm 10$ | $\pm 20$ | mV |
| Output Swing Difference Deviation | $\Delta \mathrm{VP}_{\text {P-P }}$ | Input data |  |  | $( \pm 5)$ |  | mV |
| Output Voltage Range | Vo | Input data : 00 H to 3 FH |  | Vss2 +0.1 |  | VDD2 - 0.1 | V |
| Dynamic Logic Current Consumption | IdD1 | No load, VdD2 $=3.3 \mathrm{~V}^{\text {Note2 }}$ |  |  | 0.5 | 2.5 | mA |
| Dynamic Driver Current Consumption | IdD2 | No load, VDD2 $=5.0 \mathrm{~V}^{\text {Note2 }}$ |  |  | 5.0 | 10.0 | mA |

Notes 1. Vx refers to the output voltage of analog output pins $S_{1}$ to $S_{402 / 384}$.
Vout refers to the voltage applied to analog output pins $\mathrm{S}_{1}$ to $\mathrm{S}_{402 / 384}$.
2. The STB cycle is specified at $31 \mu \mathrm{~s}$ and fclk= 16 MHz .

Switching Characteristics ( $\mathrm{T}_{\mathrm{A}}=-10$ to $+75^{\circ} \mathrm{C}, \mathrm{VdD}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}, \mathrm{~V} D 2=5.0 \mathrm{~V} \pm 0.5 \mathrm{~V}, \mathrm{Vss} 1=\mathrm{Vss2}=0 \mathrm{~V}$ )

| Parameter | Symbol | Condition |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Start Pulse Delay Time | tpLH1 | $C \mathrm{~L}=15 \mathrm{pF}$ |  |  | 7 | 12 | ns |
|  | tpHL1 |  |  |  | 7 | 12 | ns |
| Driver Output Delay Time | tpLH2 | $\left\{\begin{array}{l} \mathrm{VDD} 2=5.0 \mathrm{~V} \\ 5 \mathrm{k} \Omega+36 \mathrm{pF} \end{array}\right.$ | Vo: $0.1 \mathrm{~V} \rightarrow 4.9 \mathrm{~V}$ |  | 2.2 | 10 | $\mu \mathrm{s}$ |
|  | tpLH3 |  |  |  | 2.9 | 12 | $\mu \mathrm{s}$ |
|  | tPHL2 |  | V : $4.9 \mathrm{~V} \rightarrow 0.1 \mathrm{~V}$ |  | 2.6 | 10 | $\mu \mathrm{s}$ |
|  | tPHL3 |  |  |  | 3.6 | 12 | $\mu \mathrm{s}$ |
| Input Capacitance | $\mathrm{Cl}_{11}$ | STHR (STHL), $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  |  | 10 | 20 | pF |
|  | $\mathrm{Cl}_{12}$ | $\mathrm{V}_{0}$ to $\mathrm{V}_{9}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  |  | 100 | 150 | pF |
|  | $\mathrm{Cl}_{3}$ | STHR (STHL), other than $\mathrm{V}_{0}$ to $\mathrm{V}_{9}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  |  | 10 | 15 | pF |

## <Output Load>



Timing Requirements ( $\mathrm{T}_{\mathrm{A}}=-10$ to $+75^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD} 1}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD} 2}=5.0 \mathrm{~V} \pm 0.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{ss} 1}=\mathrm{V}_{\mathrm{ss} 2}=0 \mathrm{~V}$ )

| Parameter | Symbol | Condition | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Clock Pulse Width | PWClk |  | 20 |  |  | ns |
| Clock Low Period | PW ${ }_{\text {clk (L) }}$ |  | 4 |  |  | ns |
| Clock High Period | PWCLK (H) |  | 4 |  |  | ns |
| Data Setup Time | tsetup1 |  | 4 |  |  | ns |
| Data Hold Time | thold |  | 0 |  |  | ns |
| Start Pulse Setup Time | tsetup2 |  | 4 |  |  | ns |
| Start Pulse Hold Time | thold2 |  | 0 |  |  | ns |
| INV Setup Time | tsetup4 |  | 4 |  |  | ns |
| INV Hold Time | thold4 |  | 0 |  |  | ns |
| Start Pulse Low Period | tspL |  | 2 |  |  | CLK |
| Start Pulse Rise Time | tspr | 384 outputs |  | 128 |  | CLK |
|  |  | 402 outputs |  | 134 |  | CLK |
| STB Setup Time | tsetup3 |  | 1 |  |  | CLK |
| STB Pulse Width | PW stb $^{\text {st }}$ |  | 2 |  |  | CLK |
| Data Invalid Period | tinv |  |  | 1 |  | CLK |
| Last Data Timing | tLDT |  |  |  | 1 | CLK |
| CLK-STB Time | tclk-stb | CLK $\uparrow \rightarrow$ STB $\uparrow$ | 7 |  |  | ns |
| STB-CLK Time | tstb-clk | STB $\uparrow \rightarrow$ CLK $\uparrow$ | 7 |  |  | ns |



## 10. RECOMMENDED MOUNTING CONDITIONS

The following conditions must be met for mounting conditions of the $\mu$ PD16647.
For more details, refer to the Semiconductor Device Mounting Technology Manual(C10535E).
Please consult with our sales offices in case other mounting process is used, or in case the mounting is done under different conditions.
$\mu$ PD16647N-xxx : TCP(TAB Package)

| Mounting Condition | Mounting Method | Condition |
| :--- | :--- | :--- |
| Thermocompression | Soldering | Heating tool 300 to $350{ }^{\circ} \mathrm{C}$, heating for 2 to 3 sec ; pressure 100 g (per <br> solder) |
|  | ACF | Temporary bonding 70 to $100{ }^{\circ} \mathrm{C} ;$ pressure 3 to $8 \mathrm{~kg} / \mathrm{cm}^{2} ;$ time 3 to 5 <br> (Adhesive Conductive <br> sec. Real bonding 165 to $180^{\circ} \mathrm{C}$ pressure 25 to $45 \mathrm{~kg} / \mathrm{cm}^{2}$ time 30 to <br> 40secs(When using the anisotropy conductive film SUMIZAC1003 of <br> Sumitomo Bakelite,Ltd). |

Caution To find out the detailed conditions for mounting the ACF part, please contact the ACF manufacturing company. Be sure to avoid using two or more mounting methods at a time.

## (1) PRECAUTION AGAINST ESD FOR SEMICONDUCTORS

Note:
Strong electric field, when exposed to a MOS device, can cause destruction of the gate oxide and ultimately degrade the device operation. Steps must be taken to stop generation of static electricity as much as possible, and quickly dissipate it once, when it has occurred. Environmental control must be adequate. When it is dry, humidifier should be used. It is recommended to avoid using insulators that easily build static electricity. Semiconductor devices must be stored and transported in an anti-static container, static shielding bag or conductive material. All test and measurement tools including work bench and floor should be grounded. The operator should be grounded using wrist strap. Semiconductor devices must not be touched with bare hands. Similar precautions need to be taken for PW boards with semiconductor devices on it.
(2) HANDLING OF UNUSED INPUT PINS FOR CMOS

Note:
No connection for CMOS device inputs can be cause of malfunction. If no connection is provided to the input pins, it is possible that an internal input level may be generated due to noise, etc., hence causing malfunction. CMOS devices behave differently than Bipolar or NMOS devices. Input levels of CMOS devices must be fixed high or low by using a pull-up or pull-down circuitry. Each unused pin should be connected to VDD or GND with a resistor, if it is considered to have a possibility of being an output pin. All handling related to the unused pins must be judged device by device and related specifications governing the devices.

## (3) STATUS BEFORE INITIALIZATION OF MOS DEVICES

Note:
Power-on does not necessarily define initial status of MOS device. Production process of MOS does not define the initial operation status of the device. Immediately after the power source is turned ON, the devices with reset function have not yet been initialized. Hence, power-on does not guarantee out-pin levels, I/O settings or contents of registers. Device is not initialized until the reset signal is received. Reset operation must be executed immediately after power-on for devices having reset function.

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Standard: Computers, office equipment, communications equipment, test and measurement equipment, audio and visual equipment, home electronic appliances, machine tools, personal electronic equipment and industrial robots
Special: Transportation equipment (automobiles, trains, ships, etc.), traffic control systems, anti-disaster systems, anti-crime systems, safety equipment and medical equipment (not specifically designed for life support)
Specific: Aircraft, aerospace equipment, submersible repeaters, nuclear reactor control systems, life support systems or medical equipment for life support, etc.
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